

▶ POWER UNIT

MOHAMMED ABD-ALMAJEEED  
LOGIC

SHYAM MAWAJDEH

 POWER UNIT-JU



<http://POWER UNIT-JU.COM>

\* Raghad Imad Jaber. \*

" Shyam khated Nawaf kh "

# Logic and Computer Design Fundamentals

## Chapter 1 – Digital Systems and Information

" Shyam Nawaf kh "

Charles Kime & Thomas Kaminski

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# NUMBER SYSTEMS - Representation

(تصنيف - والترقعة العددية)

الأساس

Positive radix, positional number systems

A number with radix  $r$  is represented by a string of digits:

$$\boxed{A_{n-1}} \boxed{A_{n-2}} \dots \boxed{A_1} \boxed{A_0} \cdot \boxed{A_{-1}} \boxed{A_{-2}} \dots \boxed{A_{-m}}$$

(MSD)  $\leftarrow$   $A_{n-1}$  : أعلى قيمة موجودة  
 $\leftarrow$   $A_0$  : أول رقم  
 $\leftarrow$   $A_{-m}$  : أقل قيمة موجودة

فاصلة عشرية : radix-point

in which  $0 \leq A_i < r$  and  $r$  is the (radix point) or (Decimal-point) (LSD)

- $i$  represents the position of the coefficient (المرتبة)
- $r^i$  represents the weight by which the coefficient is multiplied (الوزن المرفقة)
- $A_{n-1}$  is the most significant digit (MSD) and  $A_{-m}$  is the least significant digit (LSD)
- The string of digits represents the power series:

في حالة وجود كسور عشرية

أعلى قيمة خاصة (المرتبة ذات أعلى وزن)

أقل قيمة خاصة (المرتبة ذات أقل وزن)

$$\left( \sum_{i=0}^{n-1} A_i r^i \right) + \left( \sum_{j=-m}^{-1} A_j r^j \right)$$

$$\sum_{i=-m}^{n-1} A_i r^i \text{ (Number)}_r$$

مجموع

- \* radix : الأساس
- \* Decimal system : النظام العشري
- \* radix-point : فاصلة عشرية
- \* Binary-point : فاصلة النظام الثنائي

# Number Systems - Examples

	General	Decimal (النظام العشري) (r=10)	Binary (النظام الثنائي) (r=2)
Radix (Base)	r	10	2
Digits	$0 \rightarrow r-1$ $(0 \leq A_i < r)$	$0 \rightarrow 9$ $(0 \leq A_i < r-1)$	$0 \rightarrow 1$
Powers of Radix	$r^0$ $r^1$ $r^2$ $r^3$ $r^4$ $r^5$ $r^{-1}$ $r^{-2}$ $r^{-3}$ $r^{-4}$ $r^{-5}$	$10^0 = 1$ $10^1 = 10$ $10^2 = 100$ $10^3 = 1000$ $10^4 = 10,000$ $10^5 = 100,000$ $10^{-1} = 0.1 = \frac{1}{10}$ $10^{-2} = 0.01 = \frac{1}{100}$ $10^{-3} = 0.001 = \frac{1}{1000}$ $10^{-4} = 0.0001 = \frac{1}{10000}$ $10^{-5} = 0.00001 = \frac{1}{100000}$	$2^0 = 1$ $2^1 = 2$ $2^2 = 4$ $2^3 = 8$ $2^4 = 16$ $2^5 = 32$ $2^{-1} = 0.5 = \frac{1}{2}$ $2^{-2} = 0.25 = \frac{1}{4}$ $2^{-3} = 0.125 = \frac{1}{8}$ $2^{-4} = 0.0625 = \frac{1}{16}$ $2^{-5} = 0.03125 = \frac{1}{32}$

رقم (عدد الخانات)  
 في النظام الذي  
 نكتبها (رقم) 10

# Example

$$1. \quad (403)_5 = 4 \times 5^2 + 0 \times 5^1 + 3 \times 5^0 = (103)_{10}$$

نوع النظام (خماسي) وهو نفسه radix (الأساس).

\* digits must be: : ثلاثة

$$0 \leq A_i \leq 4.$$

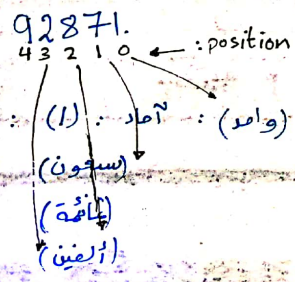
or

$$0 \leq A_i < 5.$$

(605)<sub>7</sub>  
نوع النظام (سبعي)  
2 \* 6 + 1 \* 0 + 5 \* 1 = ( )<sub>10</sub>

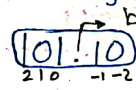
$$2. \quad (103)_{10} = 1 \times 10^2 + 0 \times 10^1 + 3 \times 10^0 = 103$$

\* Example (1):



10<sup>0</sup> : 1 : (واحد)  
10<sup>1</sup> : (10) : (عشرات)  
10<sup>2</sup> : (100) : (مئات)  
10<sup>3</sup> : (1000) : (ألفين)  
... وهكذا

\* Example (2): change from binary system to decimal system:



$$2^2 * 1 + 2^1 * 0 + 2^0 * 1 + 2^{-1} * 1 + 2^{-2} * 0$$

$$4 + 1 + 0.5 =$$

5.5 in decimal.

## BASE CONVERSION - Positive Powers of 2

Useful for Base Conversion : مفيد للتحويل بين الأنظمة \*

\* ما عليه نجمة (\*)  
- يجب حفظه ضروري  
لتسهيل الحل ...

Exponent	Value
* $0 = 2^0$	1
* $1 = 2^1$	2
* $2 = 2^2$	4
* $3 = 2^3$	8
* $4 = 2^4$	16
* $5 = 2^5$	32
* $6 = 2^6$	64
* $7 = 2^7$	128
* $8 = 2^8$	256
* $9 = 2^9$	512
* $10 = 2^{10}$	1024

Exponent	Value
* $11 = 2^{11}$	2,048
12	4,096
13	8,192
14	16,384
15	32,768
16	65,536
17	131,072
18	262,144
19	524,288
* $20 = 2^{20}$	1,048,576
21	2,097,152

$$2^{14} = \frac{2^{10}}{K} \cdot 2^4 = 16K$$

$$* 2^{30} : 1073741824$$

$$* 2^{40} : 1.09 * 10^{12}$$

$$* 2^{50} : 1.1 * 10^{15}$$

M  
صلى

# Special Powers of 2

\* memorize.

\*  $2^{10}$  (1024) is Kilo, denoted 'K'

( $2^{10}$ ) وهو (1024) هو (1000) الـ وأجزء عدد الـ 1000 = الكيلو ←  
رؤس الـ

\* Example: 20 KB  $\Rightarrow$   $20 \times 1024$  byte.

\*  $2^{20}$  (1,048,576) is Mega, denoted 'M'

\*  $2^{30}$  (1,073,741,824) is Giga, denoted 'G'

\*  $2^{40}$  (1,099,511,627,776) is Tera, denoted 'T'

\* NOTE: byte = 8 bite . البتة = 8 بايت

# Commonly Occurring Bases

2 1 0  
- - -  
b  
نوع عشري

Name	Radix	Digits
ثنائي Binary	2	0, 1
الثماني Octal	8	$0 \leq A_i \leq (r-1)$ 0, 1, 2, 3, 4, 5, 6, 7
العشري Decimal	10	0, 1, 2, 3, 4, 5, 6, 7, 8, 9
السادس عشري Hexadecimal	16	0, 1, 2, 3, 4, 5, 6, 7, 8, 9, A, B, C, D, E, F

The six letters A, B, C, D, E, and F represent the digits for values 10, 11, 12, 13, 14, 15 (given in decimal), respectively, in hexadecimal. Alternatively, a, b, c, d, e, f can be used.

السادس عشري

بدلاً من ذلك

$$(15) = (r-1)$$



# Binary System

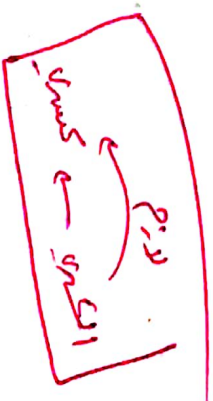
النظام الثنائي :-

- $r = 2$  الأعداد 2
- Digits =  $\{0, 1\}$  و  $\{false, True\}$  و  $\{low, high\}$  و  $\{off, on\}$  و  $\{0, 1\}$  من صفة للنظام الثنائي فقط ! تشكل قيمة (الرقم) قيمة تسمى (بت). (1, 0)
- \* Every binary digit is called a bit \*  
من صفة للنظام الثنائي فقط ! تشكل قيمة (الرقم) قيمة تسمى (بت). (1, 0)
- When a bit is equal to zero, it does not contribute to the value of the number  
يمكن استناد ال (0) من البداية قبل حل السؤال والتحويل للنظام العشري لأن الصفر عند مرتبة أي قيمة يكون الجواب النهائي = صفر.
- Example:
  - $(10011.101)_2 = (1 \times 2^4 + 0 \times 2^3 + 0 \times 2^2 + 1 \times 2^1 + 1 \times 2^0) + (1 \times 2^{-1} + 0 \times 2^{-2} + 1 \times 2^{-3})$  دون الاستناد.
  - $(10011.101)_2 = (16 + 2 + 1) + \left(\frac{1}{2} + \frac{1}{8}\right) = (19.625)_{10}$  بالاستناد.

# Octal System

النظام الثماني:

- $r = 8$
- Digits =  $\{0, 1, 2, 3, 4, 5, 6, 7\}$ :  $(0 \leq A_i \leq r-1)$ : القيمة
- Every digit is represented by 3-bits → More compact than binary \* كل قيمة رقمية في النظام الثماني تقابل (3 قيم رقمية) في النظام الثنائي.
- Example:
  - $(127.4)_8 = (1 \times 8^2 + 2 \times 8^1 + 7 \times 8^0) + (4 \times 8^{-1})$
  - $(127.4)_8 = (64 + 16 + 7) + \left(\frac{1}{2}\right) = (87.5)_{10}$



# Hexadecimal System: النظام الساسي عشري

- $r = 16$
- Digits =  $\{0, 1, 2, 3, 4, 5, 6, 7, 8, 9, \overset{10}{A}, \overset{11}{B}, \overset{12}{C}, \overset{13}{D}, \overset{14}{E}, \overset{15}{F}\}$  ( $0 \leq A_i \leq 15$ )
- Every digit is represented by 4-bits  
\* كل قيمة رقمية في النظام الساسي عشري تعادل (ع قيمة رقمية) في النظام الساسي.
- Example:
  - $(B65F)_{16} = (11 \times 16^3 + 6 \times 16^2 + 5 \times 16^1 + 15 \times 16^0)$
  - $(B65F)_{16} = (46687)_{10}$

\* انتبه! : تحوّل الرموز من النظام الساسي الى النظام العشري ووضعهم بعزّن العشر

\* الكحوليك من الأرقام العشرية المختلفة إلى النظام العشري

# Converting from any Base (r) to Decimal

$$(Number)_r = \left( \sum_{i=0}^{n-1} A_i r^i \right) + \left( \sum_{j=-m}^{-1} A_j r^j \right)$$

Integer Portion                      Fraction Portion

▪ Example: Convert  $(11010)_2$  to  $(N)_{10}$ :

\*  $(11010)_2 \rightarrow (?)_{10}$

$1 \times 2^4 + 2^3 \times 1 + 2^2 \times 0 + 2^1 \times 1 + 2^0 \times 0$

$16 + 8 + 2 = (26)_{10}$

# Conversion from Decimal to Base (r)

## Convert the Integer Part

## Convert the Fraction Part

Join the two results with a radix point

**\* Binary: (r=2)**

MSD ← 255

$(255)_{10} = (11111111)_2$   
 $(127)_{10} = (01111111)_2$   
 $(63)_{10} = (00111111)_2$   
 $(31)_{10} = (00011111)_2$   
 $(15)_{10} = (00001111)_2$   
 $(7)_{10} = (00000111)_2$   
 $(3)_{10} = (00000011)_2$   
 $(1)_{10} = (00000001)_2$   
 $(0)_{10} = (00000000)_2$

LSB →

**\* Octal: (r=8)**

MSD ← 255

$(255)_{10} = (377)_8$   
 $(127)_{10} = (177)_8$   
 $(63)_{10} = (77)_8$   
 $(31)_{10} = (37)_8$   
 $(15)_{10} = (17)_8$   
 $(7)_{10} = (7)_8$   
 $(3)_{10} = (3)_8$   
 $(1)_{10} = (1)_8$   
 $(0)_{10} = (0)_8$

LSB →

**\* Hexadecimal: (r=16)**

MSD ← 255

$(255)_{10} = (FF)_{16}$   
 $(127)_{10} = (7F)_{16}$   
 $(63)_{10} = (3F)_{16}$   
 $(31)_{10} = (1F)_{16}$   
 $(15)_{10} = (F)_{16}$   
 $(7)_{10} = (7)_{16}$   
 $(3)_{10} = (3)_{16}$   
 $(1)_{10} = (1)_{16}$   
 $(0)_{10} = (0)_{16}$

LSB →

# Conversion Details

لما نقسم الناتج حينئذ (لأنه) إنه الباقي من الباقي  
 الرقم الذي تكونه بنهاية الرقم الباقي  
 الرقم الذي تكونه بنهاية الرقم الباقي

10	15
0	1
1/2	2
1/2	4

لما نبدأ من

## To Convert the Integral Part:

1. Repeatedly divide the number by the new radix and save the remainders until the quotient is zero
2. The digits for the new radix are the remainders in reverse order of their computation
3. If the new radix is  $> 10$ , then convert all remainders  $> 10$  to digits A, B, ...

- A → 10
- B → 11
- C → 12
- D → 13
- E → 14
- F → 15

لما نبدأ من

## To Convert the Fractional Part:

1. Repeatedly multiply the fraction by the new radix and save the integer digits of the results until the fraction is zero or you reached the required number of fractional digits
2. The digits for the new radix are the integer digits in order of their computation
3. If the new radix is  $> 10$ , then convert all integers  $> 10$  to digits A, B, ...

نحفظ الباقي

باقي

الباقي هو الباقي الباقي الباقي  
 من الرقم الباقي الباقي الباقي  
 الباقي الباقي الباقي

Ex:  $\frac{1}{2} = 0.5$

# Example: Convert $46.6875_{10}$ To Base 2

Convert 46 to Base 2:

$$(46)_{10} = (101110)_2$$

32 + 8 + 4 + 2 = 32 + 14 = (46)<sub>10</sub>

Convert  $0.6875$  to Base 2:

$$(0.6875)_{10} = (0.1011)_2$$

Division	Quotient	Remainder
46/2	23	0
23/2	11	1
11/2	5	1
5/2	2	1
2/2	1	0
1/2	0	1

Multiplication	Answer
$0.6875 * 2$	1.3750
$0.375 * 2$	0.75
$0.75 * 2$	1.5
$0.5 * 2$	1.0

Join the results together with the radix point:

$$(46.6875)_{10} = (101110.1011)_2$$

# Example: Convert $153.513_{10}$ To Base 8

- Convert 153 to Base 8:

$$(153)_{10} = (231)_8$$

Division	Quotient	Remainder
153/8	19	1
19/8	2	3
2/8	0	2

الناتج
الناتج

LSD
LSD

MSD
MSD

- Convert  $0.513$  to Base 8: (Up to 3 digits)

عدد المنازل المطلوبة بالجواب

Tuncate: ترك الباقي المضاف

$$(0.513)_{10} = (0.406)_8$$

التعريب

$$\text{Round: } (0.513)_{10} = (0.407)_8$$

تقريب العدد

- Join the results together with the radix point:

$$(153.513)_{10} = (231.407)_8$$

\* in the octal system (1/2) (3) (4) (5) (6) (7)



# Example: Convert $423_{10}$ To Base 16

Division	Quotient	Remainder
$423/16$	26	7
$26/16$	1	10 (A)
$1/16$	0	1

LSD ↑

MSD

\* نتیجاً - جب تحويل الأرقام من (10 ← 15) بالرموز  
المخصصة في النظام الساس عشرى .

$$(423)_{10} = (1A7)_{16}$$

الحل باستخدام  
مضاعفات الأعداد



# Converting Decimal to Binary:

## Alternative Method

يطلب بالاصحاح  
الاصحاح

ex: -  $(28)_{10} \rightarrow (11100)_2$

$$\begin{array}{r}
 16 + 12 \\
 16 + 8 + 4 \\
 2^{(4)} + 2^{(3)} + 2^{(2)} + 2^2
 \end{array}$$

$$\begin{array}{r}
 5 \ 4 \ 3 \ 2 \ 1 \ 0 \\
 \downarrow \downarrow \downarrow \downarrow \downarrow \\
 1 \ 1 \ 1 \ 0 \ 0
 \end{array}$$

1. Subtract the largest power of 2 that gives a positive remainder and record the power

بشكل الرقم 1 اكبر عدد من مضاعفات ال 2 اقل  
من الرقم الاكبر من المضاعفات ال 2 اقل عدد

2. Repeat, subtracting from the prior remainder and recording the power, until the remainder is zero

3. Place 1's in the positions in the binary result corresponding to the powers recorded; in all other positions place 0's

$$(0.6875)_{10} = (10110)_2$$

\* Note:  $(0.4075)_8$  octal system.  $(7 \leftarrow 0)$  لأن الخانات من 0 إلى 7

$(0.410)_8$  لأن ال (7) لا يصح (8) لأنه نظام ثنائي ذلك نصف (1) للثنائية  
الثنائية ونصفها (عشر)

$$\begin{array}{r}
 0.5 + 0.1875 \\
 0.15 + 0.125 + 0.0625 \\
 \downarrow \quad \downarrow \quad \downarrow \\
 2^{-1} \quad 2^{-2} \quad 2^{-3} \quad 2^{-4} \quad 2^{-5} \\
 1 \quad 0 \quad 1 \quad 1 \quad 0
 \end{array}$$

# Example: Convert $46.6875_{10}$ To Base 2

## Using Alternative Method

Convert 46 to Base 2:

$$(46)_{10} = (101110)_2$$

5 4 3 2 1 0  
 1 0 1 1 1 0  
 (1) صواب (0) خطأ  
 رفع بر (1) صواب (0) خطأ  
 ممكن ان يكون رتبة أصغر

Convert 0.6875 to Base 2:

$$(0.6875)_{10} = (0.1011)_2$$

Join the results together with the radix point:

$$(46.6875)_{10} = (101110.1011)_2$$

Easier way to do it:

	64	32	16	8	4	2	1	.	1	1	1	1
Power	6	5	4	3	2	1	0	.	-1	-2	-3	-4
	0	1	0	1	1	0	0	.	1	0	1	1

Subtract	Remainder	Power
$46 - 32 = 14$	14	5
$14 - 8 = 6$	6	3
$6 - 4 = 2$	2	2
$2 - 2 = 0$	0	1

Subtract	Remainder	Power
$0.6875 - 0.500$	0.1875	-1
$0.1875 - 0.125$	0.0625	-3
$0.0625 - 0.0625$	0	-4

\* ملاحظة: إذا كان الرقم فردي لا بد  
 أول منزلة تكون (1) بينما زوجي أول منزلة  
 تكون (0) في النظام الثنائي.  
 زوجي:  $10110$   
 فردي:  $10111$

# Additional Issue - Fractional Part

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- Note that in this conversion, the fractional part can become 0 as a result of the repeated multiplications
- In general, it may take many bits to get this to happen or it may never happen
- Example Problem: Convert  $0.65_{10}$  to  $N_2$ 
  - $0.65 = 0.1010011001001\dots$  .البروي
  - The fractional part begins repeating every 4 steps yielding repeating 1001 forever!
- **Solution:** Specify number of bits to right of radix point and round or truncate to this number

# Checking the Conversion

- To convert back, sum the digits times their respective powers of  $r$

- From the prior conversion of  $46.6875_{10}$

$$\begin{aligned} 101110_2 &= 1*32 + 0*16 + 1*8 + 1*4 + 1*2 + 0*1 \\ &= 32 + 8 + 4 + 2 \\ &= 46 \end{aligned}$$

$$\begin{aligned} 0.1011_2 &= 1/2 + 1/8 + 1/16 \\ &= 0.5000 + 0.1250 + 0.0625 \\ &= 0.6875 \end{aligned}$$

# Octal (Hexadecimal) to Binary and Back: Method 1

## Octal (Hexadecimal) to Binary:

1. Convert octal (hexadecimal) to decimal (Slide 23)
2. Convert decimal to binary (Slide 24 or Slide 29)

الكلاس الأولى  
الكلاس الثانية

## Binary to Octal (Hexadecimal):

1. Convert binary to decimal (Slide 23)
2. Convert decimal to octal (hexadecimal) (Slide 24)

$r = 8 = 2^3$  3-bits in binary system:

000	: 0
001	: 1
010	: 2
011	: 3
100	: 4
101	: 5
110	: 6
111	: 7

hex

$r = 16 = 2^4$  4-bits in binary system:

0000	: 0
0001	: 1
0010	: 2
0011	: 3
0100	: 4
0101	: 5
0110	: 6
0111	: 7
1000	: 8
1001	: 9

4-bits in binary system:

1010	: A
1011	: B
1100	: C
1101	: D
1110	: E
1111	: F

hex

Base (r) .  
or  
octal (r = 8) .  
or  
hexadecimal (r = 16) .

↓ ↓  
↓ ↑  
Decimal (r = 10) .  
↓ ↓  
Binary (r = 2) .

## Octal (Hexadecimal) to Binary and Back: Method 2 (Easier)

octal (r=8) to hexadecimal (r=16).  
binary (r=2)

- Octal (Hexadecimal) to Binary:
  - Restate** the octal (hexadecimal) as three (four) binary digits starting at the radix point and going both ways
- Binary to Octal (Hexadecimal):
  - Group** the binary digits into three (four) bit groups starting at the radix point and going both ways, padding with zeros as needed
  - Convert each group of three (four) bits to an octal (hexadecimal) digit

Octal	0	1	2	3	4	5	6	7
Binary	000	001	010	011	100	101	110	111

Hexadecimal	0	1	2	3	4	5	6	7
Binary	0000	0001	0010	0011	0100	0101	0110	0111
Hexadecimal	8	9	A	B	C	D	E	F
Binary	1000	1001	1010	1011	1100	1101	1110	1111

ملاحظة: - إذا لم يكونا

النظامين المطلوب التحويل بينهما

قوة الأساس (2) نقول بالتحويل

للنظام العشري (decimal) ومن

ثم نحول للنظام المطلوب.

# Examples

$$\blacksquare (673.12)_8 = (110\ 111\ 011\ .\ 001\ 010)_2$$

من 2 الى 8 ضويف على منازل  
من 2 الى 16 نبوظ 5 منازل

$$\blacksquare (3A6.C)_{16} = (0011\ 1010\ 0110\ .\ 1100)_2$$

$$\blacksquare (10110001101011.11110000010)_2 = (?)_8$$

\* في آ التقسيم من ارقامه  
ذهاذا للبين و اليسار

$$(10110/001/101/011.111/100/000/1)_2 = (26153.7404)_8$$

$$\blacksquare (10110001101011.11110000010)_2 = (?)_{16}$$

$$(101100/0110/1011.1111/0000/01)_2 = (2C6B.F04)_{16}$$



# Octal to Hexadecimal via Binary

- 1. Convert octal to binary
- Use groups of four bits and convert to hexadecimal digits
- Example: Octal to Binary to Hexadecimal

$(635.177)_8$   
Octal:  $(110\ 011\ 101.\ 001\ 111\ 111)_2$

Binary:  $0001/1001/1101.\ 0011/1111/1000$

Hexadecimal:  $(19D.3F8)_{16}$

# One last Conversion Example

- Given that  $(365)_r = (194)_{10}$ , compute the value of  $r$ ?

$$3 \times r^2 + 6 \times r^1 + 5 \times r^0 = 194$$

$$3r^2 + 6r + 5 = 194$$

$$3r^2 + 6r - 189 = 0$$

$$r^2 + 2r - 63 = 0$$

\* عبارة تربيعية -  
حلها

$$(r - 7)(r + 9) = 0$$

\* الحل في (Radix) الكوبي

$$r = 7$$

\* هو و هو الحل الثاني

# Binary Numbers and Binary Coding

⊗ تلاحظ : دائماً كتابة ال (codes) باستخدام ال (binary system).

## ▪ Flexibility of representation

- Within constraints below, can assign any binary combination (called a code word) to any data as long as data is uniquely encoded

⊗ لمعرفة عدد ال (codes) اللازمة :

## ▪ Information Types

$n$ -bits  $\rightarrow 2^n$  . (using binary system,  $r=2$ )

- Numeric : عددية / رقمية
  - Must represent range of data needed
  - Very desirable to represent data such that simple, straightforward computation for common arithmetic operations permitted
  - Tight relation to binary numbers
- Non-numeric : غير عددية / غير رقمية
  - Greater flexibility since arithmetic operations not applied
  - Not tied to binary numbers

# Non-numeric Binary Codes

Given  $n$  binary digits (called bits), a binary code is a mapping from a set of represented elements to a subset of the  $2^n$  binary numbers.

\* لمعرفة عدد الـ (Codes) اللازمه :

$$\left[ \begin{array}{l} \text{عدد الـ digits} \\ \text{2}^n \\ \text{n-bits} \end{array} \right]$$

او اي السائل آلم (صن النظام المعلوم)

Example: A binary code for the seven colors of the rainbow

\* Code 100 is not used صاحه ولكن لم تستخدم عادي

عندي 7 خيارات  
خيار اي رقم ليس ايام عادي

Color	Binary Number
Red	000
Orange	001
Yellow	010
Green	011
Blue	101
Indigo	110
Violet	111

# Number of Bits Required

base = 2. (binary)

$n =$  عدد الأرقام = 2. ---

$M =$  عدد العناصر = عدد الأكواد =  $\frac{1}{2}$

(Number of elements)

- Given  $M$  elements to be represented by a binary code, the minimum number of bits,  $n$ , needed, satisfies the following relationships:

if  $n=3$  .  $2^3 \geq M > 2^2$        $2^n \geq M > 2^{n-1}$

$(8 \geq 7 > 4)$

$n = \lceil \log_2 M \rceil$ , where  $\lceil x \rceil$  is called

the ceiling function, is the integer greater than or equal to  $x$ .

تقرّب الجواب لأعلى. يتم تطبيق أكبر عدد

عدد البت المطلوبة :-

- Example: How many bits are required to represent decimal digits with a binary code?

radix = 2

عدد الأرقام للنظام العشري

$M = 10$

\*if:  $r=2$  .  $n=7$ .

\* also:  $r = *$   
 $M = *^n$  . (radix) نفس

$M = 2^7 = 128$ .  
 (maximum possible number of elements):

$n = \lceil \log_2 10 \rceil = \lceil 3.33 \rceil = 4$

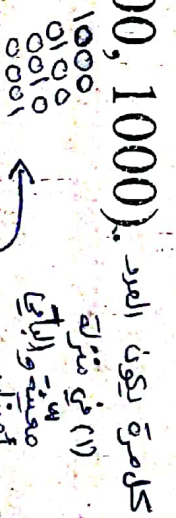
تقرّب للعدد الأكبر

# Number of Elements Represented

Given  $n$  digits in radix  $r$ , there are  $r^n$  distinct elements that can be represented.

But, you can represent  $m$  elements,  $m \leq r^n$

Examples:  
 You can represent 4 elements in radix  $r = 2$  with  $n = 2$  digits: (00, 01, 10, 11).  
 You can represent 4 elements in radix  $r = 2$  with  $n = 4$  digits: (0001, 0010, 0100, 1000).



You can represent 4 elements in radix  $r = 2$  with  $n = 4$  digits: (0001, 0010, 0100, 1000).  
 This second code is called a "one hot" code.

$n = \lceil \log_r M \rceil$ .  
 EX:  $n = \lceil \log_2 7 \rceil = 3$ .  
 EX:  $n = \lceil \log_2 365 \rceil = 9$ .

$n$ : minimum number of bits to represent  $(M)$ .

\* يمكن أيضاً معرفة أكبر قيمة (value) نصل إليها / نختارها :  
 if  $r = 10$  } then  $r^n - 1 = 100 - 1 = 99$ . the biggest value.  
 if  $n = 2$

وإذا نريد أكبر عدد واحد. (11) = 99

# DECIMAL CODES - Binary Codes for Decimal Digits (BCD)

\* السؤال الذي يجي عليه  
 كالتالي: لترتيب الرقم (10010)  
 صوله لكما في الجدول.  
 +3  
 Excess 3  
 8, 4, -2, -1

There are over 8,000 ways that you can chose 10 elements from the 16 binary numbers of 4 bits. A few are useful.

قريب BCD من  
 نظر 2 بالوضع

Decimal	8, 4, 2, 1	Excess 3	8, 4, -2, -1	Gray
0	0000	0011	0000	0000
1	0001	0100	0111	0001
2	0010	0101	0110	0011
3	0011	0110	0101	0010
4	0100	0111	0100	0110
5	0101	1000	1011	1110
6	0110	1001	1010	1010
7	0111	1010	1001	1011
8	1000	1011	1000	1001
9	1001	1100	1111	1000

\* if we have a digital system  
 M=10: elements  
 r=2  
 n: must be (4)  
 because  $2^4 = 16$   
 لا نستطيع  
 ونستعمل  
 6  
 مستخدم

(6) codes will be unused or available.  
 0000 ← 0  
 0001 ← 1  
 0010 ← 2  
 0011 ← 3  
 ...  
 1001 ← 9  
 these are used codes but codes from (1010 → 1111) will not be used.

slide (44) =  
 \* (19)<sub>10</sub> → (10011)<sub>2</sub>  
 ∴ convert (تحويل)  
 \* (19)<sub>10</sub> → (0001 1001)  
 ∴ BCD coding (ترتيب)  
 \* أهمياً يتم استخدام  
 ال codes لكي نكتب  
 القيمة الحقيقية للرقم  
 حيث يعطي الرقم قيمة  
 تختلف عن قيمته  
 الحقيقية.

Binary coded decimal  
 بترتيب الثلاثة - 3  
 لكل منزلة من الرقم الأصلي  
 يعبر عن منزلة من D ال B

\* Ex: 1: 0 1 1 1 : 4-3 = 1  
 ... وهكذا ... 9: 1 1 1 1 : 12-3 = 9

# Binary Coded Decimal (BCD)

---

- Numeric code

- The BCD code is the 8, 4, 2, 1 code

- 8, 4, 2, and 1 are weights → BCD is a weighted code

- This code is the simplest, most intuitive binary code for decimal digits and uses the same powers of 2 as a binary number, but only encodes the first ten values from 0 to 9

- Example:  $1001 (9) = 1000 (8) + 0001 (1)$

- How many “invalid” <sup>(غير مستخدم كود)</sup> code words are there?

- Answer: 6

- What are the “invalid” code words?

- Answer:  $1010 (10)$ ,  $1011 (11)$ ,  $1100 (12)$ ,  $1101 (13)$ ,  $1110 (14)$ ,  $1111 (15)$



# Warning: Conversion or Coding?

التحويل

و

الترميز

- Do NOT mix up *conversion* of a decimal number to a binary number with *coding* a decimal number with a BINARY CODE.

13<sub>10</sub> = 1101<sub>2</sub> (This is conversion)  
تحويل من الأرقام العشرية إلى الثنائية

8 4 2 1  
12+1 = 13  
كيفية

(Red)

تحويل الأرقام العشرية إلى الثنائية

13 ⇔ 000110011 (This is coding)

8 4 2 1  
4 2 1  
1 3  
تقسيم

ترميز

Excess 3

# Excess 3 Code and 8, 4, -2, -1 Code

What interesting property is common to these two codes?

Answer: Both codes have the property that the codes for 0 and 9, 1 and 8, etc. can be obtained from each other by replacing the 0's with the 1's and vice-versa. Such a code is sometimes called a

complement code.

مكمل  
 (0) → 1  
 (1) → 0

Decimal	Excess 3	8, 4, -2, -1
0	0011	0000
1	0100	0111
2	0101	0110
3	0110	0101
4	0111	0100
5	1000	1011
6	1001	1010
7	1010	1001
8	1011	1000
9	1100	1111

# ALPHANUMERIC CODES - ASCII Character Codes

- Non-numeric code

⊗ ASCII code :-  
if:  $n=7$ ,  $r=2$  } then:

Maximum number of elements =  $2^7 = 128$ .

7 bits

- ASCII stands for American Standard Code for Information Interchange (Refer to Table 1-5 in the text)

- This code is a popular code used to represent information sent as character-based data. It uses 7-bits (i.e. 128 characters) to represent:  $2^7 = 128$ .

- 95 Graphic printing characters
- 33 Non-printing characters

# ASCII Code Table (C++)

Least Significant

ASCII Code Chart

0	1	2	3	4	5	6	7	8	9	A	B	C	D	E	F
NUL	SOH	STX	ETX	EOT	ENO	ACK	BEL	BS	HT	LF	VT	FF	CR	SO	SI
DLE	DC1	DC2	DC3	DC4	NAK	SYN	ETB	CAN	EM	SUB	ESC	FS	GS	RS	US
	!	"	#	\$	%	&	'	(	)	*	+	,	-	.	/
0	1	2	3	4	5	6	7	8	9	:	;	<	=	>	?
@	A	B	C	D	E	F	G	H	I	J	K	L	M	N	O
P	Q	R	S	T	U	V	W	X	Y	Z	[	\	]	^	_
.	a	b	c	d	e	f	g	h	i	j	k	l	m	n	o
P	q	r	s	t	u	v	w	x	y	z	{		}	~	DEL

Most Significant

Small  $N = (77)_6 = (01110111)_2$

$M = (57)_{16} = (01010111)_2$

$A = (41)_6 = (01000001)_2$

$a = (61)_{16} = (01100001)_2$

# ASCII Character Codes

---

- Graphic printing characters

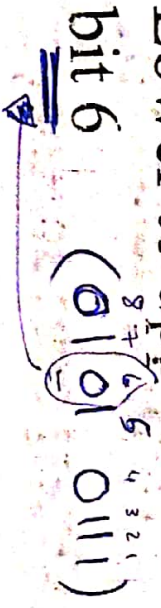
- 26 upper case letters (A-Z) *Capital*
- 26 lower case letters (a-z) *Small*
- 10 numerals (0-9)
- 33 special characters (e.g. %, @, \$) *signs.*

- Non-printing characters

- Format effectors: used for text format (e.g. BS = Backspace, CR = carriage return)
- Information separators: used to separate the data into paragraphs and pages (e.g. RS = record separator, FS = file separator)
- Communication control characters (e.g. STX and ETX start and end text areas).

# ASCII Properties

---

- ASCII has some interesting properties:
    - Digits 0 to 9 span Hexadecimal values  $30_{16}$  to  $39_{16}$
    - Upper case A-Z span  $41_{16}$  to  $5A_{16}$
    - Lower case a-z span  $61_{16}$  to  $7A_{16}$
    - Lower to upper case translation (and vice versa) occurs by flipping bit 6
- ( $0101011$ )
- 

# UNICODE

- UNICODE <sup>توسيع</sup> extends ASCII to 65,536 <sup>عالمي</sup> universal characters codes:
  - Non-numeric (غير رقمي)
  - For encoding characters in world languages
  - Available in many modern applications
  - 2 byte (16-bit) code words

\* if:  $n = 16$

$r = 2$

$2^{16} = 65,536$

# PARITY BIT Error-Detection Codes

↳ (ECC) ↳

\* يبين عن الأخطاء الفرعية فقط و تستخدم الأخطاء فقط لا تقوم  
(تحدد مكان الأخطاء فقط) بتحديد مكان الأخطاء.

detected odd number of errors.  
(تحدد)

- Non-numeric
- <sup>موت</sup>Redundancy (e.g. extra information), in the form of extra bits, can be incorporated into binary code words to detect and correct errors
- A simple form of redundancy is parity, an extra bit appended onto the code word to make the number of 1's odd or even. Parity can detect all single-bit errors and some multiple-bit errors
- A code word has even parity if the number of 1's in the code word is even
- A code word has odd parity if the number of 1's in the code word is odd





# 4-Bit Parity Code Example

Fill in the even and odd parity bits:

Even Parity Message	Odd Parity Message
0000 → parity bit.	000 <u>1</u>
001 <u>1</u>	001 <u>0</u>
010 <u>1</u>	010 <u>0</u>
011 <u>0</u>	011 <u>1</u>
100 <u>1</u>	100 <u>0</u>
101 <u>0</u>	101 <u>1</u>
110 <u>0</u>	110 <u>1</u>
111 <u>1</u>	111 <u>0</u>

\*Ex: 0001  
 ← even parity      → odd parity  
 0001      0001  
 لكن يصبح عدد (1) زوجي.  
 لكن يصبح عدد (1) فردي.

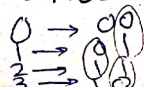
\* if we have:

0011 parity bit  
 ← parity=0 even parity 00110  
 → parity=1 odd parity 00111  
 عدد الواحدات زوجي لذلك الـ parity زوجي.  
 عدد الواحدات فردي لذلك الـ parity فردي.

The code word "1111" has even parity and the code word "1110" has odd parity. Both can be used to represent the same 3-bit data

\* Note to know:

write the gray codes for:  
 ليتم التمايز بين بت وبت واحد  
 فقط بين كل code و code  
 الا فردي



**Logic and Computer Design Fundamentals**

# **Chapter 2 – Combinational Logic Circuits**

**Part 1 – Gate Circuits and Boolean Equations**

**Charles Kime & Thomas Kaminski**

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(Hyperlinks are active in View Show mode)

# Combinational Logic Circuits

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- Digital (logic) circuits are hardware components that manipulate binary information.
  - Integrated <sup>دوائر متكاملة</sup> circuits: transistors and interconnections.
  - Basic circuits is referred to as logic gates (دوائر منطقية)
  - The outputs of gates are applied to the inputs of other gates to form a digital circuit
- Combinational? Later...

# Overview

---

- **Part 1 – Gate Circuits and Boolean Equations**<sup>قواعد</sup>
  - Binary Logic and Gates
  - Boolean Algebra
  - Standard Forms
- **Part 2 – Circuit Optimization**
  - Two-Level Optimization
  - Map Manipulation <sup>خريطة</sup>
  - Practical Optimization (Espresso)
  - Multi-Level Circuit Optimization
- **Part 3 – Additional Gates and Circuits**
  - Other Gate Types
  - Exclusive-OR Operator and Gates
  - High-Impedance Outputs

# Binary Logic and Gates

---

- **Binary variables** take on one of two values
- **Logical operators** operate on binary values and binary variables
- Basic logical operators are the logic functions **AND**, **OR** and **NOT**  
⊗ Basic operator gates:  
① AND:  $\cdot, \wedge$     ③ NOT:  $\sim, \neg, -$   
② OR:  $+, \vee$
- **Logic gates** implement logic functions
- **Boolean Algebra**: a useful mathematical system for specifying and transforming logic functions
- We study Boolean algebra as a foundation for designing and analyzing digital systems!

# Binary Variables

---

- Recall that the two binary values have different names:
  - True/False
  - On/Off
  - Yes/No
  - 1/0
- We use 1 and 0 to denote the two values
- Variable identifier examples:
  - A, B, y, z, or  $X_1$  for now
  - RESET, START\_IT, or ADD1 later

# Logical Operations

---

▪ The three basic logical operations are:

- AND
- OR
- NOT

لوجی عملیات (X)

\* ▪ AND is denoted by a dot (·) or (∧)

\* ▪ OR is denoted by a plus (+) or (∨)

\* ▪ NOT is denoted by an over-bar (¯), a single quote mark (') after, or (~) before the variable

$\bar{A} / A' / \sim A$

# Notation Examples

## Examples:

$Z = X \cdot Y = XY = X \wedge Y$ : is read "Z is equal to X AND Y"

$Z = 1$  if and only if  $X = 1$  and  $Y = 1$ ; otherwise,  $Z = 0$

$Z = X + Y = X \vee Y$ : is read "Z is equal to X OR Y"

$Z = 1$  if (only  $X = 1$ ) or if (only  $Y = 1$ ) or if ( $X = 1$  and  $Y = 1$ )

$Z = \overline{X} = X'$ : is read "Z is equal to NOT X"

$Z = 1$  if  $X = 0$ ; otherwise,  $Z = 1$  (opposite always)

Notice the difference between arithmetic addition and logical OR:

The statement:  $1 + 1 = 2$  (read "one plus one equals two")  
 \*is not the same as  $1 + 1 = 1$  (read "1 or 1 equals 1")

(+) . 9031 01211

(1) 1 or 1 = 1  
 (2) 1 or 0 = 1  
 (3) 0 or 1 = 1  
 (4) 0 or 0 = 0

$Z(X \cdot Y) \rightarrow \text{true}$

(and) 1 plus 1 is 2



# Operator Definitions

- Operations are defined on the values "0" and "1" for each operator: معروف الأعداد

<b>AND</b>
$0 \cdot 0 = 0$
$0 \cdot 1 = 0$
$1 \cdot 0 = 0$
$1 \cdot 1 = 1$

<b>OR</b>
$0 + 0 = 0$
$0 + 1 = 1$
$1 + 0 = 1$
$1 + 1 = 1$

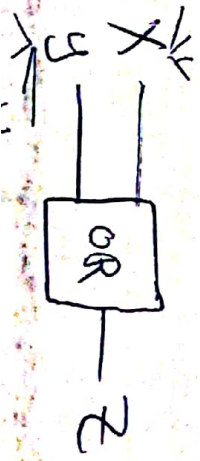
<b>NOT</b>
$\bar{0} = 1$
$\bar{1} = 0$

$r = 2$   
 $n = 2$  : } → then

$2^2 = 4$

المعاني

0 1 0 1



# Truth Tables

Handwritten scribbles and a small table of binary values:

0	0
0	1
1	0
1	1

- Truth table - a tabular listing of the values of a function for all possible combinations of values on its arguments

- Example: Truth tables for the basic logic operations:

AND

Inputs	Output
X Y	Z = X · Y
0 0	0
0 1	0
1 0	0
1 1	1

Handwritten notes: 0 →, 1 →, 2 →, 3 →, 4 →, 2<sup>2</sup> = 4 احتمالات

OR

Inputs	Output
X Y	Z = X + Y
0 0	0
0 1	1
1 0	1
1 1	1

Handwritten notes: 2<sup>2</sup> = 4 احتمالات

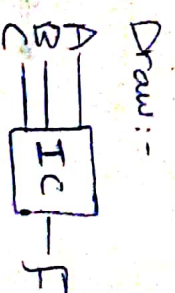
NOT

Inputs	Output
X	Z = $\bar{X}$
0	1
1	0

Handwritten notes: 1 = 2 احتمالات

Handwritten truth table for a 3-bit input:

MSB inputs:	least significant
A B C	LSbit
0 0 0	0
0 0 1	1
0 1 0	0
0 1 1	1
1 0 0	0
1 0 1	1
1 1 0	0
1 1 1	1

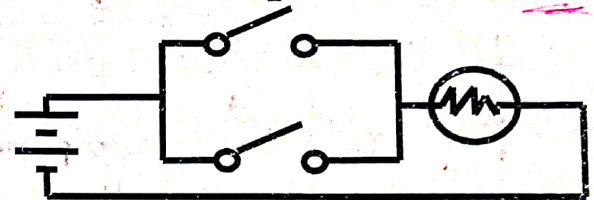


# Logic Function Implementation

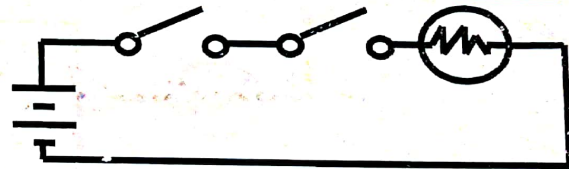
## Using Switches

- For inputs:
  - logic 1 is switch closed
  - logic 0 is switch open
- For outputs:
  - logic 1 is light on
  - logic 0 is light off
- NOT uses a switch such that:
  - logic 1 is switch open
  - logic 0 is switch closed

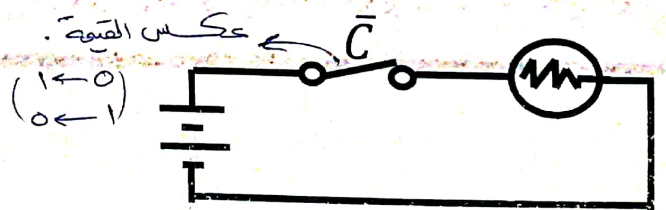
Switches in parallel => OR



Switches in series => AND



Normally-closed switch => NOT



\* عندما يكون المفتاح مغلق يعني (الدارة) (0)  
 \* عندما يكون المفتاح مفتوح يعني (الدارة) (1)

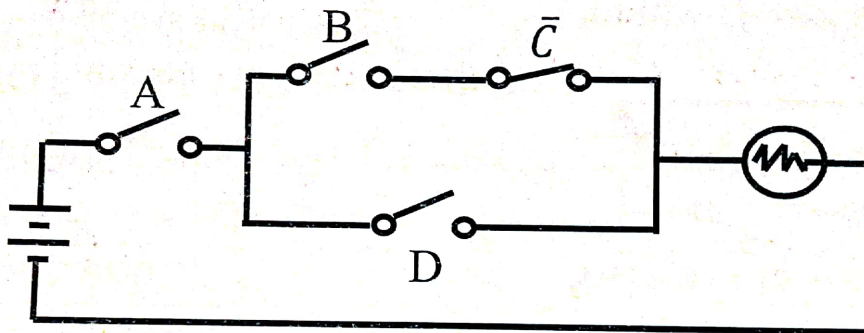
عكس القيمة  
 (1 ← 0)  
 (0 ← 1)

عكس الواقع لأن البوابة (not)

OR = 1

# Logic Function Implementation (Continued)

- Example: Logic Using Switches



- Light is

Formula ← by ← Circuit

Handwritten notes in Arabic: 'نقطة' and 'نقطة'.

$ON (L = 1)$  for  $L (A, B, C, D) = A \cdot (B\bar{C} + D) = AB\bar{C} + AD$   
 and  $OFF (L = 0)$ , otherwise.

↓  
*نقطة*  
 $A \cdot D + A \cdot B \cdot \bar{C}$

- Useful model for relay circuits and for CMOS gate circuits, the foundation of current digital logic technology

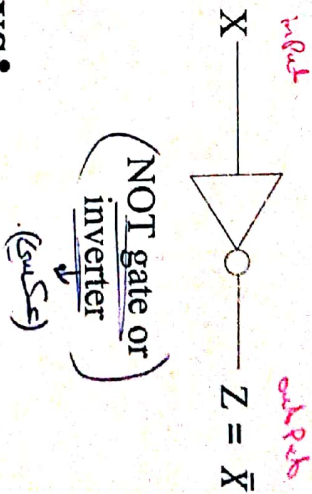
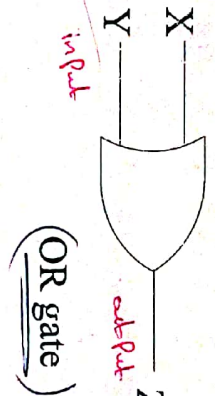
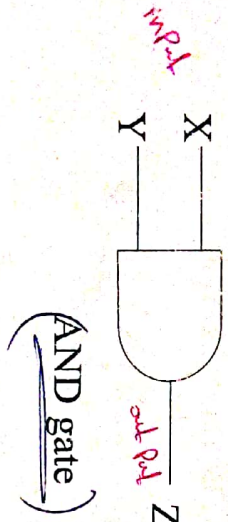
# Logic Gates

---

- In the earliest computers, switches were opened and closed by magnetic fields produced by energizing coils in *relays*. The switches in turn opened and closed the current paths
- Later, *vacuum tubes* that open and close current paths electronically replaced relays
- Today, *transistors* are used as electronic switches that open and close current paths
- **Optional:** Chapter 6 – Part 1: The Design Space

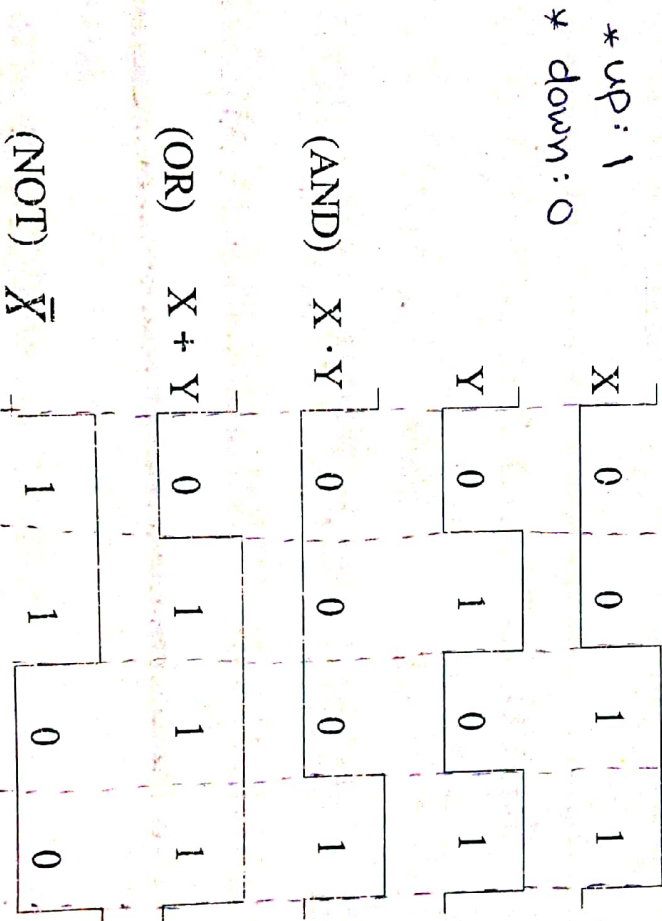
# Logic Gate Symbols and Behavior

Logic gates have special symbols: *محددات المنطق*

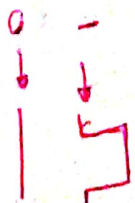


And waveform behavior in time as follows:

(a) Graphic symbols



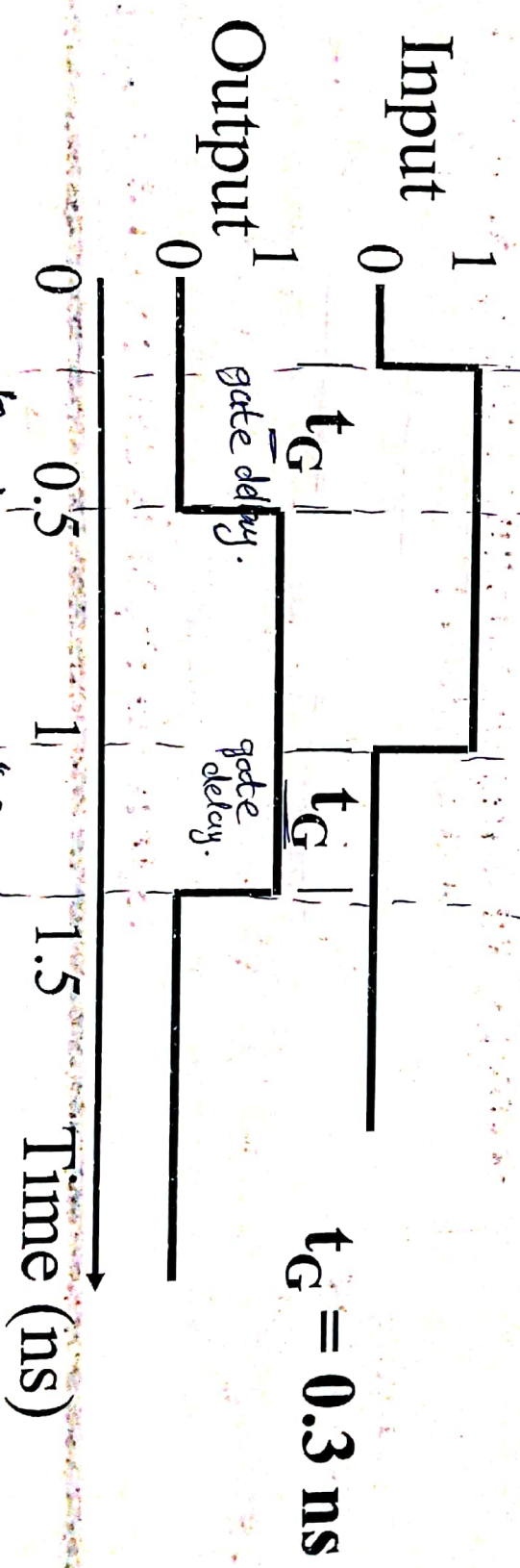
(b) Timing diagram



# Gate Delay

(سیرتایم)

- In actual physical gates, if one or more input changes causes the output to change, the output change does not occur (instantaneously) <sup>فوری</sup>
- The delay between an input change(s) and the resulting output change is the gate delay denoted by  $t_G$ :



$t_G = 0.5$  ns

$t_G = 0.5$  ns

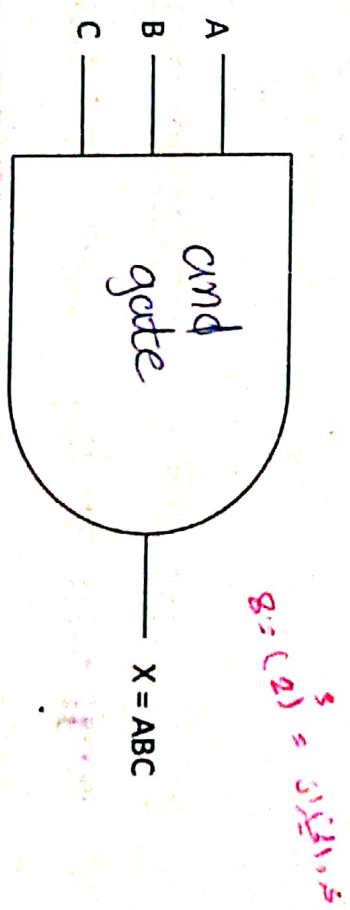
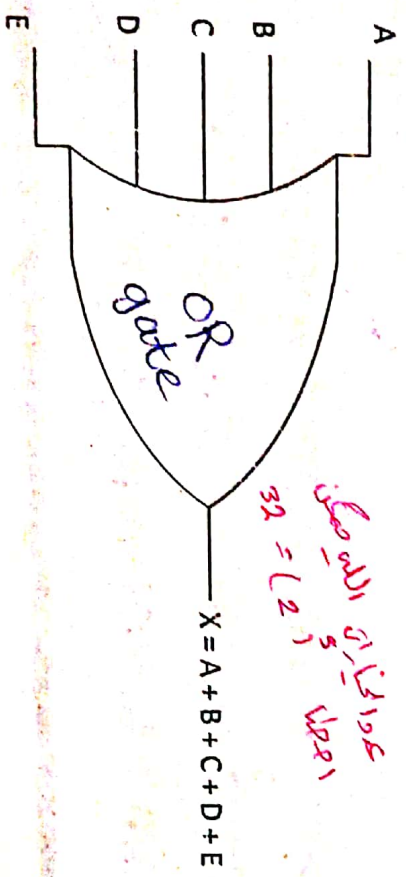
# Logic Gates: Inputs and Outputs

- NOT (inverter)

- Always one input and one output *Just*

- AND and OR gates

- Always one output
- Two or more inputs





# Boolean Algebra

عبدالله  
عبدالله



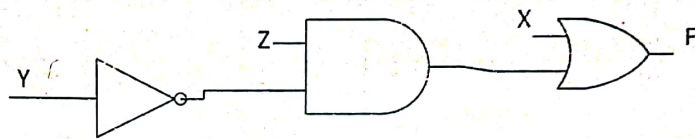
- An algebra dealing with binary variables and logic operations
  - Variables are designated by letters of the alphabet
  - Basic logic operations: AND, OR, and NOT
- A Boolean expression is an algebraic expression formed by using binary variables, constants 0 and 1, the logic operation symbols, and parentheses (المتغيرات)
- E.g.:  $X \cdot 1, A + B + C, (A + B)(C + D)$
- A Boolean function consists of a binary variable identifying the function followed by equals sign and a Boolean expression
  - E.g.:  $F = A + B + C, L(D, X, A) = DX + \bar{A}$

# Logic Diagrams and Expressions

( )  
Not  
AND  
or

1. Equation:  $F = X + \bar{Y}Z$

2. Logic Diagram:



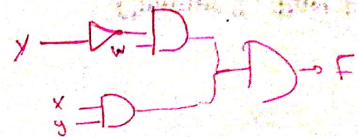
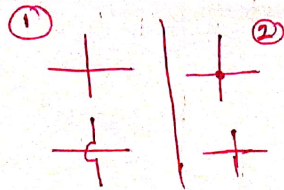
3. Truth Table:

- Boolean equations, truth tables and logic diagrams describe the same function!
- Truth tables are unique; expressions and logic diagrams are not. This gives flexibility in implementing functions.

X	Y	Z	F
0	0	0	0
0	0	1	1
0	1	0	0
0	1	1	0
1	0	0	1
1	0	1	1
1	1	0	1
1	1	1	1

3 متغیر  
نیز 8  
(2)<sup>3</sup> = 8

# Example

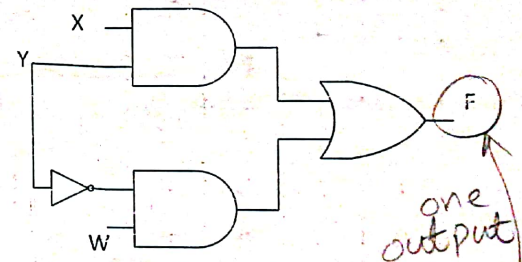


- Draw the logic diagram and the truth table of the following Boolean function:  $F(W, X, Y) = XY + W\bar{Y}$

Logic Diagram:

Truth Table:

W	X	Y	F
0	0	0	0
0	0	1	0
0	1	0	0
0	1	1	1
1	0	0	1
1	0	1	0
1	1	0	1
1	1	1	1



عدد المتغيرات = 3  
عدد الاحتمالات = 8

نوزج القيم بالاول 0 ← اصفار  
1 ← واحدات

وهكذا  
0 ← صفر  
1 ← واحد

- This example represents a **Single Output Function**

# Example

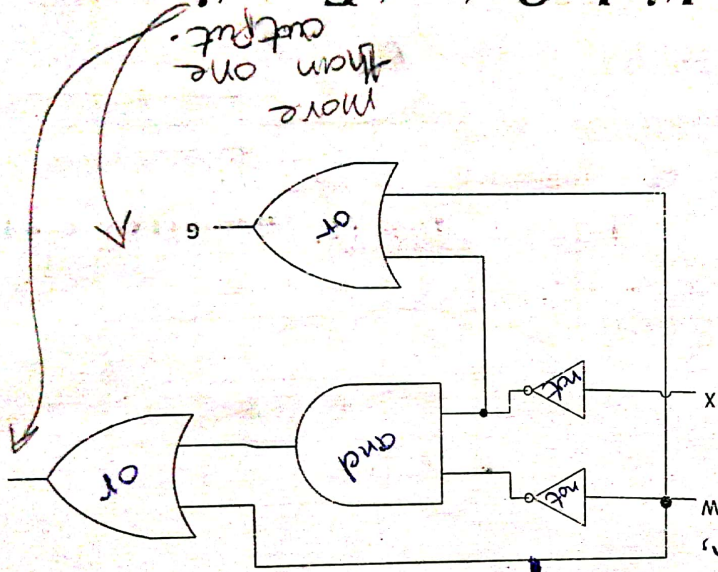
- Draw the logic diagram and the truth table of the following Boolean functions:  $F(W, X) = \overline{W}X + W\overline{X}$  and  $G(W, X) = W + \overline{X}$

- Logic Diagram:

- Truth Table:

W	X	F	G
0	0	0	1
0	1	1	1
1	0	0	0
1	1	1	1

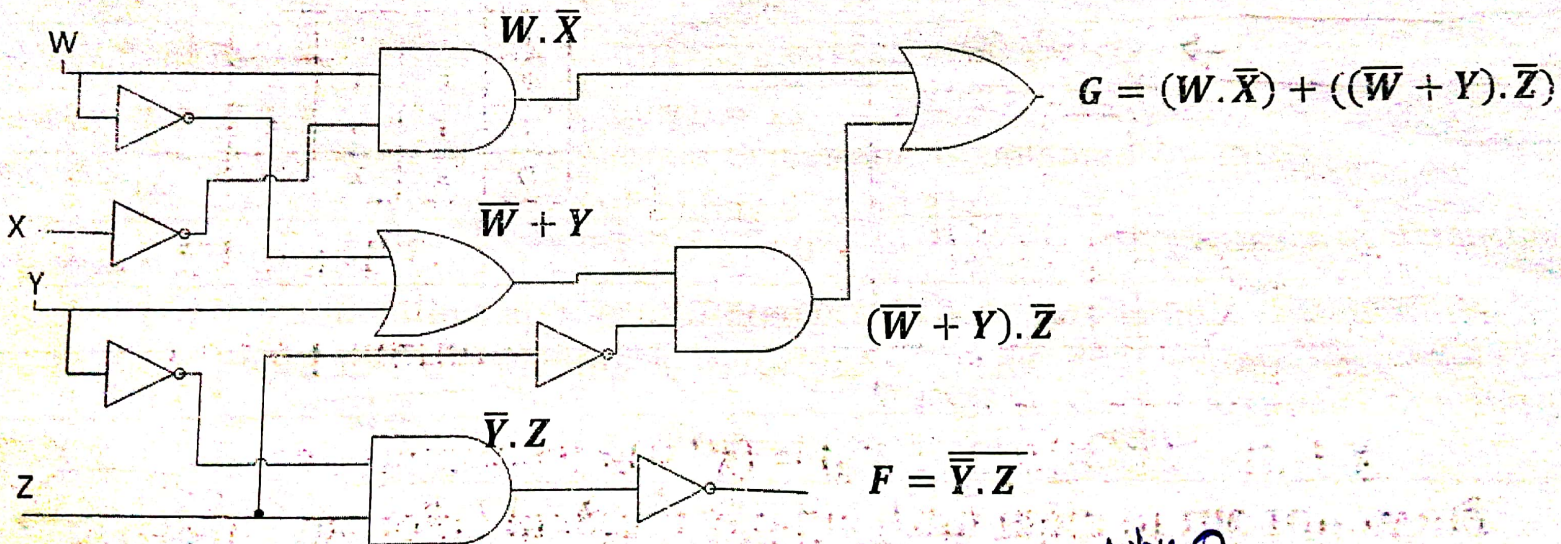
as a function of (w, x)



- This example represents a Multiple Output Function

# Example:

- Given the following logic diagram, write the corresponding Boolean equation:  
*المعادلة المنطقية التي تمثل الشكل التالي.*



- Logic circuits of this type are called combinational logic circuits (since the <sup>answer</sup> variables are combined by logical operations) *why?*

# \* Basic Identities of Boolean Algebra

\* The distributive law even though more than one variable

\* EX:  $AB \cdot (x+y+z) = AB \cdot x + AB \cdot y + AB \cdot z$

\*  $0 \cdot 0 = 0$   
 $\sum 1 = 0$   
 $(A+B+C) \cdot 0 = 0$

\* Complement = Not = bar = X. (Useful note)

\*  $x \cdot 1 = x$ ,  $0 \cdot 1 = 0$ ,  $1 \cdot 1 = 1$

1. $X + 0 = X$	2. $X \cdot 1 = X$	
3. $X + 1 = 1$	4. $X \cdot 0 = 0$	Existence of 0 and 1 <i>موجود الوجود</i>
5. $X + X = X$	6. $X \cdot X = X$	Idempotence
7. $X + \bar{X} = 1$	8. $X \cdot \bar{X} = 0$	Existence of complement
9. $\bar{\bar{X}} = X$		Involution
10. $X + Y = Y + X$	11. $XY = YX$	Commutative Laws
12. $(X + Y) + Z = X + (Y + Z)$	13. $(XY)Z = X(YZ)$	Associative Laws
14. $X(Y + Z) = XY + XZ$	15. $X + (YZ) = (X + Y)(X + Z)$	Distributive Laws
16. $\overline{X + Y} = \bar{X} \bar{Y}$	17. $\overline{X \cdot Y} = \bar{X} + \bar{Y}$	DeMorgan's Laws

هذا القانون ان يكون العلاقة بين المتغيرات انفسها

Even though more than two variables  
 \* EX:  $x+y+z = \bar{x} \cdot \bar{y} \cdot \bar{z}$

لا يوجد ما خارج القوس على القوس  
 و عمل العلاقة نفسها بين المتغيرات

نور و ال (not) و ال (and) و ال (or) العلاقة

قانون الترتيب للمتغيرات (تبادلي) العلاقة

## Some Properties of Identities & the Algebra

- If the meaning is unambiguous, we leave out the symbol ".".  

$$* AB = A \cdot B = A \cdot \text{AND} \cdot B$$

$$*) \overline{X+Y} = \overline{X} \cdot \overline{Y}$$

- The identities above are organized into pairs

نتيجة اقواس عشوائية  
 نهيمن ترتيب العليان

- The dual of an algebraic expression is obtained by interchanging (+) and (·) and interchanging 0's and 1's

عباراة متضادتين  
 تغيير / تبديل

- The identities appear in dual pairs. When there is only one identity on a line the identity is self-dual, i. e., the dual expression = the original expression.

الجواب للعبارتين  
 which means  
 المتضادتين نفسه

نعتبره متغير واحد

عبارة صحيحة:  $X + 0 = X$   
 عبارة صحيحة (عكس):  $X \cdot 1 = X$

\* Note:-  $X + Y \cdot Z = \overline{X} \cdot \overline{Y} \cdot \overline{Z}$

$\overline{X} \cdot (\overline{Y} + \overline{Z})$   
 نضع اقواس  
 لكي نحافظ على الأولوية

## Some Properties of Identities & the Algebra (Continued)

- Unless it happens to be self-dual, the dual of an expression does not equal the expression itself

Examples:

$$F = (A + \bar{C}) \cdot B + 0$$

$$\text{Dual } F = [(A \cdot \bar{C}) + B] \cdot 1 = A \cdot \bar{C} + B$$

$$G = XY + (\bar{W} + \bar{Z})$$

$$\text{Dual } G = (X + Y) \cdot \overline{W \cdot Z} = (X + Y) \cdot (\bar{W} + \bar{Z}) \rightarrow \text{Demorgan's law}$$

$$H = AB + AC + BC \quad \text{Distributive law.}$$

$$\begin{aligned} \text{Dual } H &= (A + B)(A + C)(B + C) = (A + BC)(B + C) \\ &= AB + AC + BC \end{aligned}$$

$B \cdot B \cdot C = B \cdot C$

Are any of these functions self-dual?  $\rightarrow$  H is equal to dual H

Yes, H is self-dual

$$* H: AB + AC + BC$$

$$\text{Dual } H: (A + B)(A + C)(B + C)$$

$$(A + BC)(B + C)$$

$$AB + AC + B \cdot B \cdot C + B \cdot C \cdot C$$



# Boolean Operator Precedence

- The order of evaluation in a Boolean expression is: أولويات العمليات:
  1. Parentheses الأقواس
  2. NOT not بوابة
  3. AND and بوابة
  4. OR or بوابة

- Consequence: Parentheses appear around (OR) expressions

\* لكي يتم تنفيذه قبل and أو not حسب الحاجة. \*

- Examples:

- $F = A(B + C)(C + \bar{D})$

- $F = \overline{AB} = \overline{A}B$

- $F = AB + C$

- $F = A(\bar{B} + C) = A \cdot \bar{B} + A \cdot C$

# Example

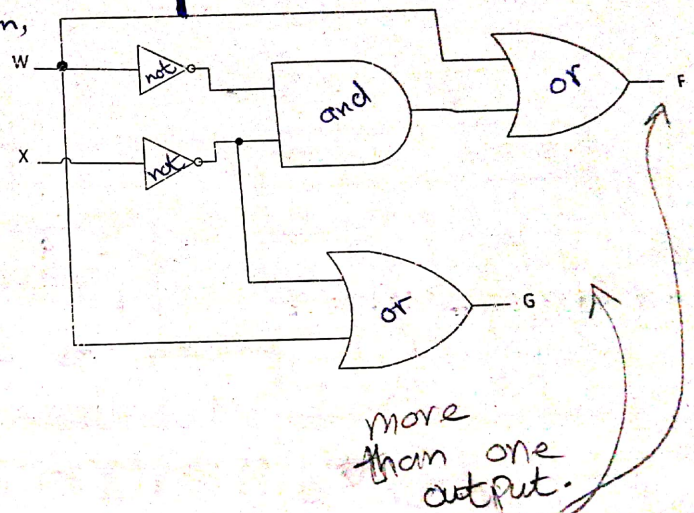
- Draw the logic diagram and the truth table of the following Boolean functions:  $F(W, X) = \bar{W}\bar{X} + W$ ,  $G(W, X) = W + \bar{X}$

- Logic Diagram:

- Truth Table:

W	X	F	G
0	0	1	1
0	1	0	0
1	0	1	1
1	1	1	1

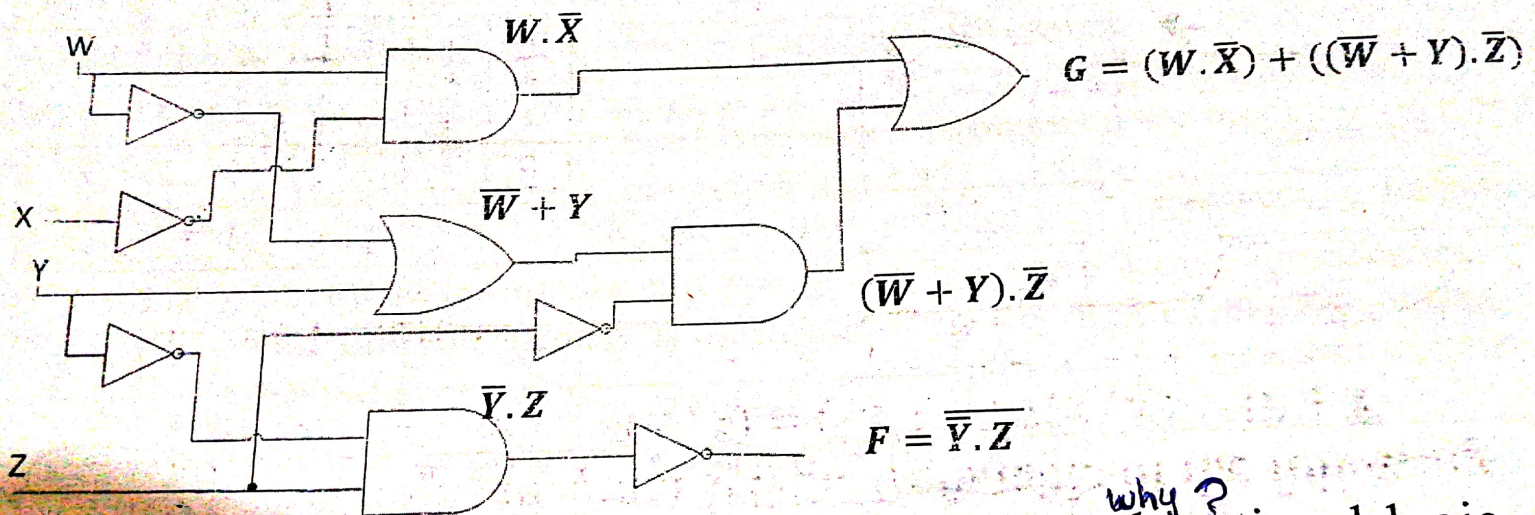
as a function, of (w, x)



- This example represents a Multiple Output Function

## Example:

- Given the following logic diagram, write the corresponding Boolean equation:



- Logic circuits of this type are called combinational logic circuits (since the <sup>answer</sup> variables are combined by logical operations) <sup>why?</sup>



# Basic Identities of Boolean Algebra

\* The distributive law even though more than one variable

\* Ex:  $AB \cdot (x+y+z) = AB \cdot x + AB \cdot y + AB \cdot z$

\*  $0 \cdot 0 = 0$   
 $0 \cdot 1 = 0$   
 $1 \cdot 0 = 0$   
 $(A+B+C) \cdot 0 = 0$

\* Complement = not = bar =  $\bar{x}$  (نفس المثل)

\*  $x \cdot 1 = x$ ,  $0 \cdot 1 = 0$ ,  $1 \cdot 1 = 1$

1. $X + 0 = X$	2. $X \cdot 1 = X$	Existence of 0 and 1 <i>amyan khalad mawqdeh</i>
3. $X + 1 = 1$	4. $X \cdot 0 = 0$ $\rightarrow$ $1 \cdot 0 = 0$ $0 \cdot 0 = 0$	
5. $X + X = X$	6. $X \cdot X = X$	Idempotence
7. $X + \bar{X} = 1$	8. $X \cdot \bar{X} = 0$	Existence of complement
9. $\bar{\bar{X}} = X$		Involution
10. $X + Y = Y + X$	11. $XY = YX$	Commutative Laws
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هذا القانون ان تكون العمليات بين المتغيرات نفسها

قانون الترتيب للمتغيرات (تبادلي العملية)

Even though more than two variables  
 \* Ex:  $x+y+z = \bar{x} \cdot \bar{y} \cdot \bar{z}$

توزيع ما خارج القوس على القوس  
 وعمل العملية نفسها بين المتغيرات

نورتي ال (not) ونعكس العملية (and  $\leftrightarrow$  or)

\* كل الاشارات التي تحت (-) تعكس

## Some Properties of Identities & the Algebra

- If the meaning is unambiguous, we leave out the symbol “.”  
 $\rightarrow * AB = A \cdot B = A \text{ AND } B$

$$*) \overline{x+y} = \bar{x} \cdot \bar{y}$$

- The identities above are organized into pairs

The dual of an algebraic expression is obtained by interchanging (+) and (·) and interchanging 0's and 1's

- The identities appear in dual pairs. When there is only one identity on a line the identity is self-dual, i. e., the dual expression = the original expression.

عباراته متضادة  
 الجوانب للجبارتين  
 المتضادتين نفسهما  
 which means  
 نعتبره متغير واحد.

عبارته صحيحة:  $x + 0 = x$   
 عبارة صحيحة (ضري):  $x \cdot 1 = x$

\* Note:-  $x + y \cdot z = \bar{x} \cdot \bar{y} \cdot \bar{z}$   
 $\bar{x} \cdot (\bar{y} + \bar{z})$   
 نضع أقواس  
 لكن نحافظ على الأولوية.

## Some Properties of Identities & the Algebra (Continued)

- Unless it happens to be self-dual, the dual of an expression does not equal the expression itself

### Examples:

*truth table ①*  
*تبسيط العبارة لتصبح متشابهة (dual H / H)*

- $F = (A + \bar{C}) \cdot B + 0$

- $Dual F = [(A \cdot \bar{C}) + B] \cdot 1 = A \cdot \bar{C} + B$

- $G = XY + (\bar{W} + \bar{Z})$

- $Dual G = (X + Y) \cdot \overline{W \cdot Z} = (X + Y) \cdot (\bar{W} + \bar{Z})$  → Demorgan's law: *التبسيط*

- $H = AB + AC + BC$  *Distributive law.*

- $Dual H = (A + B)(A + C)(B + C) = (A + BC)(B + C) = AB + AC + BC$   
*B.B.C = B.C*

- Are any of these functions self-dual?

- Yes, H is self-dual

*H is equal to dual H so it is a (self-dual).*

- $H: AB + AC + BC$

- $Dual H: (A + B)(A + C)(B + C)$

*Chapter 2 - Part 1 24*  
*نفس العبارة = نفس العبارة and note \*  $(A + B)(B + C) = AB + AC + BC + BC$   $BC = BCC$  or  $BC = BBC$   $BC = BCC$   $BC = BBC$   $BC = BCC$   $BC = BBC$*

# Boolean Operator Precedence

- The order of evaluation in a Boolean expression is: أولويات العمليات.
  1. Parentheses الأقواس
  2. NOT not بوابة
  3. AND and بوابة
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- Consequence: Parentheses appear around (OR) expressions

\* لكي يتم تنفيذه قبل and أو not حسب الحاجة. \*

- Examples:

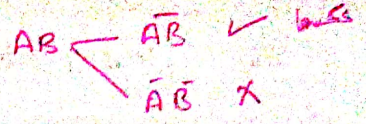
- $F = A(B + C)(C + \bar{D})$

- $F = \sim AB = \bar{A}B$

- $F = AB + C$

- $F = A(B + C) = A \cdot B + A \cdot C$

# Useful Boolean Theorems



تفسير  
 $X + \bar{X}(YZ) \Rightarrow (X + \bar{X})(X + YZ)$   
 نظرية التوزيع  
 وتوزيع في حد ذاته من التوزيع  
 نظرية التوزيع

$X + \bar{X}y = X + y \rightarrow$  simplification.  
 $x.y + \bar{x}.y = y \rightarrow$  minimization.  
 $x + x.y = x \rightarrow$  Absorption

المتكامل من المتكامل

Theorem	Dual	Name
1. $x.y + \bar{x}.y = y$	$(x + y)(\bar{x} + y) = y$	Minimization
2. $x + x.y = x$	$x.(x + y) = x$	Absorption
3. $x + \bar{x}.y = x + y$	$x.(\bar{x} + y) = x.y$	Simplification
4. $x.y + \bar{x}.z + y.z = x.y + \bar{x}.z$	$(x + y)(\bar{x} + z)(y + z) = (x + y)(\bar{x} + z)$	Consensus
$(x + y)(\bar{x} + z)(y + z) = (x + y)(\bar{x} + z)$		

\*  $F(A,B) = \sim A.B \rightarrow$  not A ①  
 A and B ②

بعض \*  $F(A,B) = \sim(A.B) \rightarrow$  A and B ①  
 not(A and B) ②



# Example 1: Boolean Algebraic Proof

▪  $A + A \cdot B = A$  (Absorption Theorem)

كأننا (A · 1) فنأخذ (A) عاملًا مباشرًا.

$  \begin{aligned}  &A + A \cdot B \\  &= A \cdot 1 + A \cdot B \\  &= A \cdot (1 + B) \quad \text{Distributive Law} \\  &= A \cdot 1 \quad \leftarrow 1 + X = 1 \\  &= A \quad \leftarrow X \cdot 1 = X  \end{aligned}  $	}	<p>dual:-</p> $  \begin{aligned}  &A \cdot (A + B) = A \\  &A \cdot A + A \cdot B \\  &\downarrow \\  &A + AB \\  &A(1 + B) \\  &A \cdot 1 = A  \end{aligned}  $
----------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------	---	------------------------------------------------------------------------------------------------------------------------------------------------------------------

\* even though:  $X + X \cdot (ABCDE) = X$

- Our primary reason for doing proofs is to learn:
  - Careful and efficient use of the identities and theorems of Boolean algebra
  - How to choose the appropriate identity or theorem to apply to make forward progress, irrespective of the application

## Example 2: Boolean Algebraic Proofs

even though more than one variables  $X \cdot yz + \bar{X} \cdot w + wyz = xyz + \bar{X} \cdot w$ .

$AB + \bar{A}C + BC = AB + \bar{A}C$  (Consensus Theorem) \*  $A + \bar{A} = 1$    
 كإن (BC) غير موجودة أي لا تؤثر. داعاً

$AB + \bar{A}C + BC$	كإن BC هي (BC.1)
$= AB + \bar{A}C + 1 \cdot BC$	$1 \cdot X = X$
$= AB + \bar{A}C + (A + \bar{A}) \cdot BC$	$X + \bar{X} = 1$
$= AB + \bar{A}C + ABC + \bar{A}BC$	Distributive Law: توزيع القوس
$= AB + ABC + \bar{A}C + \bar{A}BC$	Commutative Law ترتيب الحدود
$= AB \cdot 1 + AB \cdot C + \bar{A}C \cdot 1 + \bar{A}C \cdot B$	$X \cdot 1 = X$ and Commutative Law
$= AB(1 + C) + \bar{A}C(1 + B)$	Distributive Law: إخراج عامل مشترك
$= AB \cdot 1 + \bar{A}C \cdot 1$	$1 + X = 1$
$= AB + \bar{A}C$	$X \cdot 1 = X$

# Proof of Simplification

■  $A + \bar{A}.B = A + B$  (Simplification Theorem)

$$A + \bar{A}.B$$

توزيع على القوس.

$$= (A + \bar{A})(A + B)$$

*Distributive Law*

$$= 1.(A + B)$$

$$X + \bar{X} = 1$$

$$= A + B$$

$$X.1 = X$$

■  $A.(\bar{A} + B) = AB$  (Simplification Theorem)

$$A.(\bar{A} + B)$$

$$= (A.\bar{A}) + (A.B)$$

*Distributive Law*

$$= 0 + AB$$

$$X.\bar{X} = 0$$

$$= AB$$

$$X + 0 = X$$

# Proof of Minimization

البرهان

- $A.B + \bar{A}.B = B$  (Minimization Theorem)  
(نظرية التقليل (الاصغر))

$$A.B + \bar{A}.B$$

B : *ب*

$$= B(A + \bar{A})$$

Distributive Law

$$= B.1$$

$$X + \bar{X} = 1$$

$$= B$$

$$X.1 = X$$

- $(A + B)(\bar{A} + B) = B$  (Minimization Theorem)

$$(A + B)(\bar{A} + B)$$

B : *ب*

$$= B + (A.\bar{A})$$

Distributive Law

$$= B + 0$$

$$X.\bar{X} = 0$$

$$= B$$

$$X + 0 = X$$

$$* A + \bar{A} = 1$$

$$* A.\bar{A} = 0$$

# Proof of DeMorgan's Laws (1)

•  $\overline{X+Y} = \bar{X} \cdot \bar{Y}$  (DeMorgan's Law)

- We will show that,  $\bar{X} \cdot \bar{Y}$ , satisfies the definition of the complement of  $(X+Y)$ , defined as  $\overline{X+Y}$  by DeMorgan's Law.
- To show this, we need to show that  $\boxed{A+A'=1}$  and  $\boxed{A \cdot A'=0}$  with  $A = X+Y$  and  $A' = X' \cdot Y'$ . This proves that  $X' \cdot Y' = \overline{X+Y}$ .

• Part 1: Show  $X+Y+X' \cdot Y' = 1$

slide 32 \*  $\overline{X+Y} = \bar{X} \cdot \bar{Y}$

$A = X+Y$   
 $A' = \bar{X} \cdot \bar{Y}$   
 $A \cdot A' = 0$   
 $A + A' = 1$

slide 31.

\*  $X+Y = \bar{X} \cdot \bar{Y}$

$A + \bar{A} = (X+Y) + \bar{X} \cdot \bar{Y}$   
 $= (X+Y+\bar{X}) \cdot (X+Y+\bar{Y})$   
 $= (1+Y) \cdot (X+1)$   
 $= 1 \cdot 1 = 1$

$(X+Y) + X' \cdot Y'$	
$= (X+Y+\bar{X})(X+Y+\bar{Y})$	Distributive Law
$= (1+Y)(X+1)$	$X+\bar{X} = 1$
$= 1 \cdot 1$	$X+1 = 1$
$= 1$	$X \cdot 1 = X$

\*  $\bar{X} \cdot \bar{Y} = \overline{X+Y}$   
 $A \cdot \bar{A} = (X+Y) \cdot \bar{X} \cdot \bar{Y}$   
 $= \bar{X} \cdot \bar{Y} \cdot (X+Y)$   
 $= \bar{X} \cdot \bar{Y} \cdot X + \bar{X} \cdot \bar{Y} \cdot Y$   
 $= \bar{X} \cdot \bar{Y} \cdot X + \bar{X} \cdot \bar{Y} \cdot Y$   
 $= 0 + 0 = 0$

# Proof of DeMorgan's Laws (2)

- Part 2: Show  $(X + Y) \cdot X' \cdot Y' = 0$

$(X + Y) \cdot X' \cdot Y'$	
$= (X \cdot X' \cdot Y') + (Y \cdot X' \cdot Y')$	<i>Distributive Law</i>
$= (0 \cdot Y') + (X' \cdot 0)$	$X \cdot \bar{X} = 0$
$= 0 + 0$	$X \cdot 0 = 0$
$= 0$	$\bar{X} + 0 = X$

$x + y + x \cdot y'$   
 $x + y + y \cdot \bar{x}$   
 $x + y + \bar{x}$  →  
 $1 + y = 1$   
 $y + x + \bar{x} \cdot y$   
 $y + x + y$   
 $1 + x = 1$

- Based on the above two parts,  $X' \cdot Y' = \overline{X + Y}$
- \* The second DeMorgan's law is proved by duality
- Note that DeMorgan's law, given as an identity is not an axiom in the sense that it can be proved using the other identities.

### Example 3: Boolean Algebraic Proofs

$$\equiv \overline{(X + Y)}Z + X\bar{Y} = \bar{Y}(X + Z)$$

$$\overline{(X + Y)}Z + X\bar{Y}$$

$$= X'Y'Z + X.Y'$$

DeMorgan's law (not) توزيع

$$= Y'(X'Z + X)$$

Distributive law. عامل مشترك

$$= Y'(X + X'Z)$$

Commutative law ترتيب الحدود

$$= Y'(X + Z)$$

Simplification Theorem  $A + \bar{A}.B = A + B.$

# Boolean Function Evaluation

- $F_1 = xyz$
- $F_2 = x + \bar{y}z$
- $F_3 = \bar{x}\bar{y}z + x\bar{y}z + xy$
- $F_4 = x\bar{y} + \bar{x}z$

x	y	z	$F_1$	$F_2$	$F_3$	$F_4$
0	0	0	0	0	1	0
0	0	1	0	1	0	1
0	1	0	0	0	0	0
0	1	1	0	0	1	1
1	0	0	0	1	1	1
1	0	1	0	1	1	1
1	1	0	1	1	0	0
1	1	1	1	1	0	0

\* نأخذ على هذه الأسئلة 4 نغرد حتى يكون قيمة الناتج

النطائي (1) أو (0) حسب العبارة وحالة المتغيرات أفضل

من حل جميع العباران لجميع القيم ونأخذ للسرعة و لتسهيل

الكل



# Expression Simplification

- An application of Boolean algebra
- Simplify to contain the smallest number of literals (complemented and uncomplemented variables). *تحتوي أصغر عدد المتغيرات من التعبير لأقل عدد ممكن. have (not) gate.*
- Example: Simplify the following Boolean expression
  - $AB + A'CD + A'BD + A'CD' + ABCD$

$AB + A'CD + A'BD + A'CD' + ABCD$	
$= AB + ABCD + A'CD + A'CD' + A'BD$	<i>Commutative law</i>
$= AB(1 + CD) + A'C(D + D') + A'BD$	<i>Distributive law</i>
$= AB \cdot 1 + A'C \cdot 1 + A'BD$	$1 + X = 1$ and $X + X' = 1$
$= AB + A'C + A'BD$	$X \cdot 1 = X$
$= AB + A'BD + A'C$	<i>Commutative law</i>
$= B(A + A'D) + A'C$	<i>Distributive law</i>
$= B(A + D) + A'C \rightarrow 5 \text{ Literals}$	<i>Simplification Theorem</i>

# Complementing Functions (not) 2x3 نفي المتغير

- Use DeMorgan's Theorem to complement a function:
  1. Interchange AND and OR operators
  2. Complement each constant value and literal

■ Example: Complement  $F = x'yz' + xy'z'$

$$F' = (x + y' + z)(x' + y + z)$$

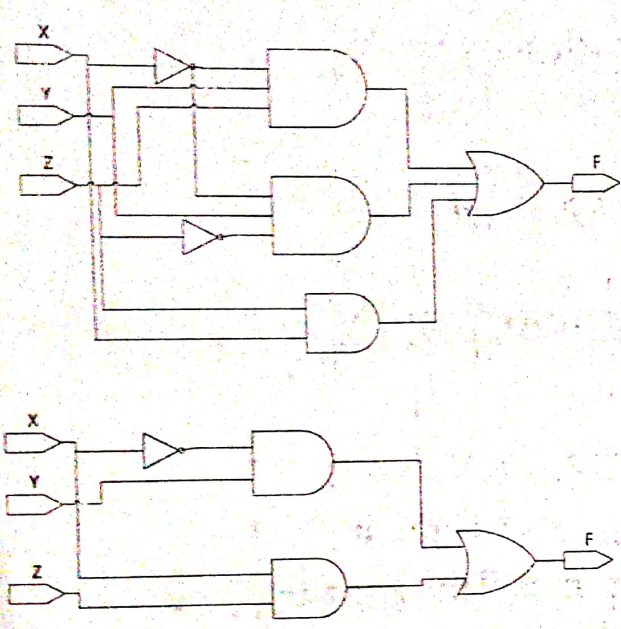
■ Example: Complement  $G = (a' + bc)d' + e$

$$G' = (a(b' + c') + d).e'$$

# Example

- Simplify the following:

- $$F = X'YZ + X'YZ' + XZ$$



$$X'YZ + X'YZ' + XZ$$

$$= X'Y(Z + Z') + XZ \quad \text{Distributive law}$$

$$= X'Y \cdot 1 + XZ \quad X + X' = 1$$

$$= X'Y + XZ \quad X \cdot 1 = X$$

x	y	z	$X'YZ + X'YZ' + XZ$	$X'Y + XZ$
0	0	0	0	0
0	0	1	0	0
0	1	0	1	1
0	1	1	1	1
1	0	0	0	0
1	0	1	1	1
1	1	0	0	0
1	1	1	1	1
			3 terms and 8 literals	2 terms and 4 literals

# Example

▪ Show that  $F = x'y' + xy' + x'y + xy = 1$

- Solution1: Truth Table

x	y	F
0	0	1
0	1	1
1	0	1
1	1	1

\* Note: عند وجود جميع احتمالات التعبير ضلال الاقران هذا يعني  
بأن الجواب دائماً (1).  
الاحتمالات جميعها هي:  
 $\left. \begin{array}{l} \overline{x}y \\ x\overline{y} \\ xy \\ x\overline{y} \end{array} \right\}$

- Solution2: Boolean Algebra

$$x'y' + xy' + x'y + xy$$

$$= y'(x' + x) + y(x' + x) \quad \text{Distributive law}$$

$$= y' \cdot 1 + y \cdot 1 \quad X + X' = 1$$

$$= y' + y \quad X \cdot 1 = X$$

$$= 1 \quad X + X' = 1$$

# Examples

- Show that  $ABC + A'C' + AC' = AB + C'$  using Boolean algebra.

$ABC + A'C' + AC'$	
$= ABC + C'(A' + A)$	Distributive law
$= ABC + C'.1$	$X + X' = 1$
$= ABC + C'$	$X.1 = X$
$= (AB + C')(C + C')$	Distributive law
$= (AB + C').1$	$X + X' = 1$
$= AB + C'$	$X.1 = X$

$\bar{C} = X$   
 $AB = Y$   
 $C = Z$

\* Distributive law  
 $X + Y.Z = (X+Y).Z$

- Find the dual and the complement of  $f = wx + y'z.0 + w'z$

•  $Dual(f) = (w + x)(y' + z + 1)(w' + z)$

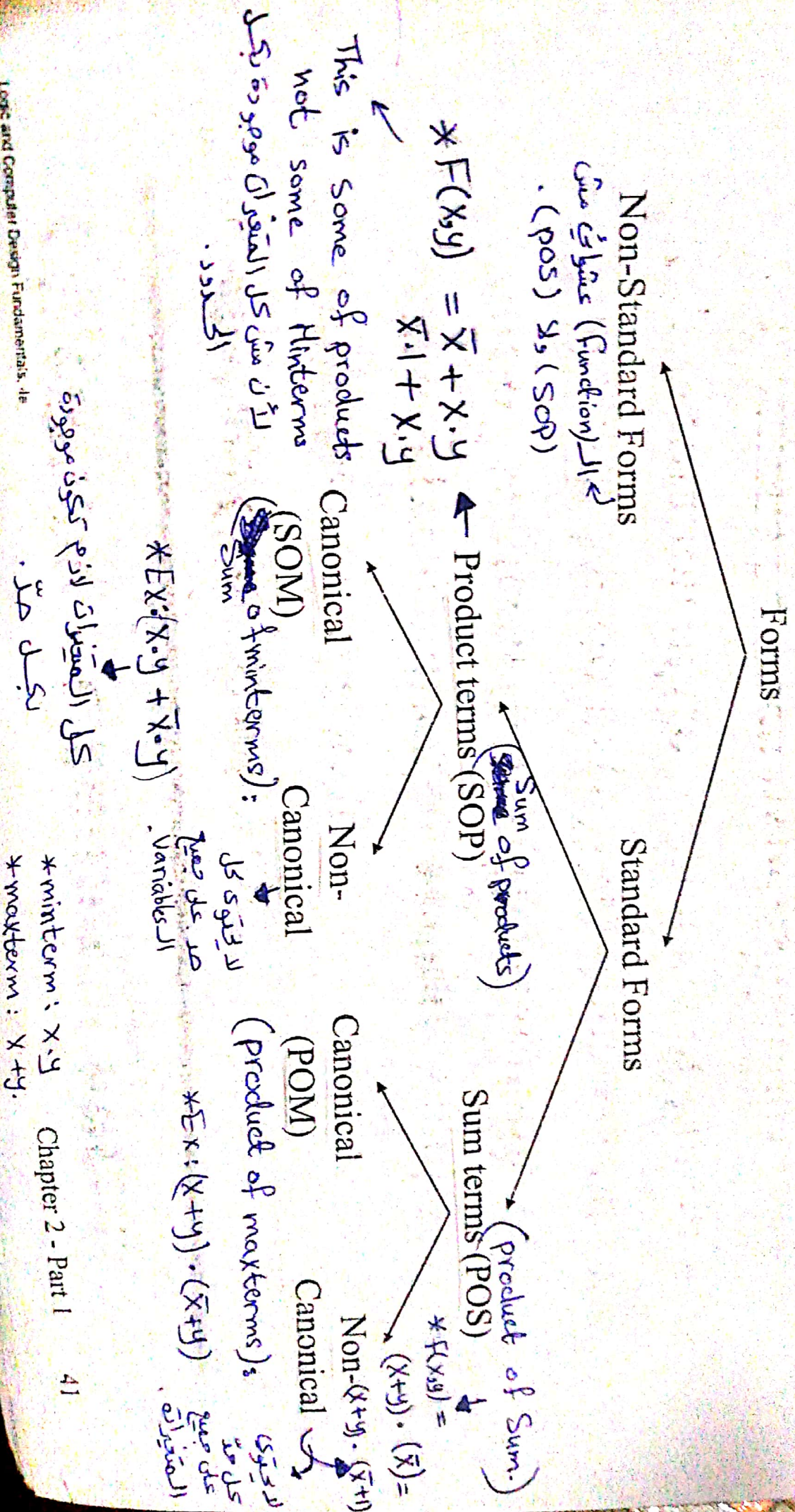
Note \* هذا السؤال ليس تافه

Complement

•  $f' = (w' + x')(y + z' + 1)(w + z')$

هذا dual و complement الـ function

# Boolean Representation Forms



# Canonical Forms

---

- It is useful to specify Boolean functions in a form that:
  - Allows comparison for equality
  - Has a correspondence to the truth tables
  - Facilitates simplification
- Canonical Forms in common usage:
  - Sum of Minterms (SOM)  $xy + \bar{x}\bar{y} + \bar{x}y \dots$
  - Product of Maxterms (POM)  $(x+y) \cdot (\bar{x} + \bar{y}) \dots$

# Minterms

Minterms are **AND** terms with **every variable** present in either **true** or **complemented** form

(X) إيجاب (X)

( $\bar{X}$ ) منقولة (X)

كتوبي على كل المتغيرات مرتبة بالترتيب الأبجدي

Given that each binary variable may appear normal (e.g.,  $x$ ) or complemented (e.g.,  $\bar{x}$ ), there are  $2^n$  minterms for  $n$  variables

Example: **Two** variables ( $X$  and  $Y$ ) produce  $2^2 = 4$  combinations:

المتغيرات **تسمى** and

عدد الاحتمالات  $2^n = 4$  (بعضها غير)

- ①  $XY$  (both normal)
- ②  $X\bar{Y}$  (X normal, Y complemented)
- ③  $\bar{X}Y$  (X complemented, Y normal)
- ④  $\bar{X}\bar{Y}$  (both complemented)

	x	y	z	minterm
0	0	0	0	$\bar{x}\bar{y}\bar{z}$
1	0	0	1	$\bar{x}\bar{y}z$
2	0	1	0	$\bar{x}y\bar{z}$
3	0	1	1	$\bar{x}yz$

Thus there are **four minterms** of two variables

$$f(x, y) = X \cdot y + X \cdot \bar{y} + \bar{X} \cdot y + \bar{X} \cdot \bar{y} \text{ (some of minterms)}$$



# Maxterms

▪ Maxterms are OR terms with every variable in true or complemented form

(X) (X-bar)  
(Y) (Y-bar)

▪ Given that each binary variable may appear normal (e.g.,  $x$ ) or complemented (e.g.,  $\bar{x}$ ), there are  $2^n$  maxterms for  $n$  variables

▪ Example: Two variables ( $X$  and  $Y$ ) produce  $2^2 = 4$  combinations:

①  $X + Y$  (both normal)

②  $X + \bar{Y}$  (X normal, Y complemented)

③  $\bar{X} + Y$  (X complemented, Y normal)

④  $\bar{X} + \bar{Y}$  (both complemented)

# Maxterms and Minterms

مخرج ال minterm في اوقات او رقم حقيقي (-)  
 مداخل مخرجه (0) و (1)  
 مخرج ال minterm في اوقات حقيقي في  
 لمدخل ال input -1 و 0  
 مخرج ال minterm (0)

Examples: Three variable (X, Y, Z) minterms and maxterms

Index	Minterm (m)	Maxterm (M)
0	$\bar{X}\bar{Y}\bar{Z}$	$X + Y + Z$
1	$\bar{X}\bar{Y}Z$	$X + Y + \bar{Z}$
2	$\bar{X}Y\bar{Z}$	$X + \bar{Y} + Z$
3	$\bar{X}YZ$	$X + \bar{Y} + \bar{Z}$
4	$X\bar{Y}\bar{Z}$	$\bar{X} + Y + Z$
5	$X\bar{Y}Z$	$\bar{X} + Y + \bar{Z}$
6	$XY\bar{Z}$	$\bar{X} + \bar{Y} + Z$
7	$XYZ$	$\bar{X} + \bar{Y} + \bar{Z}$

ال max ال minterm  
 ال minterm (0) ال max ال minterm  
 ال minterm (1) ال max ال minterm  
 \* المinterm (0) ال max ال minterm  
 \* المinterm (1) ال max ال minterm  
 minterm 0 :  $\bar{X}\bar{Y}\bar{Z}$  (m0)  
 maxterm 0 :  $X+Y+Z$  (M0)  
 minterm 1 :  $\bar{X}\bar{Y}Z$  (m1)  
 وهكذا

m0 / M0  
 m1 / M1  
 \* نلاحظ : في ال minterm  
 نبدأ بنفي جميع المتغيرات  
 فن كل ال variables يكونوا  
 complement . بينما في ال  
 maxterm نبدأ من ال true  
 ال true في حالة ال true  
 وذلك ليس في ال minterm  
 ال complemented يكون في ال minterm (0)  
 وذلك عكس ال maxterms  
 يكون ال true variable

$\bar{F}(x,y) = (x+y) + (\bar{x}y) + (x\bar{y}) + (\bar{x}\bar{y}) =$

The index above is important for describing which variables in the terms are true and which are complemented

Sum of minterms:  $F(x,y) = (xy) + (\bar{x}y) + (x\bar{y}) + (\bar{x}\bar{y})$   
 Product of maxterms:  $F(x,y) = (\bar{x}+y) \cdot (\bar{x}+y) \cdot (x+\bar{y}) \cdot (x+\bar{y})$

$\bar{F}(minterms) = F(maxterms)$   
 ال true ال true variable

# Standard Order

- Minterms and maxterms are designated with a subscript
- The subscript is a number, corresponding to a binary pattern
- The bits in the pattern represent the complemented or normal state of each variable listed in a standard order  $m_0 / m_1 / m_2 \dots$   
 $M_0 / M_1 / M_2 \dots$
- All variables will be present in a minterm or maxterm and will be listed in the same order (usually alphabetically) ترتيباً أبجدياً
- **Example:** For variables a, b, c:
  - **Maxterms:**  $(a + b + \bar{c})$ ,  $(a + b + c)$
  - **Terms:**  $(b + a + c)$ ,  $a\bar{c}b$ , and  $(c + b + a)$  are NOT in standard order.   
 موصوفين أبجدياً (عادي بوز) و لكن يكون (term) ليس (maxterm)
  - **Minterms:**  $\bar{a}bc$ ,  $abc$ ,  $\bar{a}bc$  → must contain all variables & must be in the standard order.
  - **Terms:**  $(a + c)$ ,  $\bar{b}c$ , and  $(\bar{a} + b)$  do not contain all variables

\* المinterm : هو تعبير يحتوي على جميع ال Variables (متغيرات) من السؤال ويكون فيه جميع المتغيرات بالترتيب الأبجدي. Chapter 2 - Part 1 46

# Purpose of the Index

- The index for the minterm or maxterm, expressed as a binary number, is used to determine whether the variable is shown in the true form or complemented form

\* EX :-  
(minterms):  
 $m_0 = \bar{x}\bar{y}\bar{z} = 000$   
 $m_1 = \bar{x}\bar{y}z = 001$

- For Minterms:

\* EX :-  
(maxterms):  
 $M_0 = xyz = 000$   
 $M_1 = x\bar{y}\bar{z} = 001$

use  
ver  
[...]

- “0” means the variable is “Complemented”
- “1” means the variable is “Not Complemented”

- For Maxterms:

- “0” means the variable is “Not Complemented”
- “1” means the variable is “Complemented”

# Index Example: Three Variables

\* for minterm: 0 → complemented, 1 → true.  
 \* for maxterm: 0 → true, 1 → complemented.

Index (Decimal)	Index (Binary) n = 3 Variables	Minterm (m)	Maxterm (M)
0	000	$m_0 = \bar{X}\bar{Y}\bar{Z}$	$M_0 = X + Y + Z$
1	001	$m_1 = \bar{X}\bar{Y}Z$	$M_1 = X + Y + \bar{Z}$
2	010	$m_2 = \bar{X}Y\bar{Z}$	$M_2 = X + \bar{Y} + Z$
3	011	$m_3 = \bar{X}YZ$	$M_3 = X + \bar{Y} + \bar{Z}$
4	100	$m_4 = XY\bar{Z}$	$M_4 = \bar{X} + Y + Z$
5	101	$m_5 = XYZ$	$M_5 = \bar{X} + Y + \bar{Z}$
6	110	$m_6 = XYZ$	$M_6 = \bar{X} + \bar{Y} + Z$
7	111	$m_7 = XYZ$	$M_7 = \bar{X} + \bar{Y} + \bar{Z}$

# Index Example: Four Variables

\* الفرق فقط لأنه تم زيادة عدد المتغيرات  $\leftarrow 16 = 2^4$ .

i (Decimal)	i (Binary) n = 4 Variables	المصطلح المكون (-) $m_i$	المعادلة مكون (-) $M_i$
0	0000	$\bar{a}\bar{b}\bar{c}\bar{d}$	$a + b + c + d$
1	0001	$\bar{a}\bar{b}\bar{c}d$	$a + b + c + \bar{d}$
3	0011	$\bar{a}\bar{b}cd$	$a + b + \bar{c} + \bar{d}$
5	0101	$\bar{a}b\bar{c}\bar{d}$	$a + \bar{b} + c + \bar{d}$
7	0111	$\bar{a}bcd$	$a + \bar{b} + \bar{c} + \bar{d}$
10	1010	$a\bar{b}\bar{c}\bar{d}$	$\bar{a} + b + \bar{c} + d$
13	1101	$ab\bar{c}\bar{d}$	$\bar{a} + \bar{b} + c + \bar{d}$
15	1111	$abcd$	$\bar{a} + \bar{b} + \bar{c} + \bar{d}$

\*  $(m_2 \text{ و } m_4 \text{ و } m_6 \text{ و } \dots)$  هو الجواب (function) هو.

# Minterm and Maxterm Relationship

## Review: DeMorgan's Theorem

- $\overline{x \cdot y} = \bar{x} + \bar{y}$  and  $\overline{x + y} = \bar{x} \cdot \bar{y}$

## Two-variable example:

$$\overline{m_2} = \overline{x \cdot \bar{y}} = \overline{x + y} = M_2$$

- $M_2 = \bar{x} + y$  and  $m_2 = x \cdot \bar{y} \leftrightarrow (10)_2$

- Using DeMorgan's Theorem  $\rightarrow \overline{\bar{x} + y} = \overline{\bar{x}} \cdot \overline{y} = x \cdot \bar{y}$

- Using DeMorgan's Theorem  $\rightarrow \overline{x \cdot \bar{y}} = \overline{x} + \overline{\bar{y}} = \bar{x} + y$

- Thus,  $M_2$  is the complement of  $m_2$  and vice-versa also

\*  $m_2$  is the complement of  $M_2$ .

Since DeMorgan's Theorem holds for  $n$  variables, the above holds for terms of  $n$  variables:

$$M_i = \overline{m_i} \text{ and } m_i = \overline{M_i} \quad * \text{ عكس بعضهما بعضاً}$$

Thus,  $M_i$  is the complement of  $m_i$  and vice-versa

هنا رقم الحد  
or minterm  
or maxterm.

# Function Tables for Both

## Minterms of 2 variables:

\* كل صيغة يكون الرقم (1) موجود في حجرة واحدة والباقى أصفار (0).

\* نلاحظ بأن الرقم (1) قليل كثيراً لذلك minterm.

xy	$m_0$	$m_1$	$m_2$	$m_3$
00	1	0	0	0
01	0	1	0	0
10	0	0	1	0
11	0	0	0	1

## Maxterms of 2 variables:

\* كل عود أو سطر في جدول

ال (minterm) يكون نفي لنفس العود

أو السطر في ال (maxterm). لأن  $(M_i = \bar{m}_i)$  وكذلك  $(m_i = \bar{M}_i)$

\* كما قلنا سابقاً بأن  $M_i = \bar{m}_i$  حيث أنهم عكس بعض.

\* نلاحظ بأن الرقم (1) سائد كثيراً لذلك maxterm.

\* الرقم 1 في موجود في كل حجرة

xy	$M_0$	$M_1$	$M_2$	$M_3$
00	0	1	1	1
01	1	0	1	1
10	1	1	0	1
11	1	1	1	0

عكس بعض  
كذلك

- Each column in the maxterm function table is the complement of the column in the minterm function table since  $M_i$  is the complement of  $m_i$ .

$M_i = \bar{m}_i$        $M_i = m_i$



# Observations

In the function tables:

- Each *minterm* has one and only one 1 present in the  $2^n$  terms (a minimum of 1s). All other entries are 0.   
 \* كل عبارة *minterm* لها 1 واحدة فقط في  $2^n$  الحدود (الحد الأدنى من الـ 1s). جميع الإدخالات الأخرى هي 0.   
 (ملاحظة: *minterm* = الحد الأدنى من الـ 1s)
- Each *maxterm* has one and only one 0 present in the  $2^n$  terms. All other entries are 1 (a maximum of 1s).   
 \* كل عبارة *maxterm* لها 0 واحدة فقط في  $2^n$  الحدود. جميع الإدخالات الأخرى هي 1 (الحد الأقصى من الـ 1s).   
 (ملاحظة: *maxterm* = الحد الأقصى من الـ 1s)

We can implement any function by

- "ORing" the *minterms* corresponding to "1" entries in the function table. These are called the *minterms* of the function.   
 (أي: جمع الحدود الدنيا التي تساوي 1 في الجدول)
- "ANDing" the *maxterms* corresponding to "0" entries in the function table. These are called the *maxterms* of the function.   
 (أي: ضرب الحدود العظمى التي تساوي 0 في الجدول)

This gives us two canonical forms for stating any Boolean function:

- Sum of Minterms (SOM)*  $\rightarrow f(x,y) = X\bar{y} + xy$ . (ORing)
- Product of Maxterms (POM)*  $\rightarrow f(x,y) = (x+y) \cdot (x+y)$ . (ANDing)

# Minterm Function Example

Example: Find  $F_1 = m_1 + m_4 + m_7$

$F_1(x,y,z) = x'y'z + xy'z' + xyz$  → (Some of minterm) (SOM)  
 مبرمات (1) و (4) و (7) فقط  
 مبرمات (0) و (2) و (3) و (5) و (6) فقط

$2^3 = 8$   
 squares in the K-map

xyz	Index	$m_1 + m_4 + m_7 = F_1$
000	0	$0 + 0 + 0 = 0$
001	1	$1 + 0 + 0 = 1$
010	2	$0 + 0 + 0 = 0$
011	3	$0 + 0 + 0 = 0$
100	4	$0 + 1 + 0 = 1$
101	5	$0 + 0 + 0 = 0$
110	6	$0 + 0 + 0 = 0$
111	7	$0 + 0 + 1 = 1$

# Minterm Function Example

$$F(A, B, C, D, E) = m_2 + m_9 + m_{17} + m_{23}$$

أرقام الـ (maxterm) الذين قيمتهم = (صفر) وهم (٢، ٩، ١٧، ٢٣).

$$F(A, B, C, D, E) = A'B'C'DE' + A'BC'D'E + AB'C'D'E + AB'CDE$$

٨ ٤ ٢ ١  
عليا  
من المصغر

\* ملاحظة - أرقام الـ (maxterms) نختارها عن الأماكن التي يكون فيها الـ (function) = صفر، (0).

وكذلك أرقام الـ (minterms) نختارها عن الأماكن التي يكون فيها الـ (function) = واحد، (1).

# Maxterm Function Example

Example: Implement F1 in maxterms:

$F_1 = M_0 \cdot M_2 \cdot M_3 \cdot M_5 \cdot M_6$  (0, 2, 3, 5, 6)   
 هي أوقات الأصفار:   
 بال truth table لانه maxterms

$F_1 = (x + y + z) \cdot (x + y' + z) \cdot (x + y' + z') \cdot (x' + y + z') \cdot (x' + y' + z)$

xyz	Index	$M_0 \cdot M_2 \cdot M_3 \cdot M_5 \cdot M_6 = F_1$
000	<u>0</u>	$0 \cdot 1 \cdot 1 \cdot 1 \cdot 1 = 0$
001	1	$1 \cdot 1 \cdot 1 \cdot 1 \cdot 1 = 1$
010	<u>2</u>	$1 \cdot 0 \cdot 1 \cdot 1 \cdot 1 = 0$
011	<u>3</u>	$1 \cdot 1 \cdot 0 \cdot 1 \cdot 1 = 0$
100	4	$1 \cdot 1 \cdot 1 \cdot 1 \cdot 1 = 1$
101	<u>5</u>	$1 \cdot 1 \cdot 1 \cdot 0 \cdot 1 = 0$
110	<u>6</u>	$1 \cdot 1 \cdot 1 \cdot 1 \cdot 0 = 0$
111	7	$1 \cdot 1 \cdot 1 \cdot 1 \cdot 1 = 1$

# Maxterm Function Example

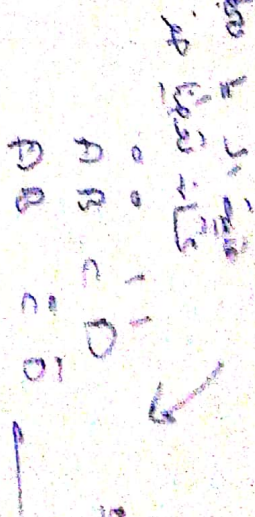
---

- $$F(A, B, C, D) = M_3 \cdot M_8 \cdot M_{11} \cdot M_{14}$$

*(10) (11) (14) (15)*  
*AP BP CP DP*
- $$F(A, B, C, D) \quad \text{[Product of Sums]}$$

$$= (A + B + C' + D') \cdot (A' + B + C + D)$$

$$(A' + B + C' + D') \cdot (A' + B' + C' + D)$$



# Canonical Sum of Minterms

From (SOP)  $\rightarrow$  (SOM)

Any Boolean function can be expressed as a Sum of Minterms (SOM):

- For the function table, the minterms used are the terms corresponding to the 1's
- For expressions, expand all terms first to explicitly list all minterms. Do this by "ANDing" any term missing a variable  $v$  with a term  $(v + \bar{v})$

Example: Implement  $f(x,y) = x + \bar{x}\bar{y}$  as a SOM? not (SOM) but

1. Expand terms  $\rightarrow f(x,y) = x(y + \bar{y}) + \bar{x}\bar{y}$  (SOP):  
(Sum of products)

2. Distributive law  $\rightarrow f = xy + x\bar{y} + \bar{x}\bar{y}$

3. Express as SOM  $\rightarrow f = m_3 + m_2 + m_0 = m_0 + m_2 + m_3$

! hsa sud dache per

# Another SOM Example

- Example:  $F = A + \bar{B}C$
- There are three variables: A, B, and C which we take to be the standard order

- Expanding the terms with missing variables:
  - $F = A(B + \bar{B})(C + \bar{C}) + (A + \bar{A})\bar{B}C$

ستأخذ كل حد ما ينقصه متغيرات

ونضعهم (A+A) أو (B+B) أو (C+C)

لكي لا نغير بقيمة التعبير الرباعي

أ يفتقر إلى A (B+B) وتأخذ (C+C) يفتقر إلى C

- Distributive law:

$$F = ABC + \bar{A}\bar{B}C + AB\bar{C} + \bar{A}\bar{B}\bar{C} + \bar{A}BC + \bar{A}\bar{B}C$$

\* لا نكرر الحدود المتكررة. ← نكتبهم مرة واحدة.

- Collect terms (removing all but one of duplicate terms):

$$F = ABC + \bar{A}\bar{B}C + AB\bar{C} + \bar{A}\bar{B}\bar{C} + \bar{A}BC$$

- Express as SOM:

$$F = m_7 + m_6 + m_5 + m_4 + m_1$$

$$F = m_1 + m_4 + m_5 + m_6 + m_7$$

# Shorthand SOM Form

- From the previous example, we started with:
  - $F = A + \bar{B}C$
- We ended up with:
  - $F = m_1 + m_4 + m_5 + m_6 + m_7$
- This can be denoted in the *formal shorthand*:
  - $F(A, B, C) = \sum_m(1, 4, 5, 6, 7)$   
(sum)
- Note that we explicitly show the standard variables in order and drop the “m” designators.



# Canonical Product of Maxterms

From (POS)  $\rightarrow$  (POM)

- Any Boolean Function can be expressed as a

## Product of Maxterms (POM):

\*For converting:-

from (SOP) to (SOM)

$$X \cdot (V + \bar{V}) = X$$

$$X + (V \cdot \bar{V}) = X$$

from (POS) to (POM)

For the function table, the maxterms used are the terms corresponding to the 0's (في جدول الحقيقة) (maxterm) \*  
 For an expression, expand all terms first to explicitly list all maxterms. Do this by first applying the second distributive law, "ORing" terms missing variable  $v$  with  $[v \cdot \bar{v}]$  and then applying the distributive law again

Example: Convert  $f(x, y, z) = x + \bar{x}\bar{y}$  to POM?

- Distributive law  $\rightarrow f = (x + \bar{x}) \cdot (x + \bar{y}) = x + \bar{y}$
- ORing with missing variable (z)  $\rightarrow f = x + \bar{y} + (z \cdot \bar{z})$
- Distributive law  $\rightarrow f = (x + \bar{y} + z) \cdot (x + \bar{y} + \bar{z})$
- Express as POS  $\rightarrow f = M_2 \cdot M_3$

# Another POM Example

Wimp

Convert  $f(A, B, C) = AC' + BC + A'B'$  to POM? (SOP)

Use  $x + yz = (x + y) \cdot (x + z)$ , assuming distributive law:

$x = AC' + BC$  and  $y = A'$  and  $z = B'$

$$f(A, B, C) = (AC' + BC + A') \cdot (AC' + BC + B')$$

Use Simplification theorem to get:  $x + \bar{x}y = x + y$

$$f(A, B, C) = (BC + A' + C') \cdot (AC' + B' + C)$$

$$x = \bar{c} \quad \bar{c} + \bar{c}B = \bar{c} + B$$

$$x = \bar{A}$$

$$\bar{A} + \bar{A}y = \bar{A} + y$$

$$\bar{A} + \bar{A}y = \bar{A} + y$$

Use Simplification theorem again to get:

$$f(A, B, C) = (A' + B + C') \cdot (A + B' + C) = M_5 \cdot M_2$$

$$f(A, B, C) = M_2 \cdot M_5 = \prod_M(2, 5) \rightarrow \text{Shorthand POM}$$

form

(multiply) (vars)

geräthel örgedel

# Function Complements

---

- The complement of a function expressed as a sum of minterms is constructed by selecting the minterms missing in the sum-of-minterms canonical forms.
- Alternatively, the complement of a function expressed by a sum of minterms form is simply the Product of Maxterms with the same indices.

- Example: Given  $F(x, y, z) = \sum_m(1, 3, 5, 7)$ , find complement  $\bar{F}$  as SOM and POM?
  - $\bar{F}(x, y, z) = \sum_m(0, 2, 4, 6)$  (SOM)
  - $\bar{F}(x, y, z) = \prod_M(1, 3, 5, 7)$  (POM)

$$\bar{F}(x, y, z) = m_1 + m_3 + m_5 + m_7.$$

$$F(x, y, z) = \sum_m(1, 3, 5, 7) = \prod_M(0, 2, 4, 6).$$

$$\bar{F}(x, y, z) = \sum_m(0, 2, 4, 6) = \prod_M(1, 3, 5, 7).$$

# Conversion Between Forms

To convert between sum-of-minterms and product-of-maxterms form (or vice-versa) we follow these steps:

- Find the function complement by swapping terms in the list with terms not in the list.   
 (تحويل)   
 وضع الأرقام التي لم تكن minterms في القائمة   
 المكملة على الموجودين بال (minterm) نقية.
- Change from products to sums, or vice versa.   
 (تحويل)   
 المكملة على الموجودين بال (minterm) نقية.

**Example:** Given  $F$  as before:  $F(x, y, z) = \sum_m(1,3,5,7)$

- Form the Complement:   
 3 Variables, means:  $2^3 = 8$ , means from (0-7)   
 من 0 إلى 7 (minterms).

$$\bar{F}(x, y, z) = \sum_m(0,2,4,6)$$

- Then use the other form with the same indices - this forms the complement again, giving the other form of the original function:

$$F(x, y, z) = \prod_M(0,2,4,6)$$

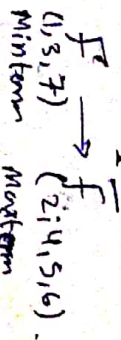
$$\bar{F}(x, y, z) = \prod_M(1,3,5,7)$$

# Important Properties of Minterms

- Maxterms are seldom used directly to express Boolean functions. \* Minterms are used more than Maxterms.

\*  $F$  (Function) يعني ال (complement) هو نقيض ال (maxterm) نلالة ال (function) (كفكرة)

- Minterms properties:
  - For  $n$  Boolean variables, there are  $2^n$  minterms (0 to  $2^n - 1$ )
  - Any Boolean function can be represented as a logical sum of minterms (SOM)



- The complement of a function contains those minterms not included in the original function. \* المنفي للاختزان يحتوي على الحدود الغير موجودة بالاختزان نفسه (مفككة لونها)

• A function that include all the  $2^n$  minterms is equal to 1.  $f(x,y) = \sum m(0,1,2,3) \Rightarrow$  The function contains all ( $2^n$ ) then it equals to (1).

\* ليس تحتوي ال (minterms) على جميع عناصر ال (function)  $f(x,y)$   $\Rightarrow$  لا يوجد (maxterms) ويكويه جوان ال (function)  $f(x,y)$  و ال (complement) هو ال (function)  $f(x,y)$  و

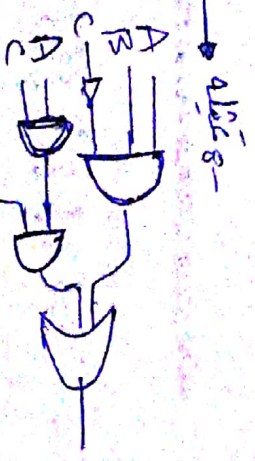
# Standard Forms

- Standard Sum-of-Products (SOP) form: equations are written as an OR of AND terms.  $F(x,y) = (x \cdot y) + (\bar{x} \cdot y) + (x \cdot \bar{y}) + (\bar{x} \cdot \bar{y})$ . \* لا يستطيع فيه اعتبار جمع الحدود على غير المتغير.
  - Standard Product-of-Sums (POS) form: equations are written as an AND of OR terms.  $F(x,y) = (x+y) \cdot (\bar{x}+y) \cdot (x+\bar{y}) \cdot (\bar{x}+\bar{y})$ . \* لا يستطيع فيه اعتبار جمع الحدود على غير المتغير.
- Examples:
- SOP:  $ABC + \bar{A}\bar{B}C + B$  → not (SOP) كل حد متعلق بالمتغيرين والحدود min terms
  - POS:  $(A+B) \cdot (A+\bar{B}+C) \cdot C$  → not (POS) متعلق بالحدود متعلق بالمتغيرين و min terms

These "mixed" forms are neither SOP nor POS

$(AB + C)(A + C)$  (SOP)   
 $ABC + AC(A + B)$  (POS)

\* محال لا   
 \* بالتوزيع يصبح (SOP) أو (POS)



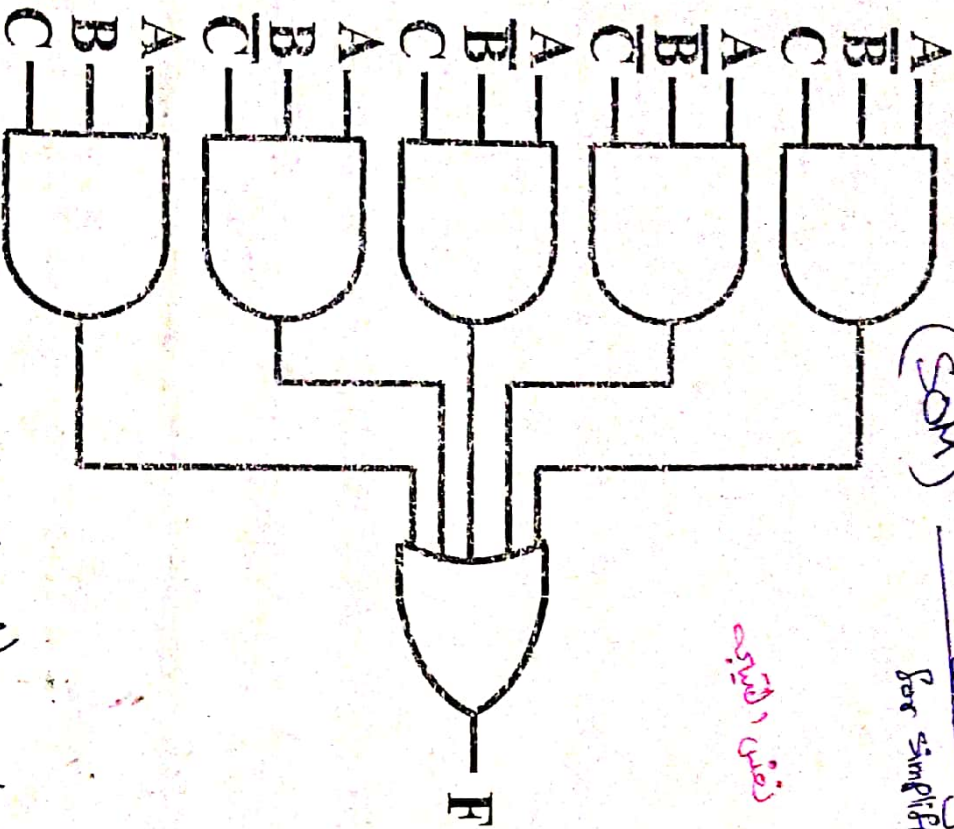






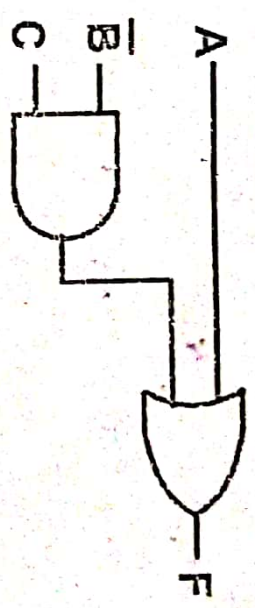
# AND/OR Two-level Implementation of SOP Expression

- The two implementations for F are shown below -- it is quite apparent which is simpler!



two levels

(and gate) then (OR gate)



two levels

(and gate) then (OR gate)

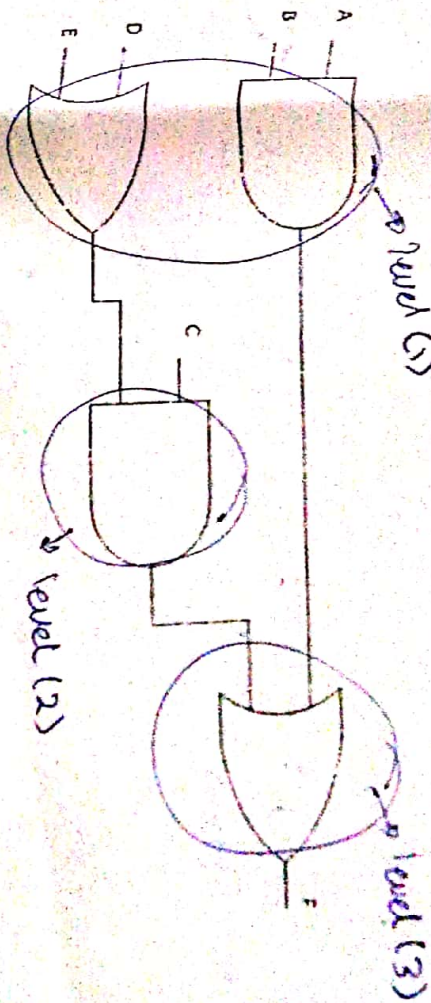
ولكن في هذه الحالة البسيط. و عدد البوابات اقل لتقليل (cost) الى

دعا البسيط \* الى (circuit) الى

# Two-level Implementation

Draw the logic diagram of the following boolean function:

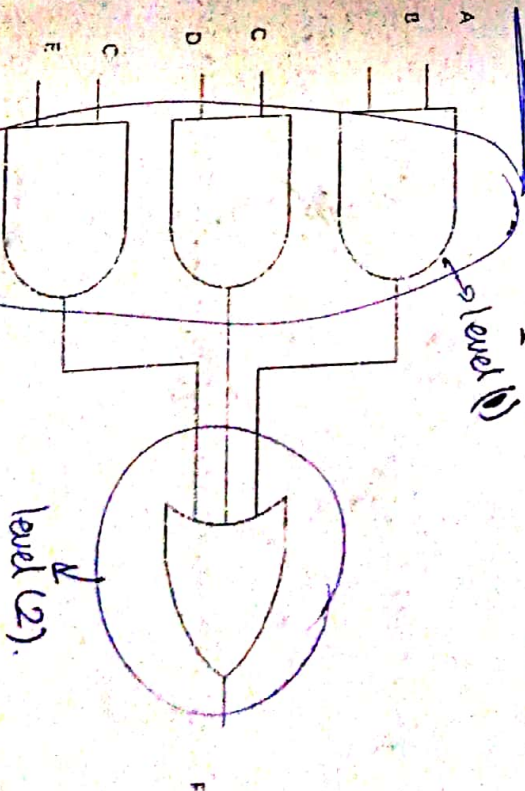
$$f = AB + C(D + E)$$



Represent the function using two-level implementation:

$$f = AB + CD + CE \rightarrow \text{SOP}$$

نسطر 8 - القليل باستعمال ال (gates) ال



# SOP and POS Observations

- The previous examples show that:
  - Canonical Forms (Sum-of-minterms, Product-of-Maxterms), or other standard forms (SOP, POS) differ in complexity
  - Boolean algebra can be used to <sup>المعادلة</sup> manipulate equations into simpler forms.
  - Simpler equations lead to simpler two-level implementations

## Questions:

- How can we attain a "simplest" expression?
- Is there only one minimum cost circuit?
- The next part will deal with these issues.

\* عند التحويل :  
SOP  $\leftrightarrow$  SOM  
POS  $\leftrightarrow$  POM  
تقل (cost)

# Literal Cost (L)

- Literal: a **variable** or its **complement** أى متغير أو متغيره المكمل
- Literal cost (L): the number of literal appearances in a Boolean expression corresponding to the logic circuit diagram

## Examples:

•  $F = BD + AB'C + AC'D'$

▪  $L = 8$  (Minimum cost  $\rightarrow$  Best solution)

$L = 8$  عدد المتغيرات هو نفس عدد المتغيرات Variables

•  $F = BD + AB'C + AB'D' + ABC'$

▪  $L = 11$

•  $F = (A + B)(A + D)(B + C + D')(B' + C' + D)$

▪  $L = 10$

# Gate Input Cost

$$(G) = L + n \text{ terms}$$

$$F = (A+B)(A+CD) + (B+C'D) + (B+C'D)$$

$G = 11 + 5 = 16$

Gate input cost (G): the number of inputs to the gates in the implementation corresponding exactly to the given equation or equations. (G: inverters not counted, GN: inverters counted)

For SOP and POS equations, it can be found from the equation(s) by finding the sum of:  $* G = L + \text{number of terms excluding single variable terms}$

All literal appearances (L)

The number of terms excluding single literal terms, (G) and

optionally, the number of distinct complemented single literals (GN).

Examples:

$F = BD + AB'C + AC'D$

$G = 11, GN = 14$  (Minimum cost  $\rightarrow$  Best solution)

$F = BD + AB'C + AB'D' + ABC'$

$G = 15, GN = 18$

$F = (A+B)(A+D)(B+C+D')(B'+C'+D)$

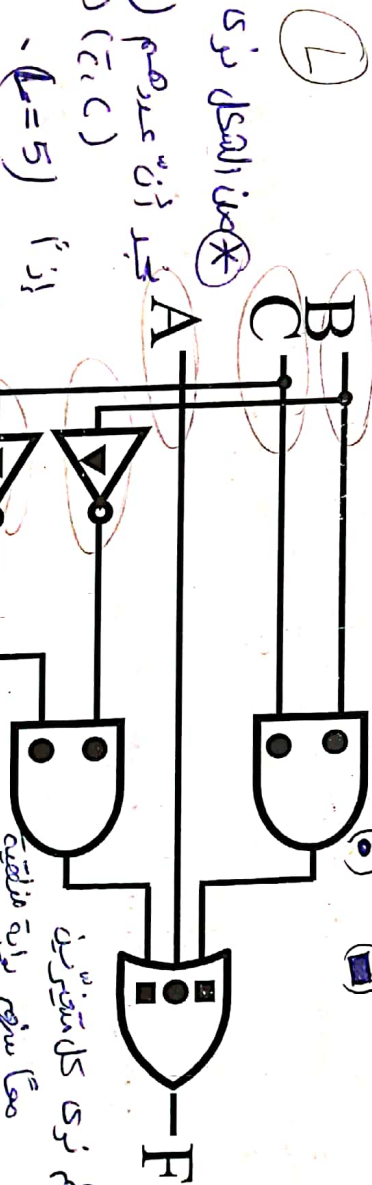
$G = 14, GN = 17$

$* 4 \text{ terms.}$   
 $* 10 \text{ number of variables}$   
 $* \text{Complemented variables: } 3$

# Cost Criteria (continued)

Example 1:  $\nabla \nabla \nabla \nabla \nabla$   $GN = G + 2 = 9$

$F = A + B + \bar{B}C + \bar{B}\bar{C}$   $L = 5$   $G = L + 2 = 7$

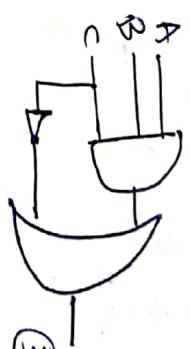


(input variables) نزي الالذي (5) وهم (B, A)  $(\bar{B}, B)$  نجد ذن عددهم (5)  $(A)$   $(\bar{C}, C)$   $(L = 5)$  اذنا

- **L** (literal count) counts the **AND** inputs and the **single** literal OR input. (5)
- **G** (gate input count) adds the remaining OR gate inputs  $5 + 2 = 7$
- **GN** (gate input count with NOTs) adds the inverter inputs  $5 + 2 + 2 = 9$

\* Example 2 -

$ABC + \bar{C}$



①  $L = 4$ . (ABC) هو الالذي  $(\bar{C})$  لا تغير term  $(\bar{C})$   $GN = 6$   $5 + 1$

②  $G = 5 = 4 + 1$   $(\bar{C})$   $GN = 6$   $(7 + 2 = GN)$  وهي  $(\bar{C})$   $GN = 6$   $(7 + 2 = GN)$  وهي

(complemented) الالذي  $(GN)$   $(7 + 2 = GN)$  وهي  $(7 + 2 = GN)$  وهي

# Cost Criteria (continued)

$G + 2 = 7$

ABCDE  
T1  
T2  
T3  
T4

## Example 2:

▪  $F = (A, B, C, D) = (ABC + D') \cdot C'$

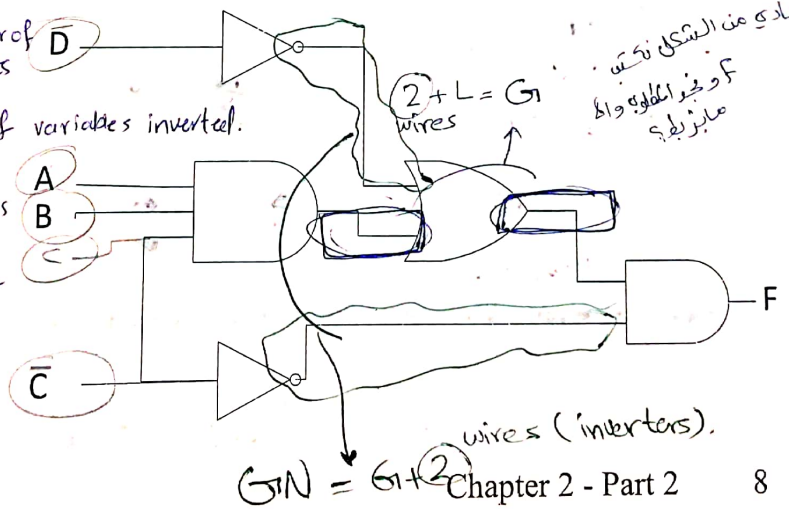
Singam  
Khaled  
MANAJEJAH

•  $L = 5$

•  $G = 5 + 2 = 7$  → because distributive law:  $\underbrace{ABC}_{\text{term 1}} + \underbrace{D'}_{\text{term 2}}$

•  $GN = 7 + 2 = 9$

- \*  $L$  = variables = inputs
- \*  $GN$  =  $G$  + number of variables inverted.
- \*  $G$  =  $L$  + number of terms that contains more than one variable.

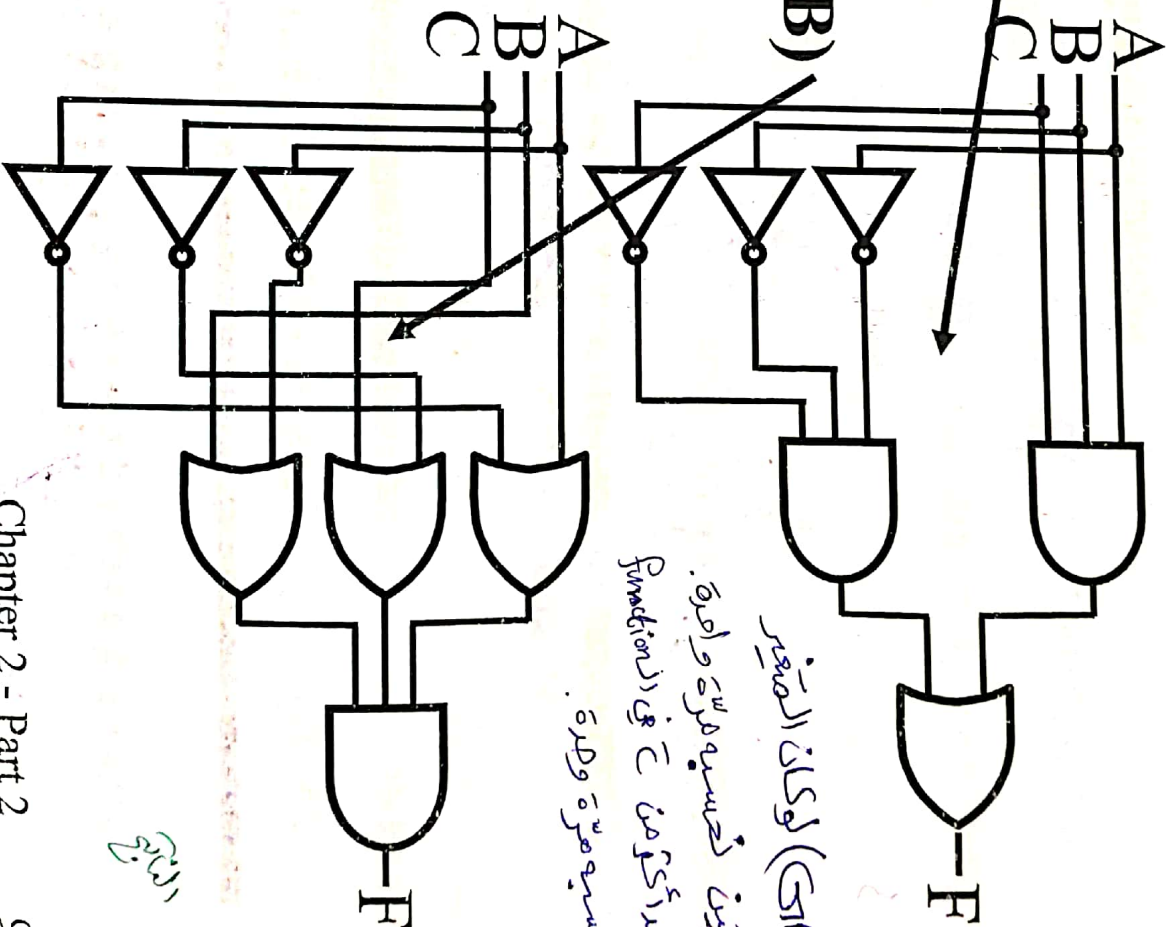


عدد من السلكين  
F ونجرب الحلوه وانا  
مبارك بدي

# Cost Criteria (continued)

## Example 3:

- $F = A B C + \bar{A} \bar{B} \bar{C}$
- $L = 6, G = 8, GN = 11$
- $F = (A + \bar{C})(\bar{B} + C)(\bar{A} + B)$
- $L = 6, G = 9, GN = 12$
- Same function and same literal cost, *same Variables number.*
- But first circuit has better gate input count and better gate input count with NOTs
- **Select it!**



\* في (GN) لو كان المتغير منفى مرتين نخصبه مرة واحدة. يوجد أكثر من C في الدوال فنحسبه مرة واحدة.

المتغير نفسه المتغير

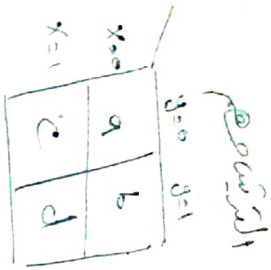


# Boolean Function Optimization

- Minimizing the gate input (or literal) cost of a (a set of) Boolean equation(s) reduces circuit cost
- We choose gate input cost
  - *Using theorems to simplify the equation*
- Boolean Algebra and graphical techniques are tools to minimize cost criteria values
  - $L = 7$
  - $G = 7 + 5 = 12$
  - $CN = 12 + 1$
  - $(A + (\overline{B}C)) \cdot D + E \cdot (C + D)$
  - $\underbrace{\underbrace{A + (\overline{B}C)}_2 \cdot D}_5 + \underbrace{E \cdot (C + D)}_3$
- Some important questions:
  - When do we stop trying to reduce the cost?
  - Do we know when we have a minimum cost?
- Treat optimum or near-optimum cost functions for two-level (SOP and POS) circuits
- Introduce a graphical technique using Karnaugh maps (K-maps, for short)

# Karnaugh Maps (K-map)

x	y	F
0	0	a
0	1	b
1	0	c
1	1	d



■ A K-map is a collection of squares  $2^n = \text{عدد المداخل}$    
 عدد المخرجات : n

- Graphical representation of the truth table
- Each square represents a minterm, or a maxterm, or a row in the truth table
- For n-variable, there are  $2^n$  squares
- The collection of squares is a graphical representation of a Boolean function  $\text{الوظيفة}$  (zeros) or (ones)  $\text{أو القيمة}$  \*   
 (maxterm)  $\text{أو الحد الأقصى}$  / (minterm)  $\text{أو الحد الأدنى}$    
 \*  $\text{تختلف}$    
 \*  $\text{بين كل مجموعتين متجاورتين مختلفتين}$    
 \*  $\text{ال (code) ؟ (one variable) ؟}$    
 001   
 010   
 011   
 110   
 111   
 (Gray code)
- Adjacent squares differ in the value of one variable  $\text{المتغير الواحد}$  (one variable)   
 001   
 010   
 011   
 110   
 111
- Alternative algebraic expressions for the same function are derived by recognizing patterns of squares  $\text{أنماط}$    
 (Gray code)

# Some Uses of K-Maps

---

- Finding optimum or near optimum
  - SOP and POS standard forms, and
  - two-level AND/OR and OR/AND circuit implementationsfor functions with small numbers of variables
- Visualizing concepts related to manipulating Boolean expressions, and
- Demonstrating concepts used by computer-aided design programs to simplify large circuits

# Two Variable Maps

## A 2-variable Karnaugh Map:

- Note that minterm  $m_0$  and minterm  $m_1$  are "adjacent" and differ in the value of the variable  $x$
- Similarly, minterm  $m_0$  and minterm  $m_2$  differ in the  $x$  variable
- Also,  $m_1$  and  $m_3$  differ in the  $x$  variable as well
- Finally,  $m_2$  and  $m_3$  differ in the value of the variable  $y$

	$y = 0$	$y = 1$
$x = 0$	$m_0 = \bar{x}\bar{y}$	$m_1 = \bar{x}y$
$x = 1$	$m_2 = x\bar{y}$	$m_3 = xy$

$$F = x + y$$

$$= x(y + \bar{y}) + x\bar{y}$$

$$= x(y + \bar{y}) + x\bar{y}$$

كل صر بعين متجاورين لمتحولين (Variable) في واحد.

# K-Map and Truth Tables

$2^n = \text{ستورودى سىز}$   $\rightarrow 2^n =$  rows and columns.

- The K-Map is just a different form of the truth table implementation of a function.
- Example: Two variable function
  - We choose a, b, c and d from the set  $\{0, 1\}$  to implement a particular function,  $F(x, y)$

Input Values (x, y)	F(x, y)
0 0	a
0 1	b
1 0	c
1 1	d

	$x \backslash y$	$y = 0$	$y = 1$
$x = 0$	0 0	a	b
$x = 1$	1 0	c	d

Truth Table

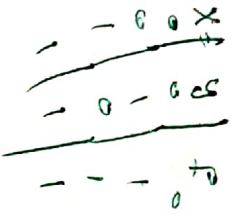
Same implementation for the function. K-Map

Truth table, K-maps.

# K-Map Function Representation

- Example:  $F(x, y) = x$

$2^2 = 4$  squares and 4 squares



$$F = \bar{x}y + x\bar{y} + xy$$

$$= \bar{x}y + x(\bar{y} + y)$$

$$= \bar{x}y + x$$

$$= y + x$$

$F(x, y) = x$	$y = 0$	$y = 1$
$x = 0$	0	0
$x = 1$	1	1

- For function  $F(x, y)$ , the two adjacent cells containing **1's** can be combined using the Minimization Theorem:

$$x \cdot (\bar{y} + y) = x$$

$$x \cdot 1 = x$$

function  $x$  Example 2:  $F(x, y) = y$ .

$\sum (1, 0, 3) \rightarrow$  مرتبة الازمنة جزيئية

$F(x, y) = y$	$y = 0$	$y = 1$
$x = 0$	0	1
$x = 1$	0	1

$x\bar{y} = m_0$      $x\bar{y} = m_2$      $x\bar{y} = m_1$      $x\bar{y} = m_3$   
 15    3    1    1

# K-Map Function Representation

■ Example:  $G(x, y) = x + y$

$$F(x, y) = \sum_m (1, 2, 3) \rightarrow$$

$$\bar{x}\bar{y} + x\bar{y} + xy \rightarrow$$

$$xy + x(y + \bar{y}) \rightarrow$$

$$x + \bar{x}y = x + y \text{ (Simplification theorem)}$$

$G(x, y) = x + y$	$y = 0$	$y = 1$
$x = 0$	0	1
$x = 1$	1	1

(x) ← اکتیو سیگنل

(1) ← سیگنال

اکتیو سیگنال (y) سیگنال

■ For  $G(x, y)$ , two pairs of adjacent cells containing 1's can be combined using the Minimization Theorem:

$$G(x, y) = (x\bar{y} + xy) + (\bar{x}y + xy)$$

$$G(x, y) = x + y$$

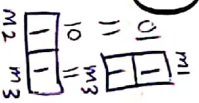
نیم مجموعہ

function 1

هو (x+y)

function 1  
یعنی جو ان

یکو ن هو جو ان

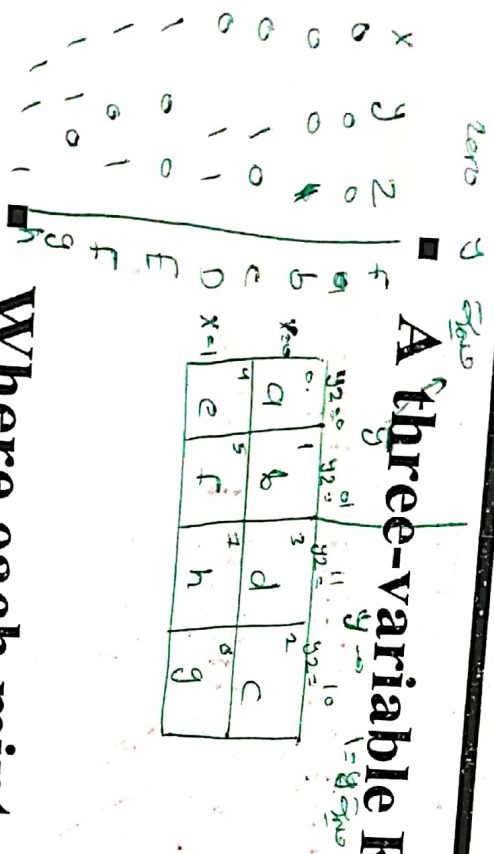


$$m1 : \bar{x}y + xy \rightarrow y(\bar{x} + x) = y \cdot 1 = y$$

$$m2 : x\bar{y} + xy \rightarrow x(\bar{y} + y) = x \cdot 1 = x$$

# Three Variable Maps

## A three-variable K-map:



	yz = 00	yz = 01	yz = 11	yz = 10
x = 0	m <sub>0</sub>	m <sub>1</sub>	m <sub>3</sub>	m <sub>2</sub>
x = 1	m <sub>4</sub>	m <sub>5</sub>	m <sub>7</sub>	m <sub>6</sub>

م تبديل المتغيرات لكي يتوافق مع المربع الثاني (1 bit) ←

Where each minterm corresponds to the product terms:

\*  $F(x, y, z) = \prod M(0, 1, 3, 4) \rightarrow$  zeros  
 minterms.

\*  $f(x, y, z) = \sum m(2, 5, 6, 7) \rightarrow$  ones  
 minterms.

	yz = 00	yz = 01	yz = 11	yz = 10
x = 0	$\bar{x}\bar{y}\bar{z}$	$\bar{x}\bar{y}z$	$\bar{x}yz$	$\bar{x}y\bar{z}$
x = 1	$x\bar{y}\bar{z}$	$x\bar{y}z$	$xyz$	$xy\bar{z}$

(m<sub>3</sub> و m<sub>2</sub>) و (m<sub>7</sub> و m<sub>6</sub>)

- Note that if the binary value for an index differs in one bit position, the minterms are adjacent on the K-Map



# Alternative Map Labeling

- Map use largely involves:
  - Entering values into the map, and
  - Reading off product terms from the map
- Alternate labelings are useful:

\* فنكر الرموز  
المشتركة فيهم  
قطر.

	$\bar{Y}$	Y	
$\bar{X}$	0	1	3
X	4	5	6

by default :  $\bar{Z}$

	YZ	00	01	11	10
X		0	1	3	2
$\bar{X}$	0	0	1	3	2
X	1	4	5	7	6

Handwritten notes on the table:  
 - Above the table:  $\bar{Y}$  by default  
 - Above the 00 column:  $\bar{X}\bar{Y}\bar{Z} = 000$   
 - Above the 01 column:  $\bar{X}Y\bar{Z} = 001$   
 - Above the 11 column:  $\bar{X}YZ = 011$   
 - Above the 10 column:  $\bar{X}Y\bar{Z} = 010$   
 - Below the table: A bracket under columns 01, 11, and 10 is labeled Z. Arrows point from this Z to the Z in the bottom row of the table.

# Example Functions

By convention, we represent the minterms of  $F$  by a "1" in the map and leave the minterms of  $\bar{F}$  blank (0) blank.

Example:

$$F(x, y, z) = \sum m(2, 3, 4, 5)$$

$$\bar{F} = \sum M(0, 1, 6, 7)$$

نصنعهم من خريطة د ر  $xy$

	$\bar{y}$	$y$	
$\bar{x}$	0	1	
$x$	4	5	
	$\bar{z}$	$z$	
			3
			2
			7
			6

most significant =  $x$

least significant =  $z$

$$G(a, b, c) = \sum m(3, 4, 6, 7)$$

$x=0$

$F = xy + \bar{x}y$  (zeros by default)

	$\bar{a}$	$a$	
	0	1	
	4	5	
	$\bar{c}$	$c$	
			3
			2
			7
			6

$F = bc + ab + a\bar{b}\bar{c}$

- Learn the locations of the 8 indices based on the variable order shown (X, most significant and Z, least significant) on the map boundaries

# Steps for using K-Maps to Simplify Boolean Functions



Enter the function on the K-Map

Function can be given in truth table, shorthand notation, SOP, ... etc

	0	1
0	1	1
1	0	1

- Example:
  - $F(x, y) = \bar{x} + xy$  (SOP)
  - $F(x, y) = \sum m(0, 1, 3)$  (SOM)

$$\bar{x} \cdot (y + \bar{y}) + x \cdot y = \bar{x}y + \bar{x}\bar{y} + xy = m_0 + m_1 + m_3$$

2<sup>2</sup> = 4 squares  
The K-map will be.

x	y	F(x, y)
0	0	1
0	1	1
1	0	0
1	1	1

	0	1
0	1	1
1	0	1

$F = \bar{x} + y$

Combining squares for simplification

- Rectangles that include power of 2 squares {1, 2, 4, 8, ...}
- Goal: Fewest rectangles that cover all 1's → as large as possible

Determine if any rectangle is not needed (not essential)

Read-off the SOP terms

ليكن ما يطرحه لنا (المصطلح) من الـ SOP

4 أحسن من 4  
2 أحسن من 2  
ليكن ما يطرحه لنا

نقطة ضعف الـ SOP هي الـ SOP التي فيها  
شيء إضافي لا يلزمه الـ SOP

# Combining Squares

نخل اکبر دانه شاموی اکبر عدد ۴

- By <sup>2.07</sup> combining squares, we reduce number of literals in a product term, reducing the literal cost, thereby reducing the other two cost criteria

\* On a 2-variable K-Map:  $F(x, y)$ .

- One square represents a minterm with two variables
- Two adjacent squares represent a product term with one variable
- Four "adjacent" terms is the function of all ones (no variables) = 1.  $2^2 = 4$

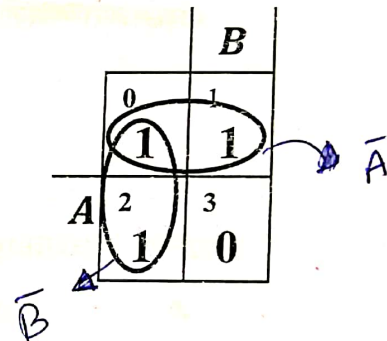
\* On a 3-variable K-Map:  $F(x, y, z)$

- One square represents a minterm with three variables
- Two adjacent squares represent a product term with two variables
- Four "adjacent" terms represent a product term with one variable
- Eight "adjacent" terms is the function of all ones (no variables) = 1.  $2^3 = 8$

# Example: Combining Squares

- Example:  $F(A, B) = \sum_m(0,1,2)$

$$F(A, B) = \overset{m_0}{\bar{A}\bar{B}} + \overset{m_1}{\bar{A}B} + \overset{m_2}{A\bar{B}}$$



- Using Distributive law

- $F(A, B) = \bar{A} + A\bar{B}$

- Using simplification theorem

- $F(A, B) = \underline{\bar{A} + \bar{B}}$

- Thus, every two adjacent terms that form a  $2 \times 1$  rectangle correspond to a product term with one variable**

# Example: Combining Squares

- Example:  $F(x, y, z) = \sum_m(2,3,6,7)$

$$F(x, y, z) = \sum_m(1,3,5,7) = \prod_M(0,2,4,6)$$

$$F(x, y, z) = \bar{x}\bar{y}\bar{z} + \bar{x}y\bar{z} + x\bar{y}\bar{z} + xyz$$

- Using Distributive law

$$F(x, y, z) = \bar{x}y + xy$$

\* when we have a function with 2 variables,

- Using Distributive law again.

$$F(x, y, z) = y$$

- Thus, the four adjacent terms that form a  $2 \times 2$  square correspond to the term "y"

By Boolean Expressions using theorems.

$F(x, y, z) = \sum_m(0, 2, 4, 6) = \prod_M(1, 3, 5, 7)$

	$F(x, y, z)$		
	0	1	3
x	4	5	2
			7
			6
			1
			z

Handwritten notes on the table: A box around cells (0,1), (1,1), (0,3), (1,3) is labeled "y". A box around cells (0,3), (1,3), (0,7), (1,7) is labeled "y".

\* عند تبسيط الـ (ones) صلاحيات عدد الـ Variables بعد الـ (1) .

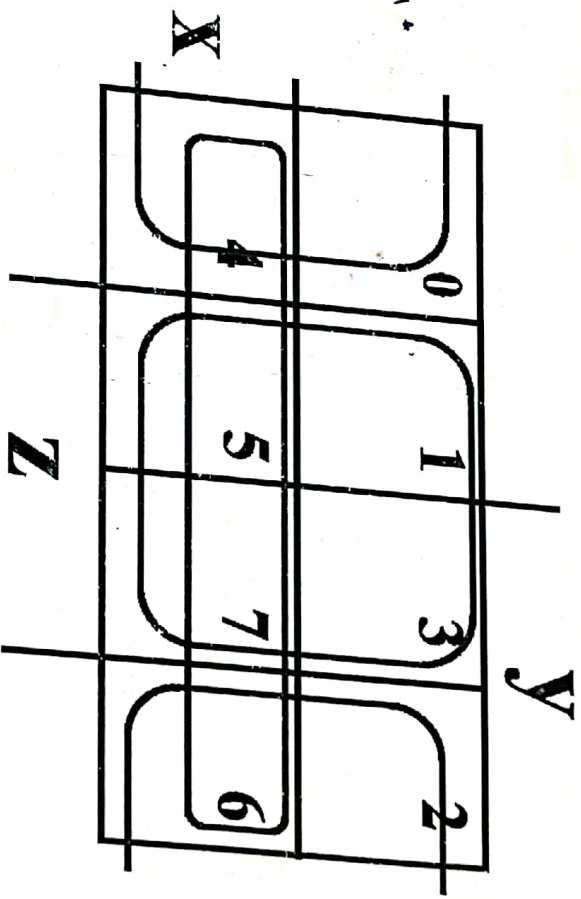
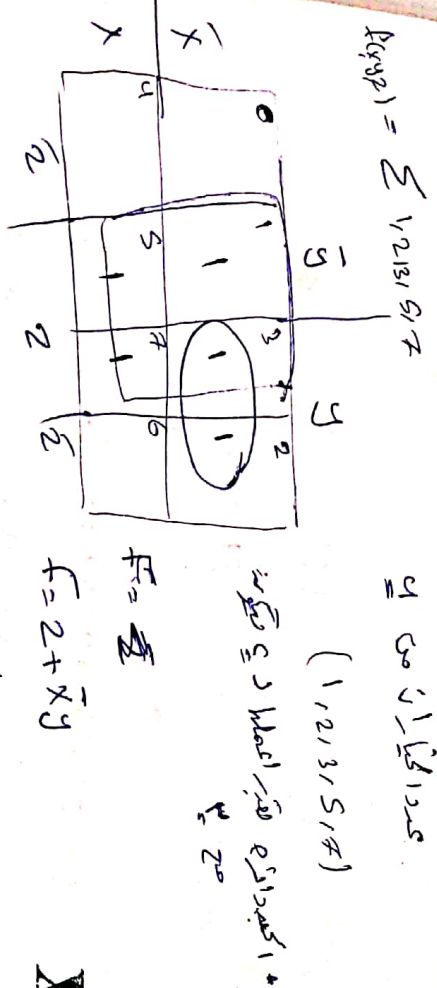






# Three-Variable Maps

Example shapes of 4-cell Rectangles:



Read off the product terms for the rectangles shown:

- $Rect(1,3,5,7) = Z$
- $Rect(0,2,4,6) = \bar{Z}$
- $Rect(4,5,6,7) = X$

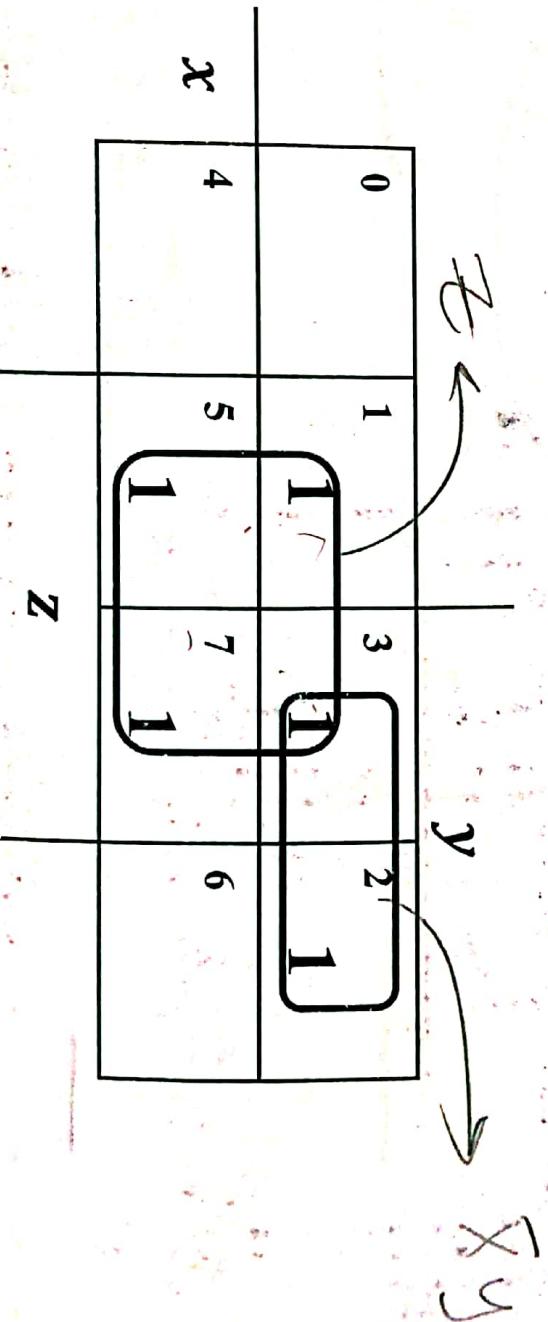
$$(0,1,3,2) = \bar{X}$$

$$(4,5,7,6) = X$$

# Three Variable Maps

- **K-maps** can be used to **simplify Boolean functions** by **systematic methods**. Terms are **selected to cover the "1s" in the map**.

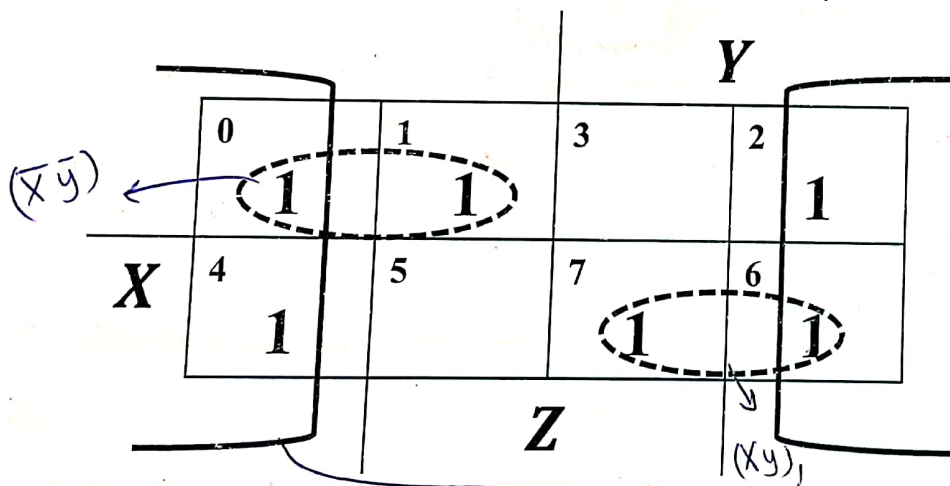
- Example: Simplify  $F(x, y, z) = \sum m(1, 2, 3, 5, 7)$



$$F(x, y, z) = z + \bar{x}y$$

# Three-Variable Map Simplification

- Use a K-map to find an optimum SOP equation for  $F(X, Y, Z) = \sum_m(0,1,2,4,6,7)$



اول احسنه فيدور مع (4) اذا حاقص  
 بعد (2)  
 اهننا الاصطلاح (2)

$$F(X, Y, Z) = \bar{Z} + \bar{X}\bar{Y} + XY$$

$$\bar{F} = \bar{z} \cdot \overline{xy} \cdot \overline{xy}$$

$$\bar{F} = z \cdot (x+y) \cdot (\bar{x}\bar{y})$$

# Four Variable Maps

- Map and location of minterms

$F(W, X, Y, Z)$ :

$\bar{X}, X$      $\bar{Y}, Y$      $\bar{Z}, Z$   
 اطاقات بضمير ال  
 من الورد  
 الورد  
 حتمه سطره ال  
 الورد

بضمير ال    الورد

الورد سطره الورد كل الورد  
 الورد سطره الورد  
 الورد الورد

	$\bar{Z}$		$Z$		
	00		01		
$\bar{W}$	0 $\bar{W}\bar{X}\bar{Y}\bar{Z}$ 0000	1 $\bar{W}\bar{X}YZ$ 0001	3 $\bar{W}X\bar{Y}Z$ 0011	2 $\bar{W}XY\bar{Z}$ 0010	
11	4	5	7	6	
10	12	13	15	14	
$W$	8	9	11	10	
					$X$
					$\bar{X}$
					$Y$
					$\bar{Y}$
					$Z$



# Four Variable Terms

Four variable maps can have rectangles corresponding to:

- A single 1: 4 variables (i.e. Minterm)  $F = (4)$  variables.
- Two 1's: 3 variables  $F = (3)$  variables.
- Four 1's: 2 variables  $F = (2)$  variables.
- Eight 1's: 1 variable  $F = (1)$  variable.
- Sixteen 1's: zero variables (function of all ones)  $F = 1$ .

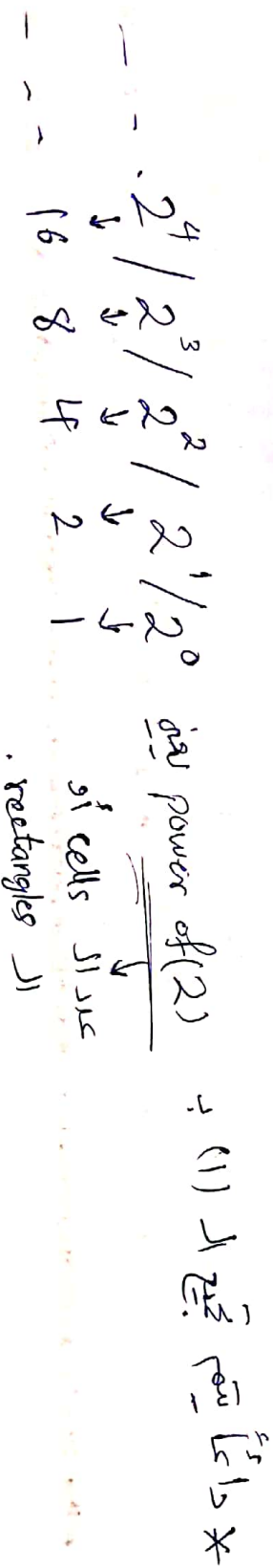
which means:

المصطلح (مصطلح)  $F = (4)$  variables.

المصطلح (مصطلح)  $F = (3)$  variables.

المصطلح (مصطلح)  $F = (2)$  variables.

المصطلح (مصطلح)  $F = (1)$  variable.



# Four-Variable Maps

- Example shapes of 4-cell rectangles:

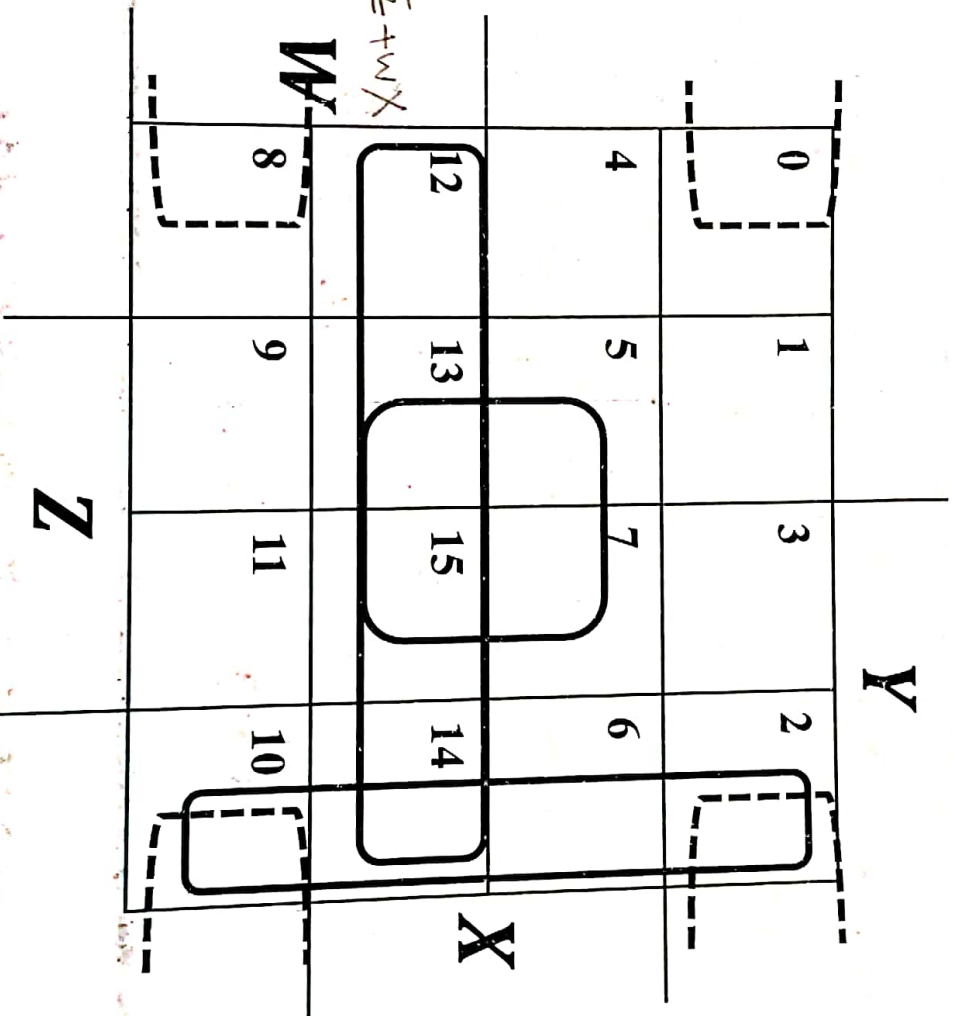
$$\text{rec}(5, 7, 13, 15) : XZ.$$

$$\text{rec}(0, 2, 8, 10) : \bar{X}\bar{Z}.$$

$$\text{rec}(2, 6, 10, 14) : Y\bar{Z}.$$

$$\text{rec}(12, 13, 14, 15) : WX.$$

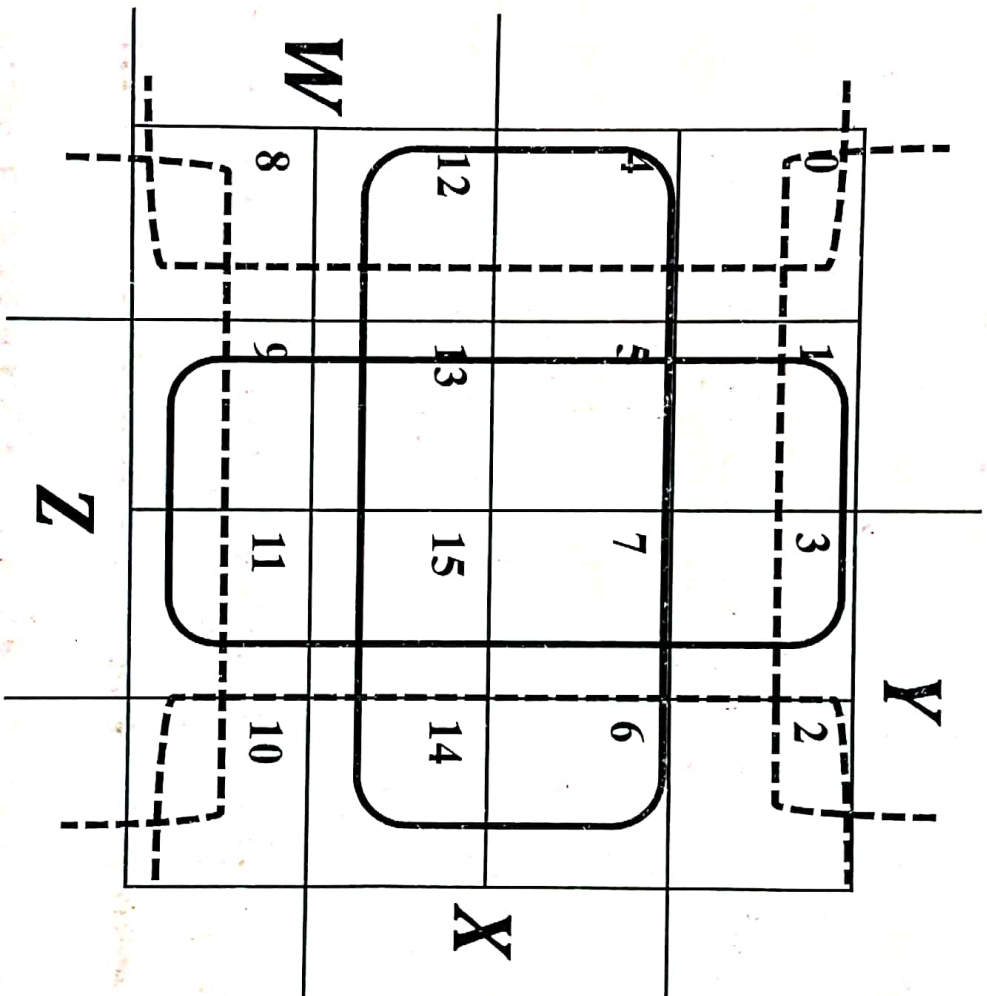
$$f(W, X, Y, Z) = XZ + \bar{X}\bar{Z} + Y\bar{Z} + WX$$



# Four-Variable Maps

---

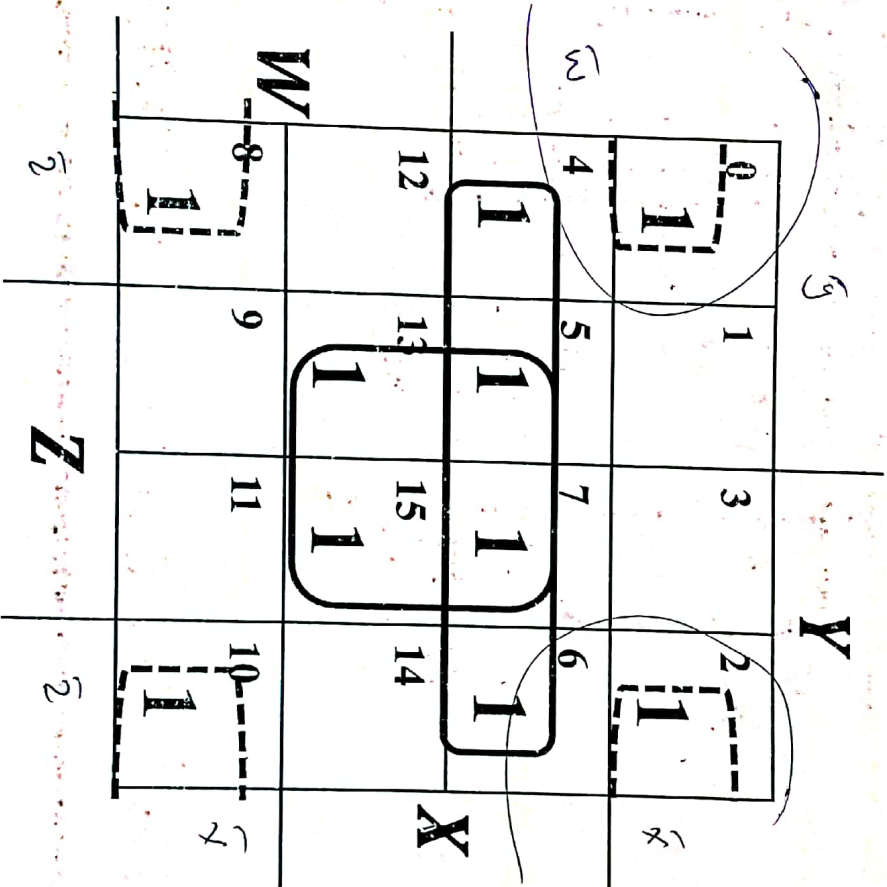
- Example shapes of 8-cell rectangles:



# Four-Variable Map Simplification

$F(W, X, Y, Z) = \sum m(0, 2, 4, 5, 6, 7, 8, 10, 13, 15)$

Min عدد 10  
 10 1 1 1 1 1 1 1 1 1  
 1 1 1 1 1 1 1 1 1 1



\* ال (prime implicant) هو أن نجعل أكبر عدد من ال (ones) بتجميعهم

0, 1, 2, 3  
 0, 1, 2, 3  
 0, 1, 2, 3

$F(W, X, Y, Z) = XZ + \bar{X}\bar{Z} + W\bar{X}$



# Four-Variable Map Simplification

▪  $F(W, X, Y, Z) = \sum m(3, 4, 5, 7, 9, 13, 14, 15)$

\* Essential Implicants:-

- \* rec (9,13)
- \* rec (3,7)
- \* rec (4,5)
- \* rec (14,15)

5 - prime implicants.

rec (5,7) → Prime implicant

لا يوجد rec (5,7,13,15) Prime implicant  
لأنه يوجد rec (5,7,13,15) وهي أكبر

	$\bar{Y}$		Y	
	0	1	3	2
$\bar{X}$	4	5	7	6
X	12	13	15	14
W	8	9	11	10
	$\bar{Z}$		Z	

\* Non-Essential :-

rec (5, 7, 13, 15) رتبة 2  
مكرر وال (ones) فيه صوتين بأكثر  
من صريح 1 في

\* لا توجد هنا الجداول فاصلي ومرتبة  
Function ولا يوجد أي الجداول 1,0  
بعض نوبة 1 في مكانه كل قاطع فيه أكثر الالام  
بعض 2 في مكانه قاطعها  
بعض 1 في مكانه قاطعها

$F(W, X, Y, Z) = \bar{W}YZ + \bar{W}X\bar{Y} + WXY + W\bar{Y}Z$

# Systematic Simplification

الدالة المنطقية  
Function & K-map term.

\* Implicant  $\equiv$  term.

Essential

Non-Essential

**Prime Implicant:** is a product term obtained by combining the **maximum** possible number of adjacent squares in the map into a rectangle with the number of squares a power of 2

\* The maximum possible number of adjacent terms.  
\* الحد الأقصى لعدد المصطلحات المجاورة الممكنة.

A prime implicant is called an **Essential Prime Implicant** if it is the **only** prime implicant that covers (includes) one or more minterms

مصطلح الدالة المنطقية الوحيد الذي يغطي

أحد أو أكثر من المصطلحات المنطقية

أحد مصطلح الدالة المنطقية فقط

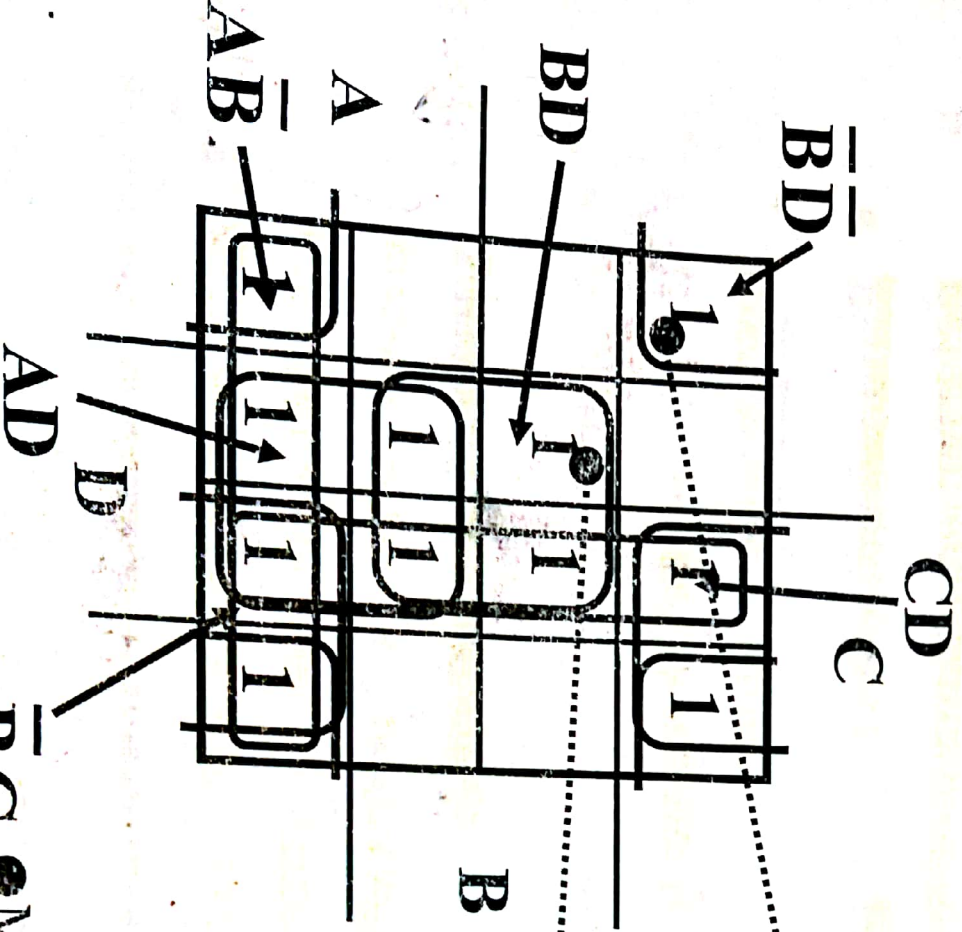
بعض المصطلحات المنطقية فقط

Prime Implicants and Essential Prime Implicants can be determined by inspection of a K-Map

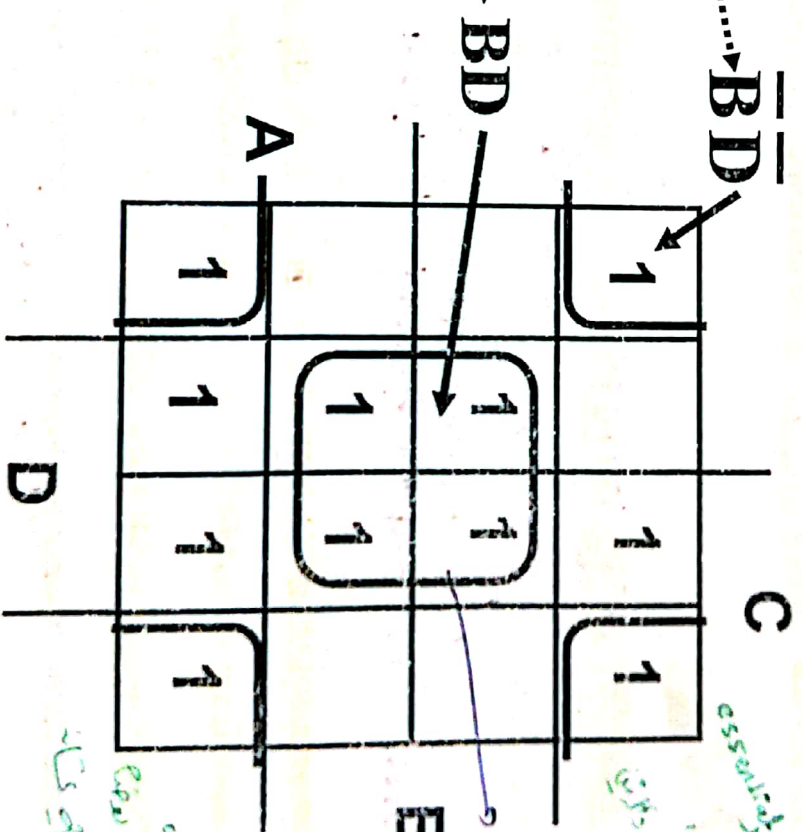
- A set of prime implicants "covers all minterms" if, for each minterm of the function, at least one prime implicant in the set of prime implicants includes the minterm

# Example of Prime Implicants

- Find ALL Prime Implicants



## ESSENTIAL Prime Implicants



$\bar{B}C$  ● Minterms covered by single prime implicant

Essential prime implicants  
 Essential prime implicants  
 Essential prime implicants  
 Essential prime implicants  
 Essential prime implicants

Prime implicant  
 Prime implicant  
 Prime implicant  
 Prime implicant

# Prime Implicant Practice

- Find all prime implicants for:

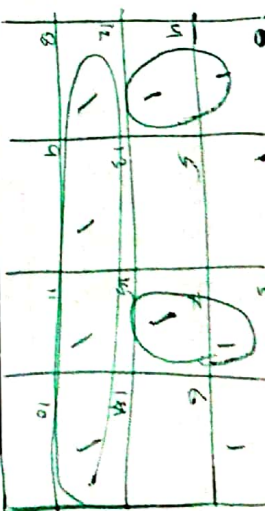
$$F(A, B, C, D) = \sum_m (0, 2, 3, 8, 9, 10, 11, 12, 13, 14, 15)$$

- Prime Implicants: [Essentials]

- $A$  *essential*
- $\bar{B}C$  *essential*
- $\bar{B}\bar{D}$  *essential*



# Another Example



- Find all prime implicants for:

$$G(A, B, C, D) = \sum_m (0, 2, 3, 4, 7, 12, 13, 14, 15)$$

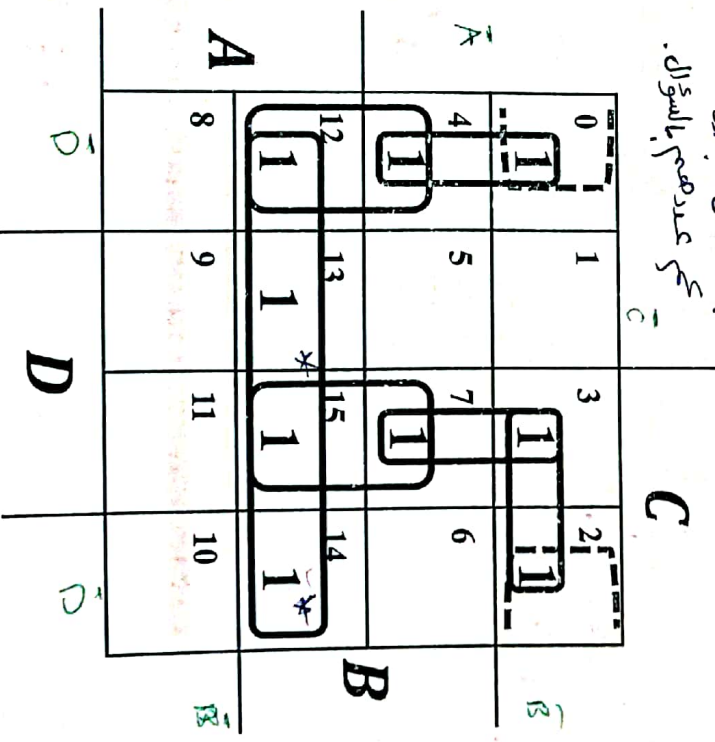
- Hint: There are seven prime implicants!

- Prime Implicants:

- $AB \rightarrow$  (the only one (Essential).)
- $BCD$  (7, 11, 15)
- $BC\bar{D}$  (11, 12)
- $\bar{A}CD$  (5, 4)
- $\bar{A}C\bar{D}$  (10, 11)
- $\bar{A}\bar{B}C$  (2, 3)
- $\bar{A}\bar{B}\bar{D}$  (0, 1, 2, 3)

\* كل (one) سوكلي  
 لا يترك من مبرقة لا يترك  
 بل يترك نوا (non-essential)

فان عددهم كثير  
 بالامتحان نختار  
 التي عددتهم بالسؤال.



# Optimization Algorithm

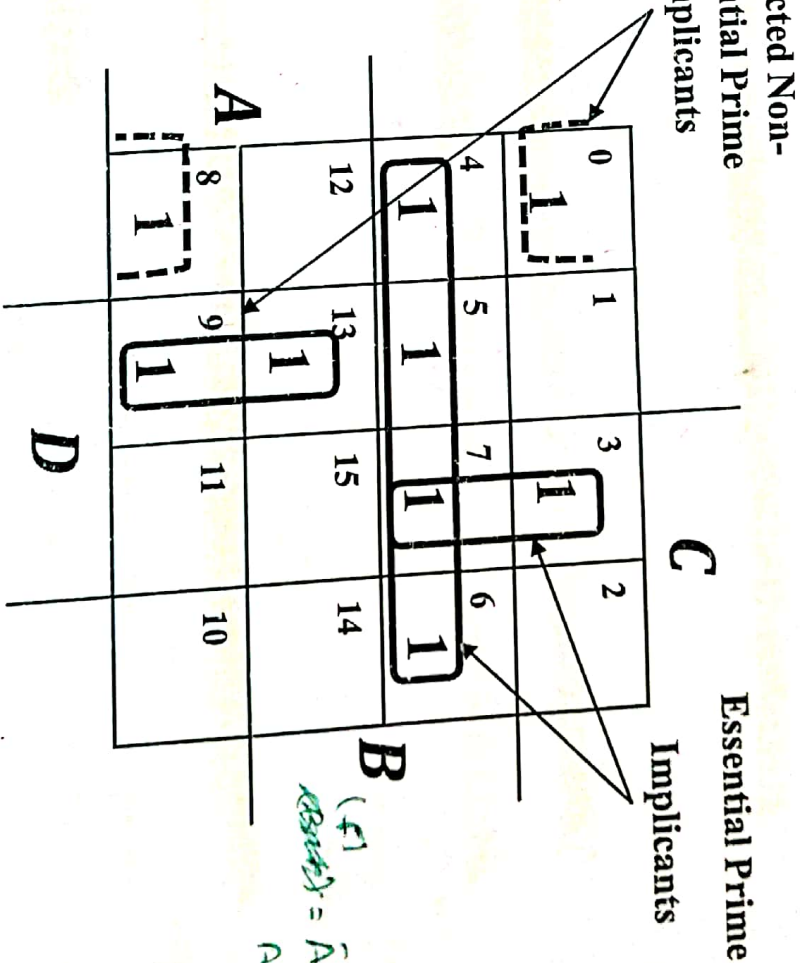
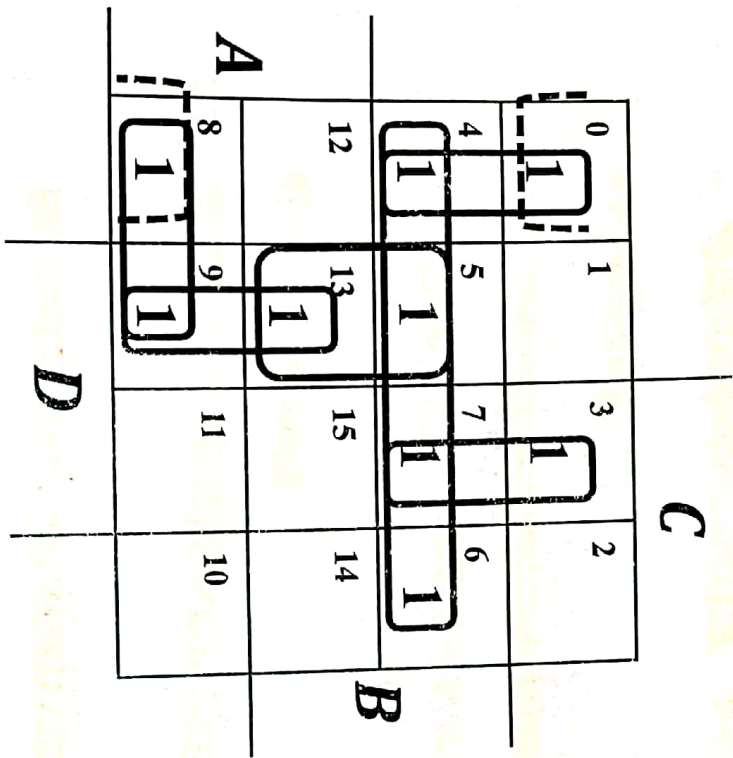
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1. Find all prime implicants
2. Include all essential prime implicants in the solution
3. Select a <sup>1</sup> minimum <sup>2</sup> cost <sup>3</sup> set of non-essential prime implicants to cover all minterms not yet covered
  - Selection Rule: Minimize the overlap among prime implicants as much as possible. In particular, in the final solution, make sure that each prime implicant selected includes at least one minterm not included in any other prime implicant selected

no need to add

# Selection Rule Example

- Simplify  $F(A, B, C, D)$  given on the K-map



$(F)$   
 $(B+C) = \bar{A}CD + \bar{A}B + A\bar{C}D + \bar{B}\bar{C}\bar{D}$

Prime Implicants

Essential and Selected Non-essential

Prime Implicants

Handwritten note: "Joi (ones) are used in joi's (prime implicants)"

Handwritten note: "not prime implicant."

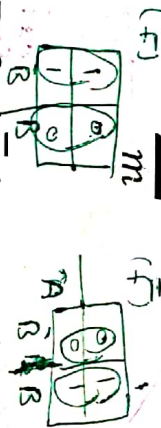
Handwritten note: "a prime implicant."

# Product of Sums Example

(1)  $m$ 's  $\leftarrow F$   $\leftarrow$  minterms  $\leftarrow$  SOP  $\leftarrow$  K-map  
 (2)  $M$ 's  $\leftarrow \bar{F}$   $\leftarrow$  maxterms  $\leftarrow$  POS  $\leftarrow$  K-map  
 in the k-map

- Find the optimum POS solution for:

$$F(A, B, C, D) = \sum (1, 3, 9, 11, 12, 13, 14, 15)$$



Solution:

- Find optimized SOP for  $F$  by combining 1's in K-Map of  $F$
- Complement  $\bar{F}$  to obtain optimized POS for  $F$

$$\bar{F}(A, B, C, D) = \bar{A}\bar{B} + \bar{B}\bar{D}$$

- Using Demorgan's Law:

$$F(A, B, C, D) = (A + \bar{B})(B + D)$$

	C			
	0	1	3	2
4	0	1	1	0
5	0	0	0	0
7	0	0	0	0
6	0	0	0	0
	B			
12	1	1	1	1
13	1	1	1	1
15	1	1	1	1
14	1	1	1	1
	D			
8	0	1	1	0
9	0	1	1	0
11	0	1	1	0
10	0	1	1	0
	A			



# Example

- Find the optimum POS and SOP solution for:

$$F(A, B, C, D) = \prod (0, 2, 4, 5, 6, 7)$$

$$F(A, B, C, D) = \sum_m (1, 3, 8, 9, 10, 11, 12, 13)$$

- POS solution (Red):
  - Find optimized SOP for  $\bar{F}$  by combining 0's in K-Map of  $F$
  - Complement  $\bar{F}$  to obtain optimized POS for  $F$

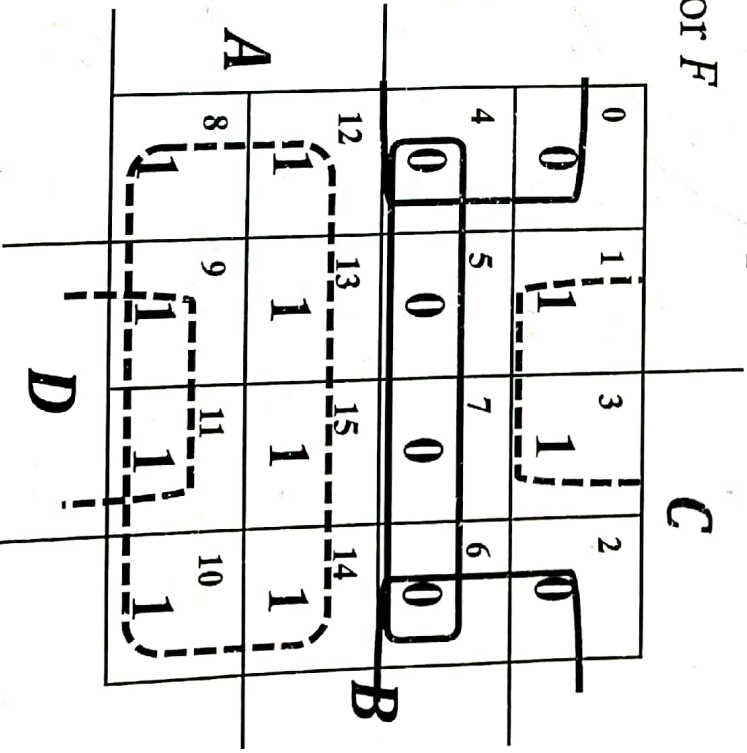
$$\bar{F}(A, B, C, D) = \bar{A}B + \bar{A}\bar{D}$$

$$F(A, B, C, D) = (A + \bar{B})(A + D)$$

- SOP solution (Blue):

- Combining 1's in K-Map of  $F$

$$F(A, B, C, D) = A + \bar{B}D$$



SOP( $\bar{F}$ )  $\Rightarrow$  POS( $F$ )

# Don't Cares in K-Maps

BCD (0-9)

- Incompletely specified functions: Sometimes a function table or map contains entries for which it is known:

  - the input values for the minterm will never occur, or
  - The output value for the minterm is not used
- In these cases, the output value is defined as a "don't care"
- By placing "don't cares" (an "x" entry) in the function table or map, the cost of the logic circuit may be lowered
- Example: A logic function having the binary codes for the BCD digits as its inputs. Only the codes for 0 through 9 are used. The six codes, 1010 through 1111 never occur, so the output values for these codes are "x" to represent "don't cares"
- "Don't care" minterms cannot be replaced with 1's or 0's because that would require the function to be always 1 or 0 for the associated input combination

\* استخراج قيم لا تظهر كقيم  
 حسب (الاضداد في ال  
 (0) ال  
 من ال  
 (K-map)  
 ال

# Example: BCD "5 or More"

$>= 5$

The map below gives a function  $F(w, x, y, z)$  which is defined as "5 or more" over BCD inputs. With the don't cares used for the 6 non-BCD combinations:

If don't cares are treated as 1's (Red):

$F_1(w, x, y, z) = w + xy + xz$

$G = 7$

$L = 5$

If don't cares are treated as 0's (Blue):

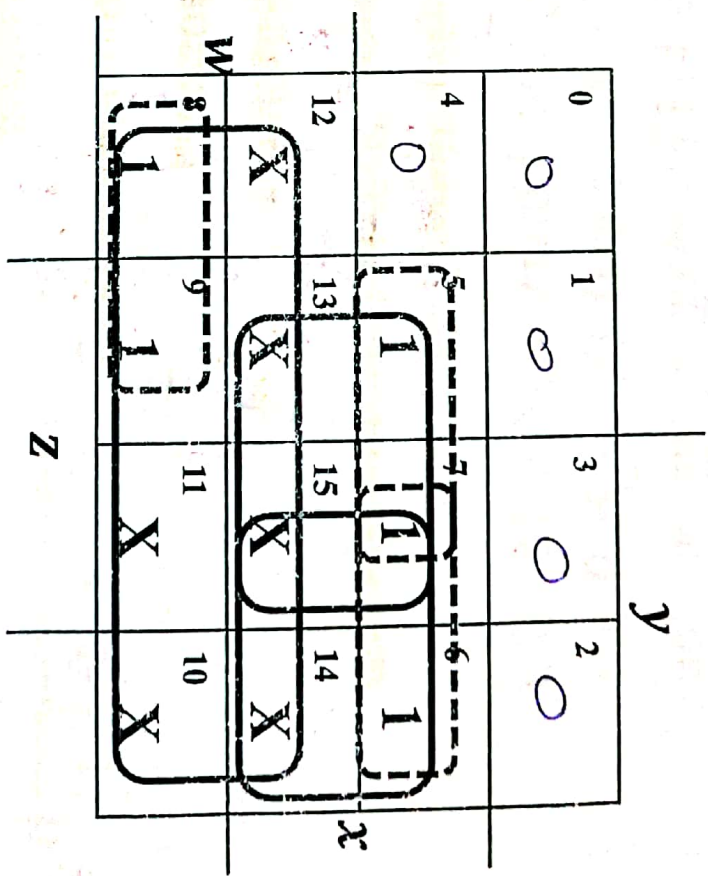
$F_2(w, x, y, z) = \bar{w}xz + \bar{w}xy + w\bar{x}\bar{y}$

$G = 12$

$L = 9$

$G_N = 15$

(red) (function)  $\rightarrow$   $\bar{w}xz$   $\bar{w}xy$   $w\bar{x}\bar{y}$  (ones)  $\rightarrow$   $\bar{w}xz$   $\bar{w}xy$   $w\bar{x}\bar{y}$   $\bar{w}xz$   $\bar{w}xy$   $w\bar{x}\bar{y}$   $\bar{w}xz$   $\bar{w}xy$   $w\bar{x}\bar{y}$   $\bar{w}xz$   $\bar{w}xy$   $w\bar{x}\bar{y}$  \*

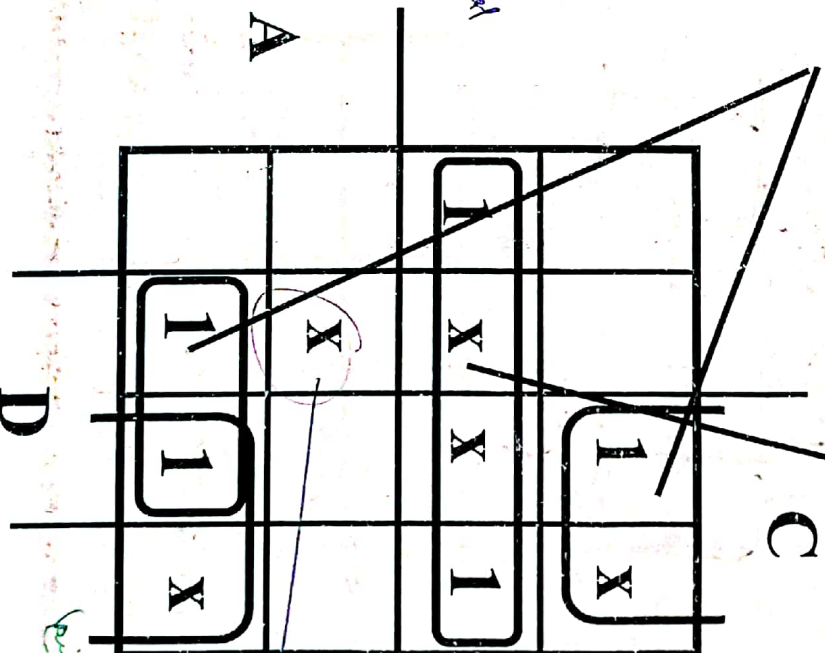
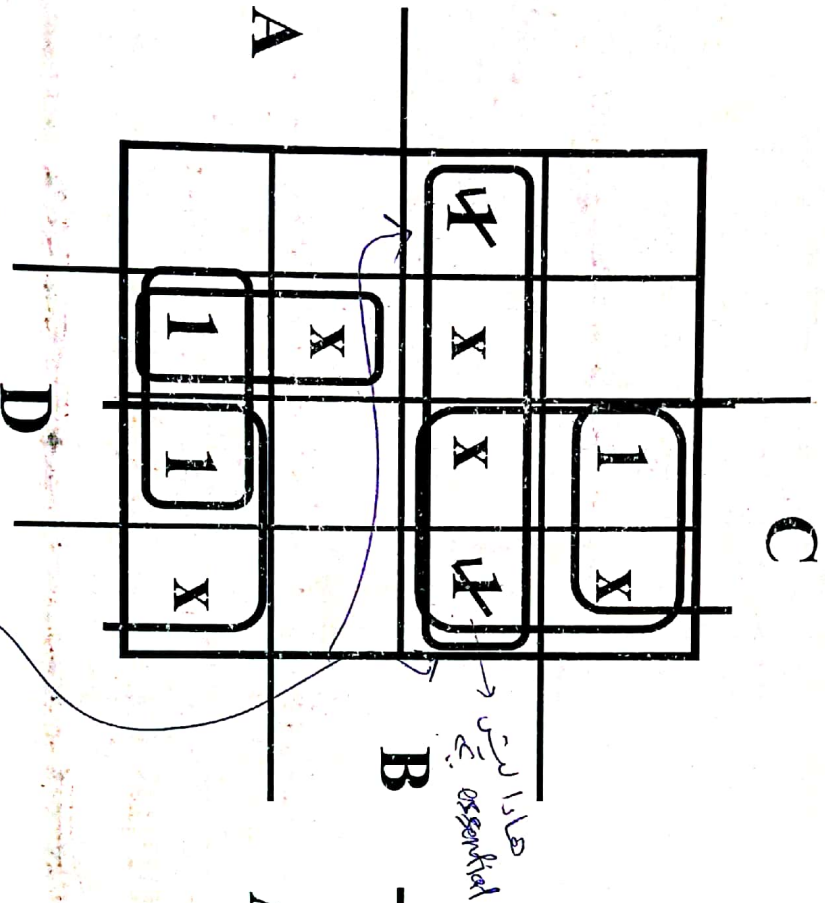


For this particular function, cost  $G$  for the POS solution for  $F(w, x, y, z)$  is not changed by using the don't cares

- Choose the one less inverters (i.e. less  $G_N$ )

# Selection Rule Example with Don't Cares

- Simplify  $F(A, B, C, D)$  given on the K-map.



Selected Essential

\* لاحظ: تغيير الـ (X)  
هنا كما يتبين (1)

التي تظهر أكبر عدد من الـ (ones) ولكننا نراهم يتم عمله (X) في مبرينات لا يوجد (على)

لا فضل (X) من (1) في كارد

Minterms covered by essential prime implicants

$F(A, B, C, D) = \bar{A}B + \bar{B}C + A\bar{C}D$   
 $G_1 = 10$ .  $G_2 = 13$ .

\*  $F(A, B, C, D) = \bar{A}B + \bar{B}C + AB\bar{D}$

# Product of Sums with Don't Care

## Example

- Find the optimum POS solution for:

$$F(A, B, C, D) = \sum_m (3, 9, 11, 12, 13, 14, 15) + \sum_{d \text{ (not in } m)} (1, 4, 6)$$

(ones)

	C			
	0	1	3	2
A	12	1	1	0
	8	X	0	X
D	9	0	1	0
	5	1	0	1
	B			
	13	15	14	10
	1	1	1	0

	C			
	0	1	3	2
A	12	1	1	0
	8	X	0	X
D	9	0	1	0
	5	1	0	1
	B			
	13	15	14	10
	1	1	1	0

~~$F(A, B, C, D) = (A + B + C + D)$~~

\*  $F_{\text{opt}} = AB + \bar{B}D$ .  $\rightarrow$  11 (ones)

$\bar{F}(A, B, C, D) = \bar{A}\bar{B} + \bar{B}\bar{D}$

$F(A, B, C, D) = (A + \bar{B})(B + D)$

# Buffer

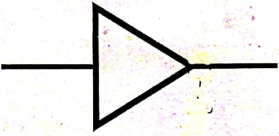
بجهد غير امكن ان

😊 (بار) لا (بار)

▪ A **buffer** is a gate with the function  $F = X$ :

كأنه سلك لا يتغير في قيمة الـ (inputs) : داعاً

inputs = outputs.



$$F \equiv X \quad \bar{X} \bar{X} = X$$

X	F
0	0
1	1

بطينا زي الحلال ( المدخل زي المخرج )

▪ In terms of Boolean function, a buffer is the same as a connection! (wire) <sup>سلك</sup> <sub>عاضى لا يتغير</sub>

▪ So why use it? <sup>قيمة الـ</sup> <sub>(inputs)</sub>

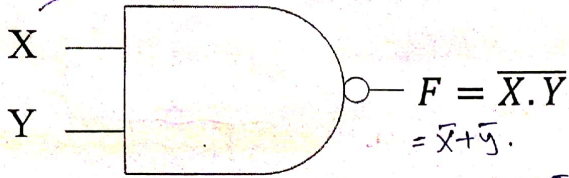
- A buffer is an **electronic amplifier** used to **improve circuit voltage levels** and **increase the speed of circuit operation** (signal) <sup>صوتى</sup>
- **Protection and isolation** between circuits

# NAND Gate

\* يتم تنفيذ ال (and) بولاجم ال (not) ←

- The **NAND** gate has the following symbol and truth table:

AND Not  
Not AND

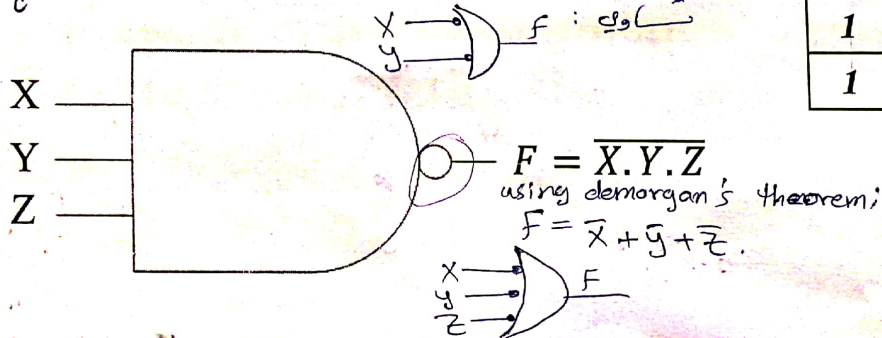


X	Y	F
0	0	1
0	1	1
1	0	1
1	1	0

AND ال ال ال

$\overline{XY} \neq \overline{X} \overline{Y}$   
"  
 $\overline{X} + \overline{Y}$

تنبه لفتح ال (AND)



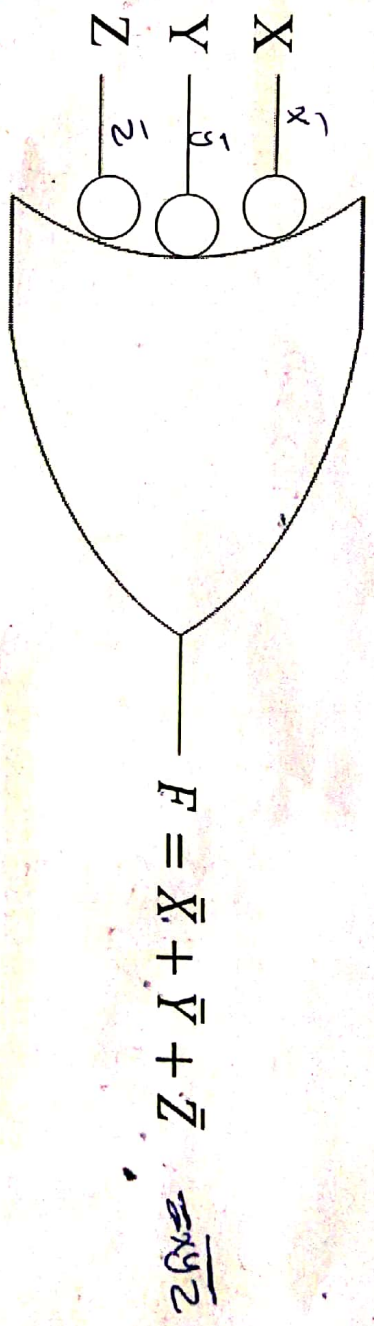
- NAND** represents **NOT-AND**, i.e., the AND function with a NOT applied. The symbol shown is an **AND-Invert**. The **small circle** ("bubble") represents the **invert function**



# NAND Gates (continued)

- Applying DeMorgan's Law gives Invert-OR (NAND)

بعض الـ OR



- This NAND symbol is called Invert-OR, since inputs are inverted and then ORed together
- AND-Invert and Invert-OR both represent the NAND gate. Having both makes visualization of circuit function easier



# NAND Gates (continued)

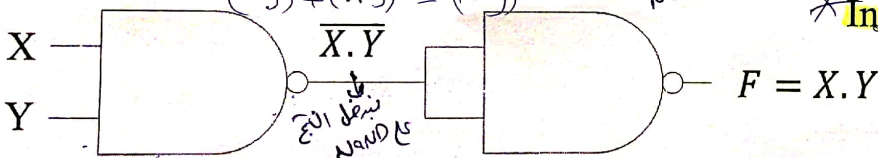
- Universal gate:** a gate type that can implement any Boolean function. **The NAND gate is a universal gate:**

$$F = (\overline{X \cdot Y}) \cdot (\overline{X \cdot Y})$$

Demorgan's law:

$$(\overline{X \cdot Y}) + (\overline{X \cdot Y}) =$$

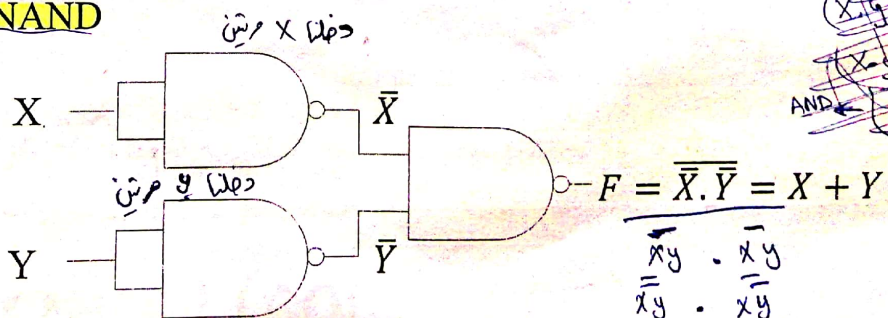
$$(X \cdot Y) + (\overline{X \cdot Y}) = (X \cdot Y)$$



**AND using NAND**

$$X \cdot Y = \overline{\overline{X \cdot Y}}$$

$$X \cdot X = \overline{\overline{X \cdot X}}$$



**OR using NAND**

$$\overline{\overline{X \cdot Y}} \cdot \overline{\overline{X \cdot Y}}$$

$$\overline{\overline{X \cdot Y}} \cdot \overline{\overline{X \cdot Y}}$$

$$X \cdot Y + X \cdot Y = X \cdot Y$$

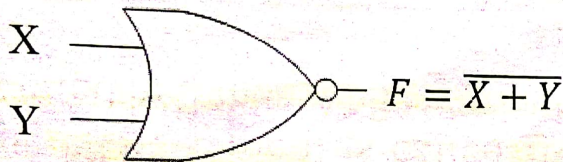
Chapter 2 - Part 3

# NOR Gate

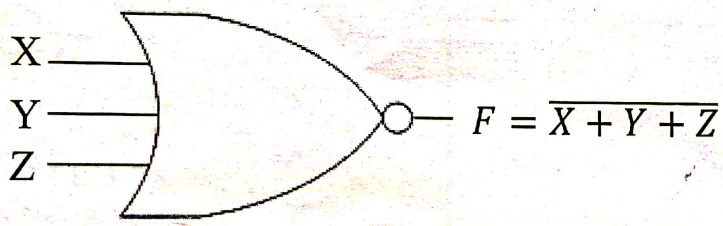
~~$(\bar{A} + \bar{B}) = \overline{A \cdot B}$~~   
 ~~$(\bar{A} + \bar{B}) = \overline{A \cdot B}$~~

- The NOR gate has the following symbol and truth table:

تنبیه الی: ع



X	Y	F
0	0	1
0	1	0
1	0	0
1	1	0

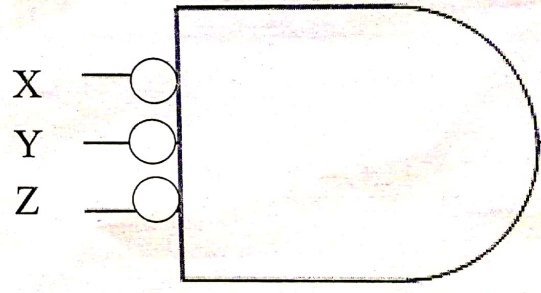
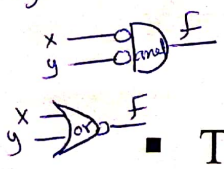


- NOR** represents **NOT-OR**, i.e., the **OR** function with a **NOT** applied. The symbol shown is an **OR-Invert**. The small circle (“bubble”) represents the invert function

# NOR Gates (continued)

- Applying DeMorgan's Law gives **Invert-AND** (NOR)

\* عندنا تغير مكان ال inverter من ال input الى ال output .  
نعكس نوع ال gate .



$$F = \overline{X+Y+Z} \rightarrow \bar{X} \cdot \bar{Y} \cdot \bar{Z}$$

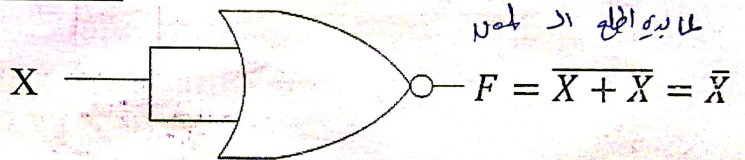
$$F = \bar{X} \cdot \bar{Y} \cdot \bar{Z}$$

- This **NOR** symbol is called **Invert-AND**, since inputs are inverted and then ANDed together
- OR-Invert** and **Invert-AND** both represent the NOR gate. Having both makes visualization of circuit function easier

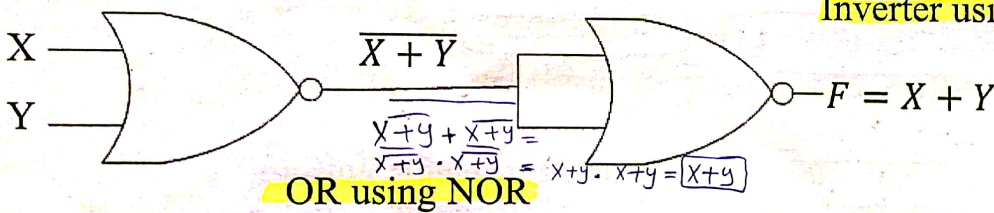
لتنو

# NOR Gates (continued)

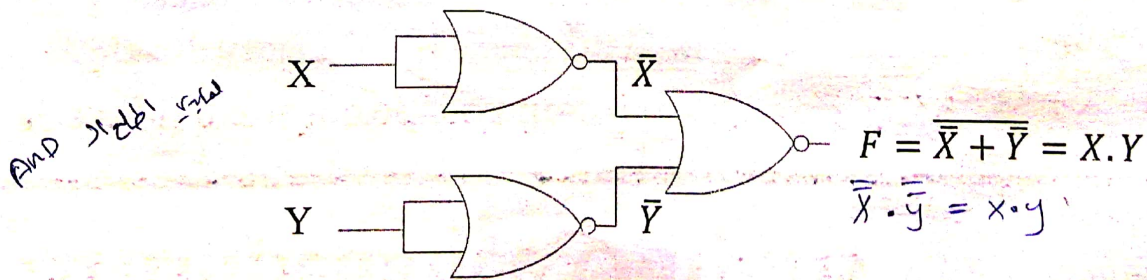
- The NOR gate is a universal gate:



Inverter using NOR



اذا صليت (....) ف = 2 و طلبنا نرسم بالبيانات المنطقية او دوائى بديل لها والنتيجة على map K و نطلع ال function المنطقى و نعينه بخرج نقيده ببوليان منطقية و نرى ال output منه مخرجات ال بوابه المنطقية .



(NAND) بال (OR) نفس

# Hi-Impedance Outputs

(buffer) مرنية ال

التي مرنية ال (Primitives) quite

Logic gates introduced thus far

- have 1 and 0 output values, Values *بديهي*
- cannot have their outputs connected together, and *the circuit is damaged*
- transmit signals on connections in only one direction

(2-output) لنينا \*

ممكنين بديهي



not(0) and not(1) then is a third new value.

Three-state logic adds a third logic value, **Hi-Impedance (Hi-Z)**, giving three states: 0, 1, and Hi-Z on the outputs.

Primary gates:

- AND
- OR
- NOT

(1, 0, Z) outputs لا كاي ال (1, 0, Z)

Hi-Z can be also denoted as **Z or z**

The presence of a Hi-Z state makes a gate output as described above behave quite differently:

- "1 and 0" become "1, 0, and Hi-Z"
- "cannot" becomes "can," and
- "only one" becomes "two"

\*Two-state buffers-



\*Tri-state buffer (3 state buffer)



Chapter 2 - Part 3 12 (EN = Switch)

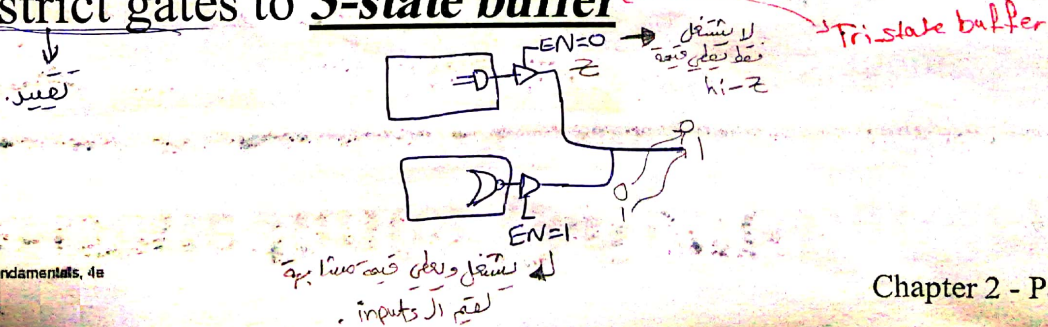
EN	IN	F
0	0	0
0	1	1
1	0	Z
1	1	Z

# Hi-Impedance Outputs (continued)

## What is a Hi-Z value?

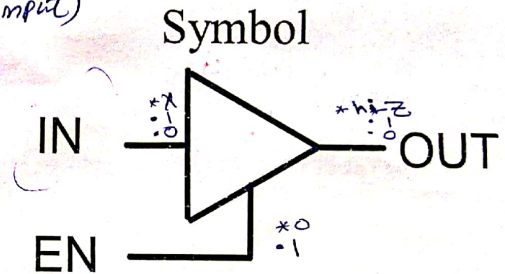
- The **Hi-Z value** behaves as an open circuit
- This means that, looking back into the circuit, the output appears to be disconnected
- It is as if a switch between the internal circuitry and the output has been opened

## Hi-Z may appear on the output of any gate, but we restrict gates to **3-state buffer**



# Tri-State Buffer (3-State Buffer)

- For the **symbol** and truth table, **IN** (input) is the **data input**, and **EN** is the **control input** (Enable)



- For **EN = 0**, regardless of the value on IN (denoted by X), the output value is Hi-Z
- For **EN = 1**, the output value follows the input value

Truth Table

EN	IN	OUT
0	X	Hi-Z
1	0	0
1	1	1

عندما يكون الـ EN ← 0 فإن الـ output يكون (Hi-Z) لا يوجد إشارة في اذرع البوابة.

عندما يكون الـ EN ← 1 فإن الـ output يكون (input) تابع الـ function.

لا يوجد إشارة في اذرع البوابة.

لا يوجد إشارة في اذرع البوابة.

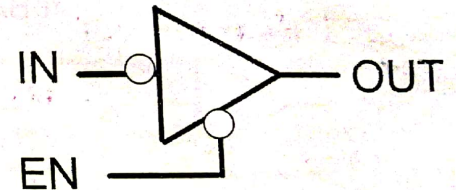
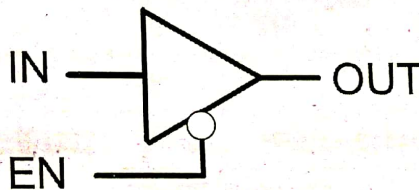
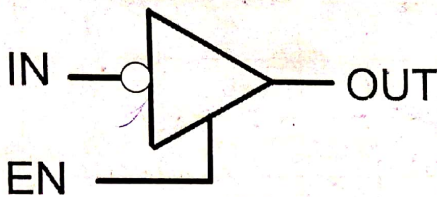
# Tri-State Buffer Variations

By adding "bubbles" to signals:

- Data input, IN, can be inverted (by bubbles)
- Control input, EN, can be inverted (by bubble).

إذا كان الـ EN = 0 الجواب Hi-Z فقط  
النظر عن المدخل الـ IN

input bubble



EN	IN	OUT
0	X	Hi-Z
1	0	1
1	1	0

دائماً  
نظرة عن  
الـ input

EN	IN	OUT
0	0	0
0	1	1
1	X	Hi-Z

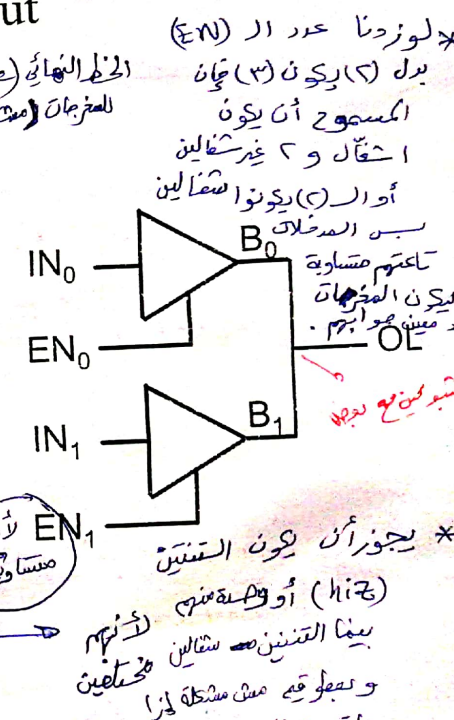
EN	IN	OUT
0	0	1
0	1	0
1	X	Hi-Z



# Resolving 3-State Values on a Connection

- Connection of two tri-state buffer outputs,  $B_1$  and  $B_0$ , to a wire, OL (Output Line)  $\rightarrow$  Multiplexed Output

$EN_1$	$EN_0$	$IN_1$	$IN_0$	$B_1$	$B_0$	OL
0	0	X	X	Hi-Z	Hi-Z	Hi-Z
0	1	X	0	Hi-Z	0	0
0	1	X	1	Hi-Z	1	1
1	0	0	X	0	Hi-Z	0
1	0	1	X	1	Hi-Z	1
1	1	0	0	0	0	0
1	1	0	1	0	0	0
1	1	1	0	1	1	1
1	1	1	1	1	1	1



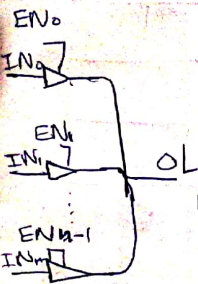
The two enables are (on)

circuit ال circuit ال circuit ال circuit ال circuit ال circuit ال circuit ال circuit ال circuit ال circuit ال

# Resolving 3-State Values on a Connection

▪ **Resulting Rule:** At least one buffer output value must be Hi-Z. Why? \* At least one tri-state buffer must be inactive (hi-Z)

- Because any data combinations including (0,1) and (1,0) can occur. If one of these combinations occurs, and no buffers are Hi-Z, then high currents can occur, destroying or damaging the circuit



▪ **How many valid buffer output combinations exist?** (Fire)

- 5 valid output combination

number of legal outputs = 5. The maximum number of outputs is  $2^n + 1$ .  
 Example:  $n=2$  →  $2^2 + 1 = 5$

▪ **What is the rule for "n" tri-state buffers connected to wire, OL?**

- \* At least "n-1" buffer outputs must be Hi-Z. \* Ex: n=2 → n=4  
 (1) must be hi-Z (3) must be hi-Z  
 (1) at most buffer must be active
- **How many valid buffer output combinations exist?**  
 Each of the n-buffers can have a 0 or 1 output with all others at Hi-Z. Also all buffers can be Hi-Z. So there are  $2^n + 1$  valid combinations.

\* at most 1 tri-state buffer must be active  
 slide

State buffer 2, 5

# Tri-State Logic Circuit

بشكل بديهي لكي تتقلل مساحة الدارة

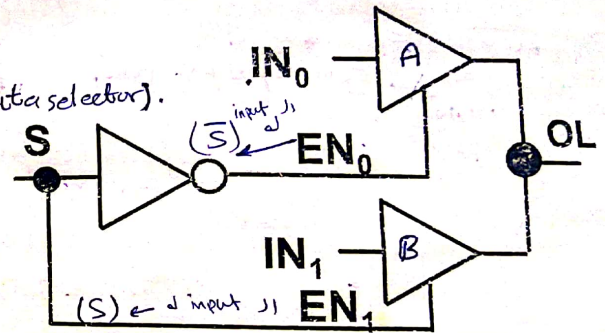
- **Data Selection Function:** If  $s = 0$ ,  $OL = IN_0$ , else  $OL = IN_1$
- Performing data selection with tri-state buffers:

نفرجهما لتأخذين A, B

3 input

S	EN <sub>1</sub>	EN <sub>0</sub>	IN <sub>1</sub>	IN <sub>0</sub>	OL
0	0	1	0	0	0
0	0	1	X	1	1
1	1	0	0	X	0
1	1	0	1	X	1

(data selector).



عنا في وحدة من (EN) ال input له (S) والباقي (EN) input له (S) إذا لم يكونوا متساويين دائماً (0,0) و (1,0) بالباقي واحد

- Since  $EN_0 = \bar{s}$  and  $EN_1 = s$ , one of the two buffer outputs is always Hi-Z.

# Logic Functions using Tri-State Buffers

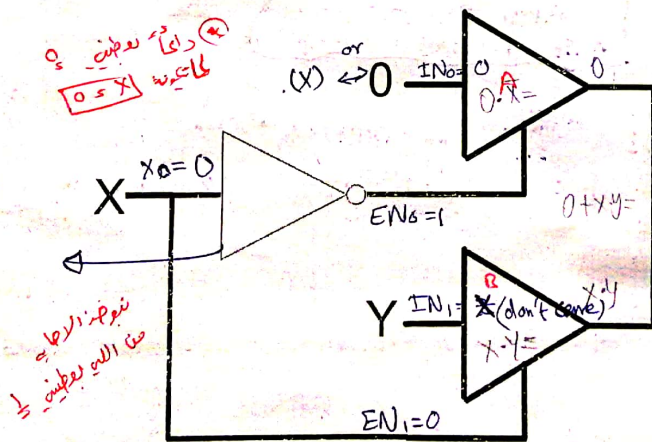
- Implement AND gate using 3-State buffers and inverters *Tri state buffers + inverters to implement any gate we want.*

$$F(X, Y) = X \cdot Y$$

- Use X as control input: *(2-input AND gate) \* this implementation is \**
  - When  $X = 0$ ,  $F = 0$  regardless of the value of Y
  - When  $X = 1$ ,  $F = Y$  *or  $F = X$  (dis. cpn. En = 0) = 0*

X	Y	F
0	0	0
0	1	0
1	0	0
1	1	1

*F = constant. F = 0, A f = X F = Y B F ≠ constant*



\* الخواص لتقبل أي gate يستخدم  
 - inverters + Tri-state buffer  
 ① نوسم ال (Truth table) ال (function) المراد  
 ② نوسم ال (truth table) من اليف بناء  
 على ال most significant و ال (X) Variable  
 ③ بعد ما نوسم الجول نساو نطلع ال function جولة  
 Constant  $X$  أو  $Y$   
 0 أو 1

\* نلاحظ: دائماً أن ال outputs مشبوكون معاً في سلك واحد بالأحرى  
 ① نجزز ببول ال bubble على  
 ال buffer متبوعه  
 أي لا يلزم كتابه إشارة ال inverter ال bubble ال  
 ال bubble ال

Shaym  
Khaled  
Mawgadah

# Logic Functions using Tri-State Buffers

- Implement the following function using 3-State buffers and inverters:  $F(w, x, y) = \bar{w}x + w\bar{y} + xy$

function (0) يكون لا function (1) يصبح (غير 0) و يصبح (غير 0) function (0) يكون لا function (1) يصبح (غير 0) و يصبح (غير 0)

Use  $w$  as control input:

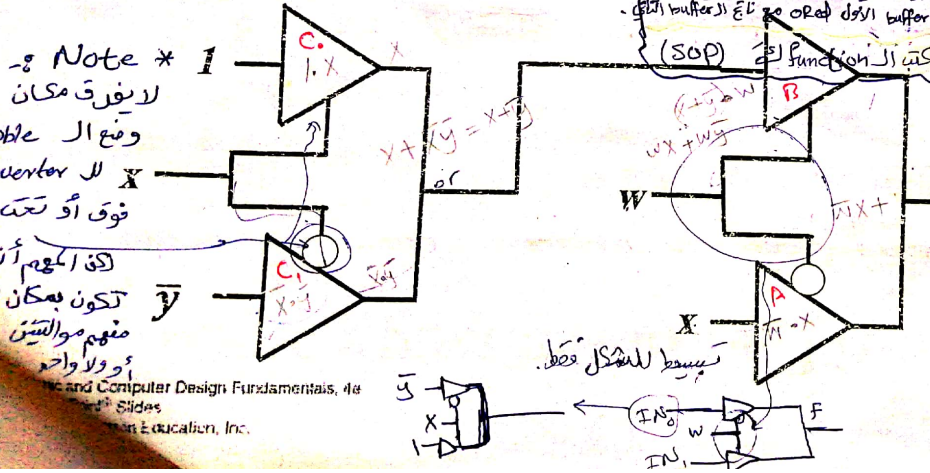
- When  $w = 0$ ,  $F = x$  regardless of the value of  $Y$
- When  $w = 1$ 
  - If  $x = 0$ ,  $F = \bar{y}$
  - If  $x = 1$ ,  $F = 1$

\* Note \*  
X: Variable (x)  
X: don't care value

w	x	y	F
0	0	0	0
0	0	1	0
0	1	0	1
0	1	1	1
1	0	0	1
1	0	1	0
1	1	0	1
1	1	1	1

F=x  
صنعتنا الجرد  
بالرغم بناء على (w)  
f=y  
F=x/F=w  
F=1 (const)

\* إذا أعطيت بالسؤال الرسم وطول كتابة  
ال function لمثل لها: الخطوات:  
1 عن كل مثل buffer نكتب ال input  
2 ضواء كانوا true أو inverted  
3 ثم بين ضلتي ال buffer نقل ال  
ال buffer ال or ال buffer ال  
ثم نكتب ال function ال (SOP)



Note \*  
لا يفرق مكان  
وضع ال  
inverter ال  
فوق أو تحت  
لكن المهم أن  
تكون مكان ال  
منهم موالتين  
أو ولا واحد

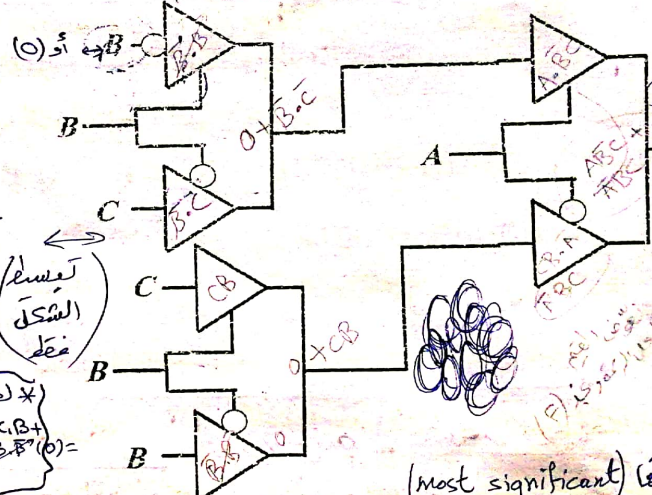
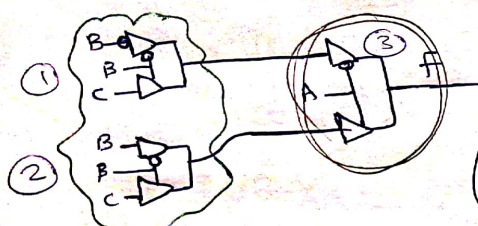
صنعتنا الجرد  
مرة ثانية بالنسبة  
بناء  
على (x)  
الستة لا يكون  
the next most  
significant  
variable.

# Logic Functions using Tri-State Buffers

\* هذا أكيد جاي عليه سؤال

- Write the Boolean expression of  $F(A, B, C)$  given the diagram below:  $F(A, B, C) = ABC + \bar{A}BC$

\* مقطع الرسم -



نرسم الـ (truth table) وذلك  
 n = 2^n : عدد المتغيرات  
 3 = 2^3 = 8 = 2^3 = 8

A	B	C	F
0	0	0	0
0	0	1	0
0	1	0	0
0	1	1	1
1	0	0	0
1	0	1	1
1	1	0	0
1	1	1	0

implementation function  
 1.  $\bar{B} \cdot B \rightarrow \bar{B} \cdot B + B \cdot C = \bar{B} \cdot C$   
 2.  $C \cdot B \rightarrow C \cdot B + B \cdot \bar{B} = C \cdot B$   
 3.  $A \cdot B \cdot C + A \cdot \bar{B} \cdot C = A \cdot C$

نقسم الجداول بالنسبة لـ (A) لأنها (most significant variable) ولأنها (most significant variable) ولأنها (most significant variable)

$$F(A, B, C) = ABC + \bar{A}BC$$

نبدأ الرسم من الـ output (F) ثم نضع الـ (ENable) (A) ونضع الـ (ENable) (B) ونضع الـ (ENable) (C) ونضع الـ (ENable) (A) ونضع الـ (ENable) (B) ونضع الـ (ENable) (C)

The complex gates

## Exclusive OR/ Exclusive NOR

- The **eXclusive OR (XOR)** function is an important Boolean function used extensively in logic circuits
- The XOR function may be:
  - implemented directly as an electronic circuit (truly a gate) or
  - implemented by interconnecting other gate types (used as a convenient representation)
- The **eXclusive NOR (XNOR)** function is the complement of the XOR function
- By our definition, **XOR and XNOR gates are complex gates**

# Exclusive OR/ Exclusive NOR

- Uses for the XOR and XNORs gate include:
  - Adders/subtractors/multipliers
  - Counters/incrementers/decrementers
  - Parity generators/checkers

XOR (A)  $\oplus$   $\bar{A}B + A\bar{B}$   
 و  $\bar{A}B + A\bar{B}$  (B)  $\oplus$

## Definitions

- The XOR function is:  $X \oplus Y = \bar{X}Y + X\bar{Y}$
- The XNOR function is:  $X \odot Y = \bar{X} \oplus \bar{Y} = XY + \bar{X}\bar{Y}$

XNOR (A)  $\odot$   $XY + \bar{X}\bar{Y}$   
 و  $XY + \bar{X}\bar{Y}$  (B)  $\odot$

- Strictly speaking, XOR and XNOR gates do not exist for more than two inputs. Instead, they are replaced by odd and even functions



# Proof: XNOR is the complement of XOR

- $$\overline{X \oplus Y} = \overline{\overline{XY} \oplus X\overline{Y}}$$

\* البطلان (نفي) (OR)   
 د. د. د. (demorgan's law)
- $$\overline{X \oplus Y} = \overline{\overline{XY} \cdot \overline{X\overline{Y}}}$$

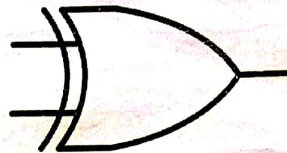
(primitive gates) و (complex gates)
- $$\overline{X \oplus Y} = (X + \overline{Y}) \cdot (\overline{X} + Y)$$

distributive law.
- $$\overline{X \oplus Y} = \overline{X \cdot \overline{X}} + XY + \overline{X\overline{Y}} + \overline{Y \cdot \overline{Y}}$$

(0) (0)
- $$X \odot Y = \overline{X \oplus Y} = XY + \overline{X\overline{Y}}$$

# Symbols For XOR and XNOR

- XOR symbol:



- XNOR symbol:



لا تسمح أكثر من 2 مدخل

- Shaped symbols exist only for two inputs

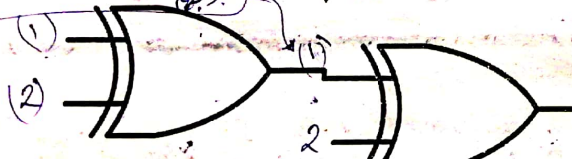
\* مهمة جداً : فقط

لا يسمح لهم بـ (2-1N)

wrong!  
3-inputs.



$(1) = 1 \oplus 2 \rightarrow$  ناتج بوابة



Right!  
2-inputs only

\* إذا أردنا زيادة عدد الـ (inputs) عن (two) نقوم بإضافة (gates) أخرى.

Chapter 2 - Part 3  
كله قبل (XOR) 25

# Truth Tables for XOR/XNOR

X	Y	$X \oplus Y$
0	0	0
0	1	1
1	0	1
1	1	0

X	Y	$X \odot Y (X \equiv Y)$
0	0	1
0	1	0
1	0	0
1	1	1

معادلة الـ (1)  
 $(\bar{x}y + x\bar{y})$   
 Function

دليل أن  
 complement  
 to each  
 other

معادلة الـ (1)  
 $(\bar{x}y + x\bar{y})$

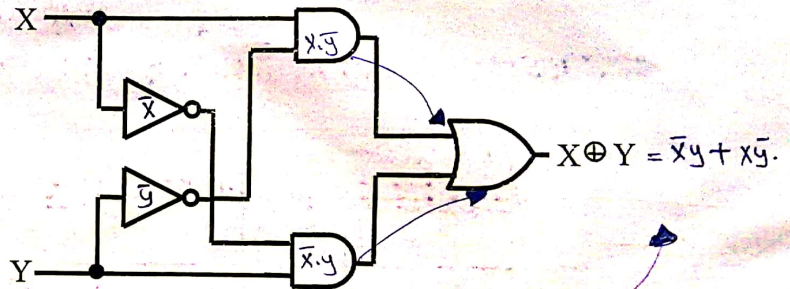
\* يجب أن يكونوا (x, y) متساويين ليكون الناتج (1) \* يجب أن يكونوا (x, y) مختلفين ليكون ناتج الـ (1) Function

- The XOR function means: **X OR Y, but NOT BOTH**
- Why is the XNOR function also known as the **equivalence** function, denoted by the operator  $\equiv$ ?
  - Because the function equals 1 if and only if **X = Y**

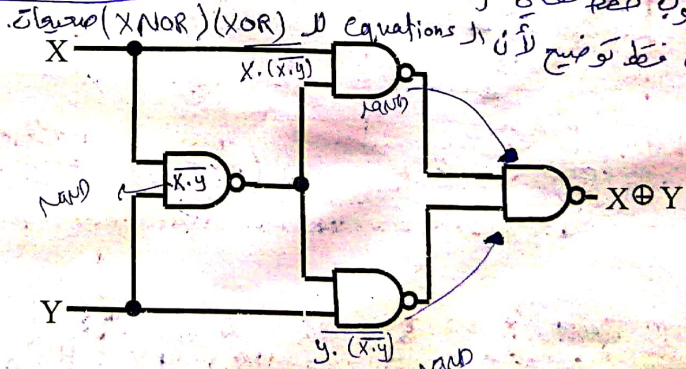
(XNOR Gate)  
 لأنها تتأكد من  
 أن (x, y) قيمتين  
 متساويتين.

# XOR Implementations

- The simple SOP implementation uses the following structure:



- A NAND only implementation is:



\* غير مطلوب فقط هاهي الرساله  
ولكن هي فقط توضيح لأن ال Equations (XOR) (XNOR) صحيات.

\* نريد الكهول على (XOR) gate  
فقط كالتالي -  
تستخدم (NAND) gates



# XNOR

لا ليس شرط فقط هذه ال (identities) حيث لو عوضنا القيم (0 و 1) بالامتحان نستطيع ان نجد الجواب بسهولة.

## The XNOR identities:

$\textcircled{1} X \odot 0 = \bar{X}$	$\textcircled{2} X \odot 1 = X$
$\textcircled{3} X \odot X = 1$	$\textcircled{4} X \odot \bar{X} = 0$
$X \odot Y = Y \odot X \quad (\text{خاصية تبيلية})$	
$X \odot Y \odot Z = (X \oplus Y) \odot Z = X \odot (Y \oplus Z) \quad (\text{خاصية تجميعية})$	

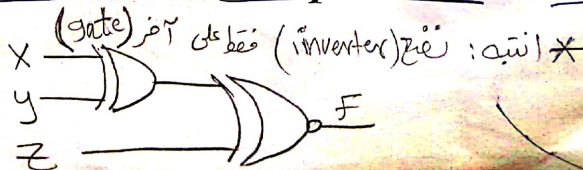
①  $1 \odot 0 = 0$  متساويين  
 $0 \odot 0 = 1$  متساويين  
 ②  $1 \odot 1 = 1$  متساويين  
 $0 \odot 1 = 0$  متساويين  
 ③  $0 \odot 0 = 1$  دائماً متساويين لإدخال الناتج دائماً (1).  
 $1 \odot 1 = 1$  دائماً متساويين لإدخال الناتج دائماً (1).  
 ④  $1 \odot 0 = 0$  دائماً متساويين لإدخال الناتج دائماً (0).  
 $0 \odot 1 = 0$  دائماً متساويين لإدخال الناتج دائماً (0).

- The XNOR function can be extended to 3 or more variables. For more than 2 variables, it is called an **even function**, not an XNOR:  $\ast$  إذا كان عدد ال (ones) في ال (input) زوجي يكون ال (output) 1 (XNOR) و ال (output) 0 (XOR) function

$$X \odot Y \odot Z = \bar{X}YZ + X\bar{Y}Z + XY\bar{Z} + \underline{XYZ} \quad (\text{Even \# of 1's})$$

- The even function is the complement of the odd function**

Reason: Number of ones = even number of ones.



# Odd and Even Functions

- The 1s of an **odd function** <sup>(XOR)</sup> correspond to minterms having an index with an odd number of 1s.

	y			
	0	1	3	2
x	4	5	7	6
	1	0	1	0
	z			

(K-maps)

نبدأ (0)

	C			
	0	1	3	2
	4	5	7	6
A	12	13	15	14
	8	9	11	10
	D			

\* شكلهم  
مميز  
مربع (1) و  
مربع جنبه (0)  
وزلاظ  
عزم قدرتنا  
على جمع ال (ones)  
أبداً

- The 1s of an **even function** <sup>(XNOR)</sup> correspond to minterms having an index with an even number of 1s.

	y			
	0	1	3	2
x	4	5	7	6
	1	0	1	0
	z			

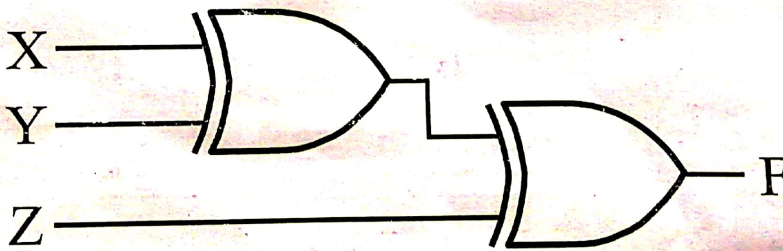
نبدأ (1)

	C			
	0	1	3	2
	4	5	7	6
A	12	13	15	14
	8	9	11	10
	D			

ولكن من الشكل  
نعرف أنه  
even/odd  
function

## Example: Odd Function Implementation

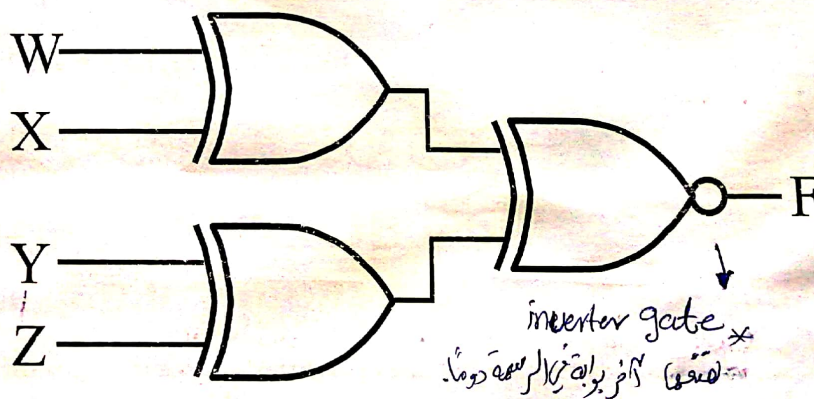
- Design a 3-input odd function  $F = X \oplus Y \oplus Z$  with 2-input XOR gates
- Factoring,  $F = (X \oplus Y) \oplus Z$
- The circuit:





## Example: Even Function Implementation

- Design 4-input even function  $F = \overline{W \oplus X \oplus Y \oplus Z}$  with 2-input XOR and XNOR gates
- Factoring,  $F = \overline{(W \oplus X) \oplus (Y \oplus Z)}$
- The circuit:



# Overview

## ▪ Part 1 – Design Procedure

### • Steps

- Specification
- Formulation
- Optimization
- Technology Mapping
- Verification

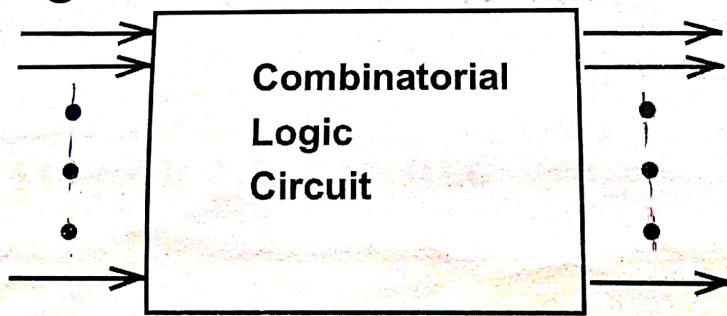
\* نريد استخدام التالي في تصميم دوائر كهربائية نعمل بوظيفة معينة.

Technology Mapping <sup>using:</sup> AND, OR, and NOT to NAND or NOR

# Combinational Circuits (doesn't have memory).

- A combinational logic circuit has:
  - A set of  $m$  Boolean inputs,
  - A set of  $n$  Boolean outputs, and
  - $n$  switching functions, each mapping the  $2^m$  input combinations to an output such that the current output depends only on the current input values

▪ A block diagram:



$m$  Boolean Inputs

$n$  Boolean Outputs

\* يكون هناك أكثر من  
input وبتلك أكثر  
من output ولكي  
كل output يعتمد على  
input الذي  
يخضع له

# Design Procedure

خطوات لكي نعمل من احتمال الخطأ في تصميم الدارة (circuit).

## 1. Specification (الوصف)

- Write a specification for the circuit if one is not already available. ***What does the circuit do? Including names or symbols for inputs and outputs***

## 2. Formulation (التمثيل)

- Derive a ***truth table*** or ***initial Boolean equations*** that define the required relationships between the inputs and outputs, if not in the specification

## 3. Optimization (التقليل) / (الاعتقار)

- Apply 2-level optimization using K-maps
- Draw a logic diagram for the resulting circuit using ANDs, ORs, and inverters (*basic-gates*)

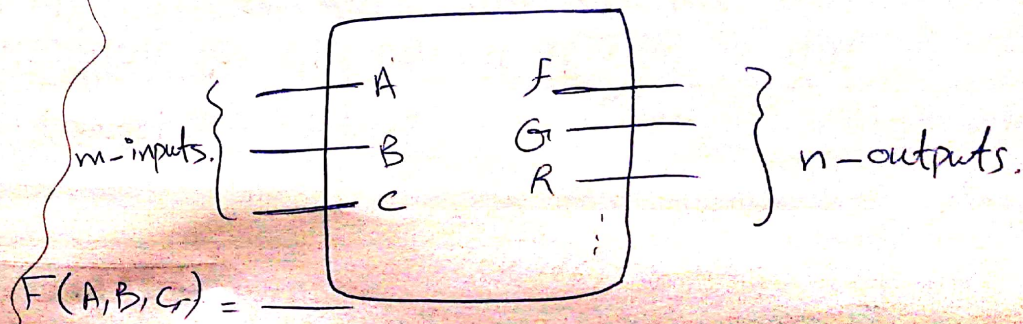
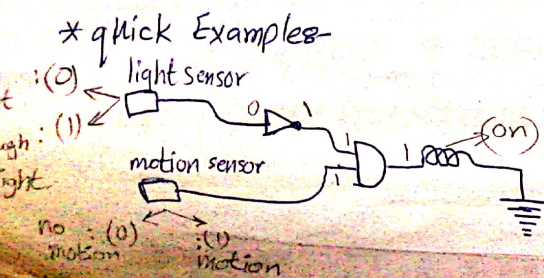
# Design Procedure

## 4. Technology Mapping (التعيين بالرسم)

- Map the logic diagram to the implementation technology selected

## 5. Verification (using lab logic).

- Verify the correctness of the final design manually or using simulation → (محاكاة)



# Design Example1

- Specification:** Design a combinational circuit that has 3 inputs ( $X, Y, Z$ ) and one output  $F$ , (such that  $F = 1$  when the number of 1's in the input is greater than the number of 0's (i.e. number of 1's  $\geq 2$ )).

الهدف  
 المجموعة التي  
 توصف مع  
 يعطين الfunction  
 (1) و (0).  
 \* شرط متى يكون ناتج (F) هو (1)  
 (المعظم)  
 الأغلبية، الأكثرية.

  - This is called majority function (i.e. majority of inputs must be 1 for the function to be 1)

- Formulation:**

نقل ال (truth table)

\* specification the table  
 تعيين الجدول  
 مكان ال (1)  
 عن طريق السطر  
 المعطى بالاستعانة.

	X	Y	Z	F
m <sub>0</sub>	0	0	0	0
m <sub>1</sub>	0	0	1	0
m <sub>2</sub>	0	1	0	0
m <sub>3</sub>	0	1	1	1
m <sub>4</sub>	1	0	0	0
m <sub>5</sub>	1	0	1	1
m <sub>6</sub>	1	1	0	1
m <sub>7</sub>	1	1	1	1

عدد الأصفار  
 أكبر من عدد الواحد  
 عدد الواحد  
 أكبر من عدد  
 الأصفار.

# Design Example 1 Cont.

(cost) ال دوائر

Optimization: *using (K-maps) دوائر ال function*

$$F(X, Y, Z) = XY + XZ + YZ$$

as a sum of minterms:  $\Sigma m(3, 5, 6, 7)$ .

	Y	
X	0	1
4		
	3	2
5	1	
7	1	
	6	1
	Z	

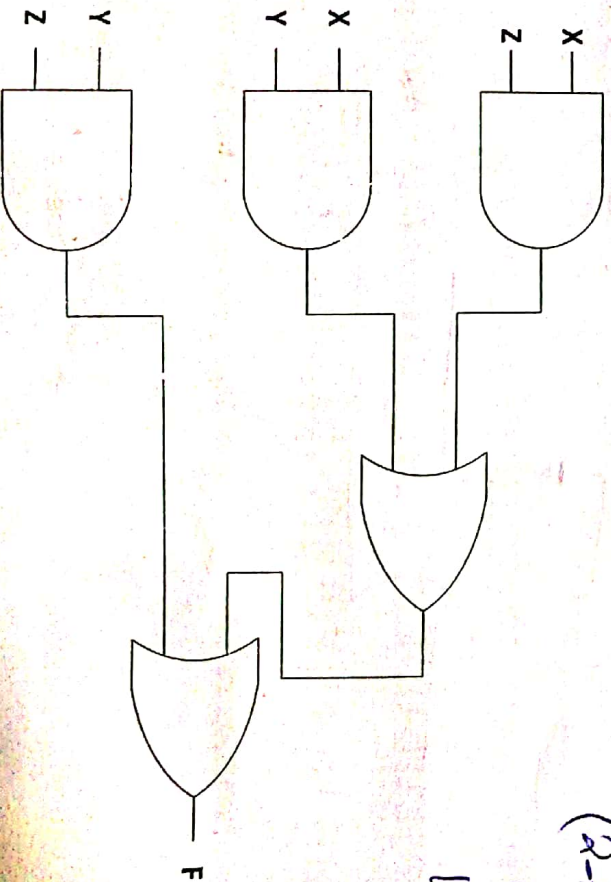
## Technology Mapping:

- Mapping with a library containing inverters, 2-input AND, 2-input

OR

\* *سلسلة نقطه (2-inputs)*

*للاصله نعمل اكواد  
برايه لكي نربطها  
باللح.*



# Design Example 2

- Specification:** Design a combinational circuit that compares 2-bit Binary number ( $A, B$ ) and produce two outputs ( $O_1, O_0$ ), such that:
  - $O_1 O_0 = 00$  When  $A = B$  and Both are even
  - $O_1 O_0 = 01$  When  $A < B$
  - $O_1 O_0 = 10$  When  $A > B$
  - $O_1 O_0 = 11$  When  $A = B$  and Both are odd

$O_1 O_0 = 00$	When $A = B$ and Both are even
$O_1 O_0 = 01$	When $A < B$
$O_1 O_0 = 10$	When $A > B$
$O_1 O_0 = 11$	When $A = B$ and Both are odd

- Formulation:** (truth table)
  - (decimal B) تعبئة الـ (A)
  - (decimal A) تعبئة الـ (B)

Note \* تعبئة الـ (A) و (B) كالتالي (4-variable inputs) يعني لا دخل لأن A ← متغير لها (2 bits) تعبئة الـ (B) متغيرين.

\* 2 binary numbers and each number consists of 2 bits.

$A(A_1 A_0)$	$A=B$	$B(B_1 B_0)$	$O(O_1 O_0)$
00	A=B	00	00
00	A<B	01	01
00		10	01
00		11	01
01	A>B	00	10
01	A=B	01	11
01		10	01
01		11	01
10		00	10
10		01	10
10		10	00
10		11	01
11		00	10
11		01	10
11		10	10
11		11	11

bit لكل bit output (Kmap) نقل الـ

odd decimal (1)

decimal odd (3)

(0) ← (Kmap) لـ

(1) ← (Kmap) لـ

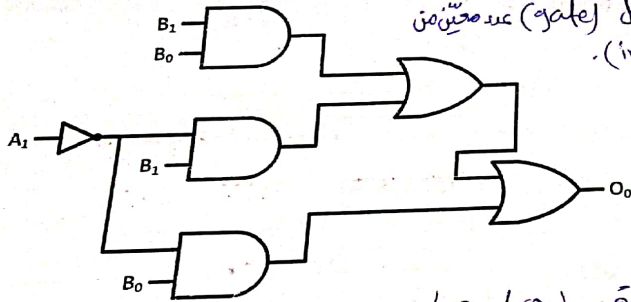


# Design Example2 Cont. (output) نقل (k-map) \* لكل

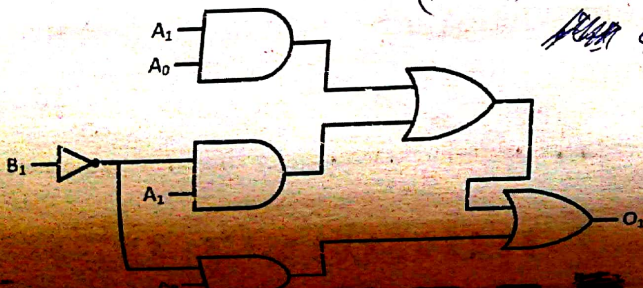
منفصلة . وجميعها (16 مربع) لأن  $2^4 = 16$  . 1: شجرات على اعتبار  $A_1, A_0, B_1, B_0$  (من 4 متغيرات)

## Optimization and Technology Mapping:

$O_0 = B_1 B_0 + \overline{A_1} B_1 + \overline{A_1} B_0$  لا تأخذ عدد لمبرمن ال (gates) لأن لكل (gate) عدد متغيرات ال (inputs)



$O_1 = A_1 A_0 + A_0 \overline{B_1} + A_1 \overline{B_1}$



\* يتم أخذ قيم  $(O_1/O_0)$  من الجدول ووضعها في ال (k-map) حسب الترتيب

$O_0$

	$B_1$			
$A_1$	0	1	3	2
$A_0$	4	5	7	6
	$B_0$			
$A_1$	8	9	11	10
$A_0$	12	13	15	14

*both essential prime implicants*

$O_1$

	$B_1$			
$A_1$	0	1	3	2
$A_0$	4	5	7	6
	$B_0$			
$A_1$	8	9	11	10
$A_0$	12	13	15	14

# Design Examples

## 1. Specification

- **BCD to Excess-3 code converter**. (Excess 3) ← (BCD) دائرة (circuit) لـ
- Transforms <sup>(4-bits)</sup> BCD code for the decimal digits to <sup>(4-bits)</sup> Excess-3 code for the decimal digits
- BCD code words for digits 0 through 9: 4-bit patterns 0000 to 1001, respectively BCD : (0 → 9)   
 بنية لفظة (9) كود (Excess) هي (2)
- Excess-3 code words for digits 0 through 9: 4-bit patterns consisting of 3 (binary 0011) added to each BCD code word
- **BCD input is labeled A, B, C, D**
- **Excess-3 output is labeled (W, X, Y, Z)**

# Design Examples Cont.

## 2. Formulation

ABCD	WXYZ
0000	0011
0001	0100
0010	0101
0011	0110
0100	0111
0101	1000
0110	1001
0111	1010
1000	1011
1001	1100
1010	XXXX
1011	XXXX
1100	XXXX
1101	XXXX
1110	XXXX
1111	XXXX

don't care.

\* لهم خطي بال كل صلا افعال  
 بال (illegal) (Outputs) من (10 ← 15) في  
 (BCD) موهوبين بال (9 ← 0) في

# Design Examples3 Cont.

## 3. Optimization

$$W = A + BC + BD$$

$$X = \bar{B}D + \bar{B}C + B\bar{C}\bar{D}$$

$$Y = \bar{C}\bar{D} + CD$$

$$Z = \bar{D}$$

	W			
	0	1	3	2
A	X	1	1	1
	4	5	7	6
	12	13	15	14
	8	9	11	10
	1	1	X	X
			D	B

	X			
	0	1	3	2
A	1	1	1	1
	4	5	7	6
	12	13	15	14
	8	9	11	10
	X	X	X	X
			D	B

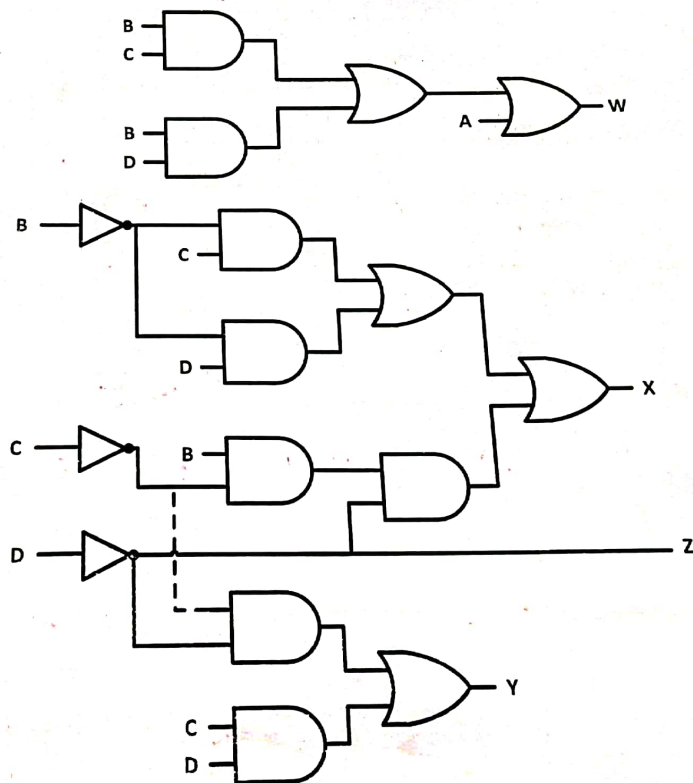
	Y			
	0	1	3	2
A	1	1	1	
	4	5	7	6
	12	13	15	14
	8	9	11	10
	1	X	X	X
			D	B

	Z			
	0	1	3	2
A	1	1	1	1
	4	5	7	6
	12	13	15	14
	8	9	11	10
	1	X	X	X
			D	B

# Design Example3 Cont.

## 4. Technology Mapping

- Mapping with a library containing inverters, 2-input AND, 2-input OR



# Homework: BCD to 7-Segment

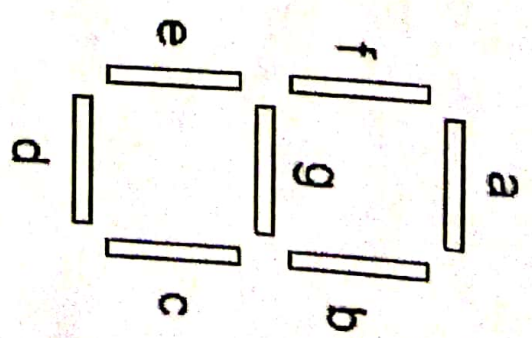
## Specification:

- Inputs: (A, B, C, D) BCD code from 0000-to-1001
- Outputs: (g, f, e, d, c, b, a)

## Formulation:

ABCD	gfedcba
0000	0111111
0001	0000110

Handwritten notes: "desired" with an arrow pointing to the first row, and "(1)" written below the second row.



## Optimization:

- How many K-maps?

1001	1100111
1010	0000000

Handwritten notes: "(BCD) لا يوجد (10) لا يوجد" and "(1)" written below the second row.

1111	0000000
------	---------

Handwritten notes: "(BCD) لا يوجد" and "(1)" written below the row.

Handwritten notes: "خارج 1: output (1) : يضيء الحرف", "بينما 0: output (0) : لا يضيء الحرف"

Handwritten notes: "المدخلات", "المخرجات", "خارج 1", "بينما 0", "K-map"

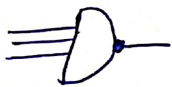
# Technology Mapping

## Mapping Procedures

- To NAND gates
- To NOR gates

\* better to make a circuit only with one gate type. (using one type universal gate).

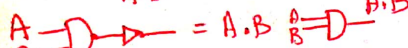
\* NAND Gate:-



\* NOT:-



\* AND:-



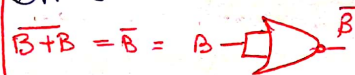
\* OR:-



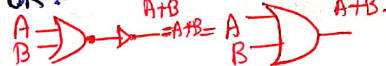
\* NOR Gate:-



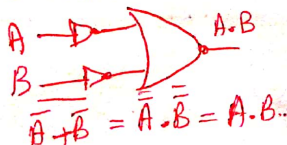
\* NOT:-



\* OR:-



\* AND:-



# Mapping to NAND gates

---

- Assumptions:

- Gate loading and delay are ignored
- Cell library contains an inverter and  $n$ -input NAND gates,  $n = 2, 3, \dots$
- An AND, OR, inverter schematic for the circuit is available

- The mapping is accomplished by:

- Replacing AND and OR symbols,  $\rightarrow$  to NAND
- Pushing inverters through circuit fan-out points, and

- Canceling inverter pairs  $\rightarrow$  to NAND.

Use 0's in 1's  $\leftarrow$  (2-inverters)  
 Use 1's in 0's  $\leftarrow$  (2-inverters)

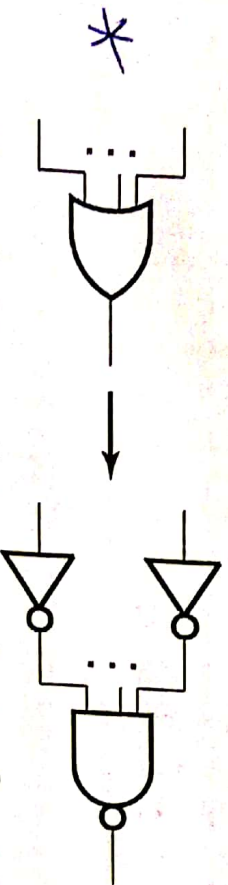
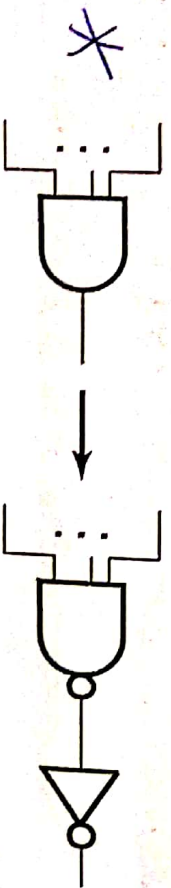




# NAND Mapping Algorithm

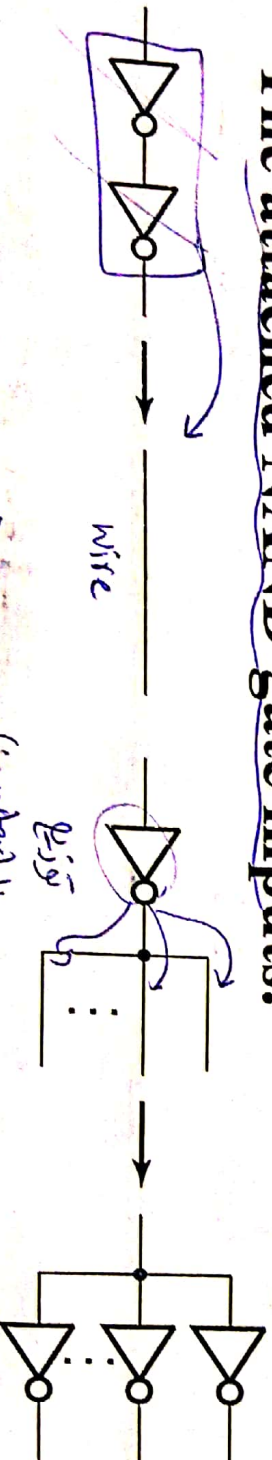
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1. Replace ANDs and ORs:

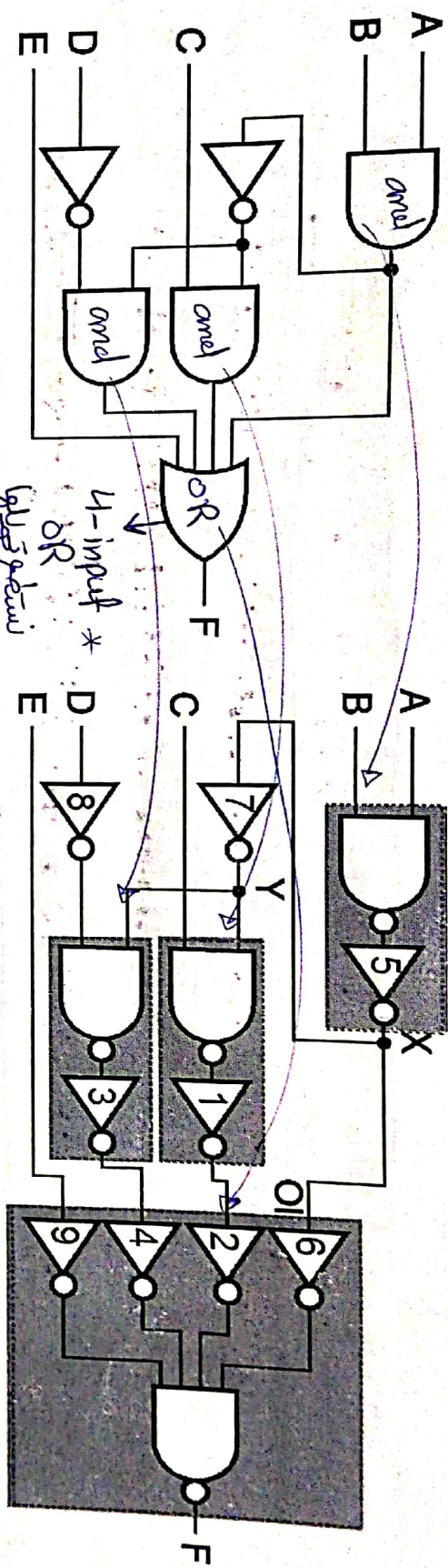


2. Repeat the following pair of actions until there is at most one inverter between :

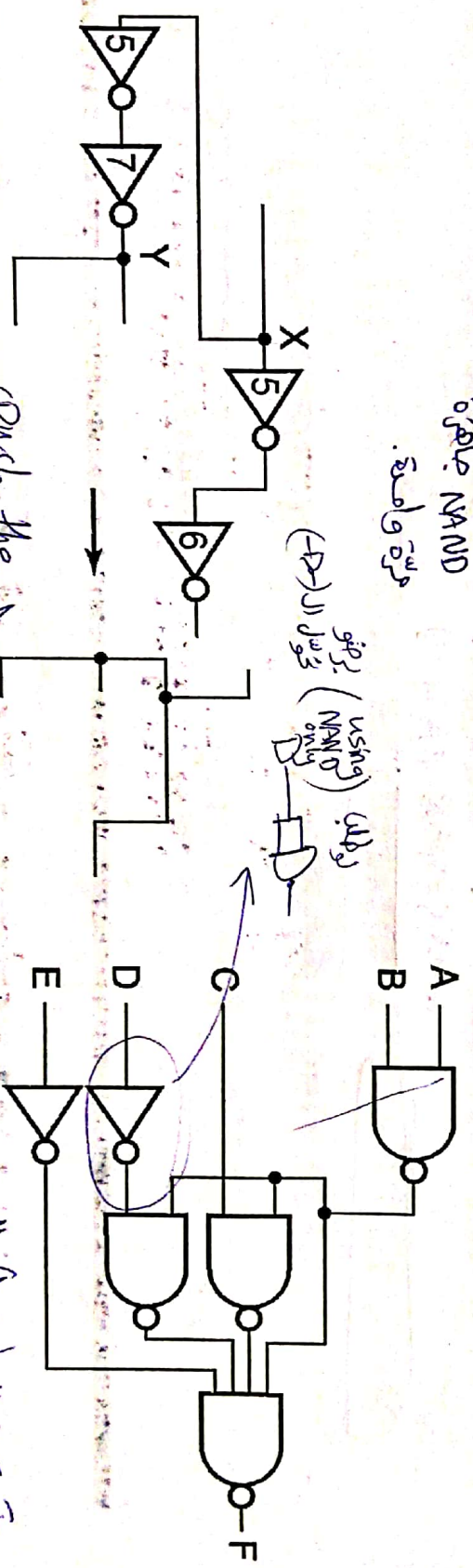
- A circuit input or driving NAND gate output, and
- The attached NAND gate inputs.



# NAND Mapping Example



(a)  
 4-input OR  
 4-input NAND  
 مبروكه و مبروكه  
 نستعمل قوتها



(b)  
 4-input NAND  
 مبروكه و مبروكه  
 نستعمل قوتها  
 Push the (c) inverter  
 \*to reduce the number of inverters.

# Mapping to NOR gates (نفس ال NAND)

## ▪ Assumptions:

- Gate loading and delay are ignored
- Cell library contains an inverter and  $n$ -input NOR gates,  $n = 2, 3, \dots$  (لا يوجد داعي لتحويل ال (inverter) فقط ال (and/or) gates)
- An AND, OR, inverter schematic for the circuit is available

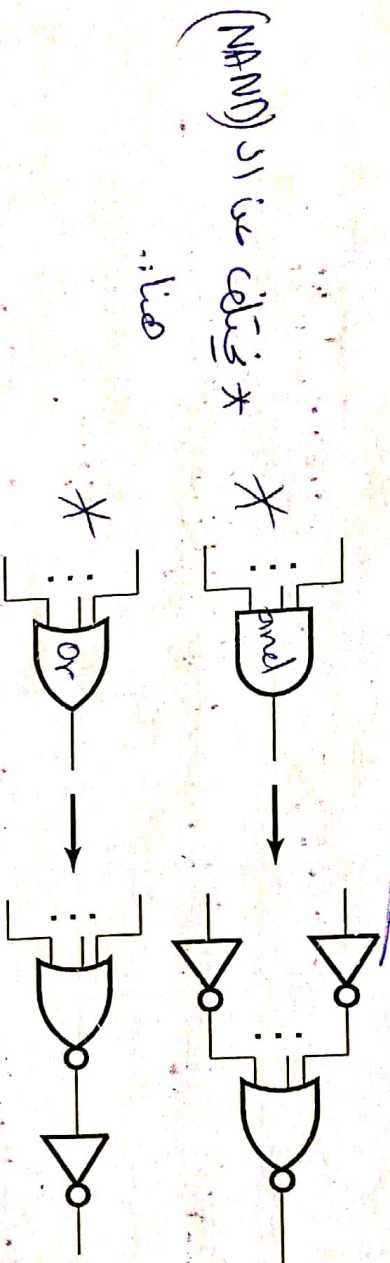
## ▪ The mapping is accomplished by:

- Replacing AND and OR symbols,
- Pushing inverters through circuit fan-out points,  
and
- Canceling inverter pairs

# NOR Mapping Algorithm

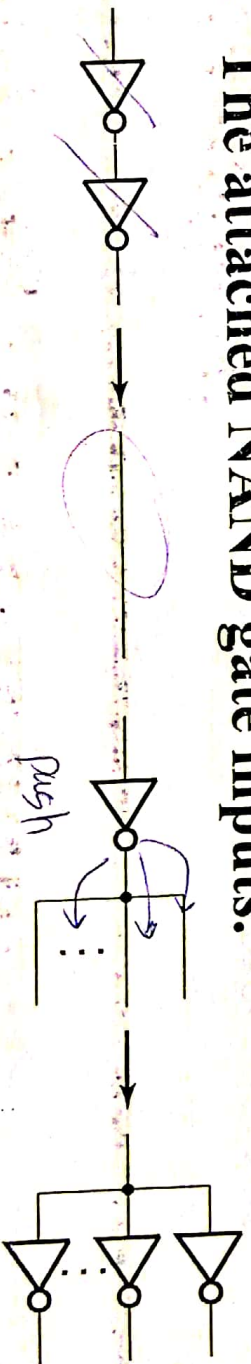
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## 1. Replace ANDs and ORs:

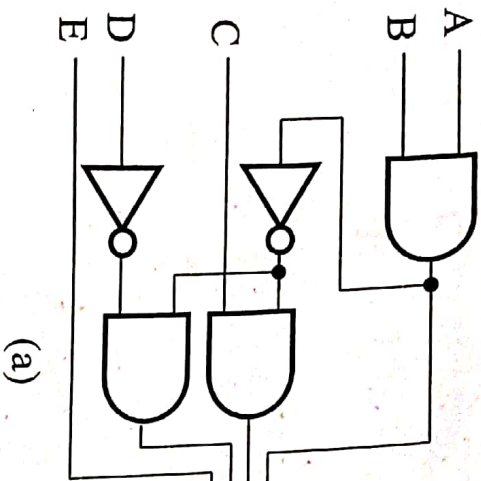


## 2. Repeat the following pair of actions until there is at most one inverter between :

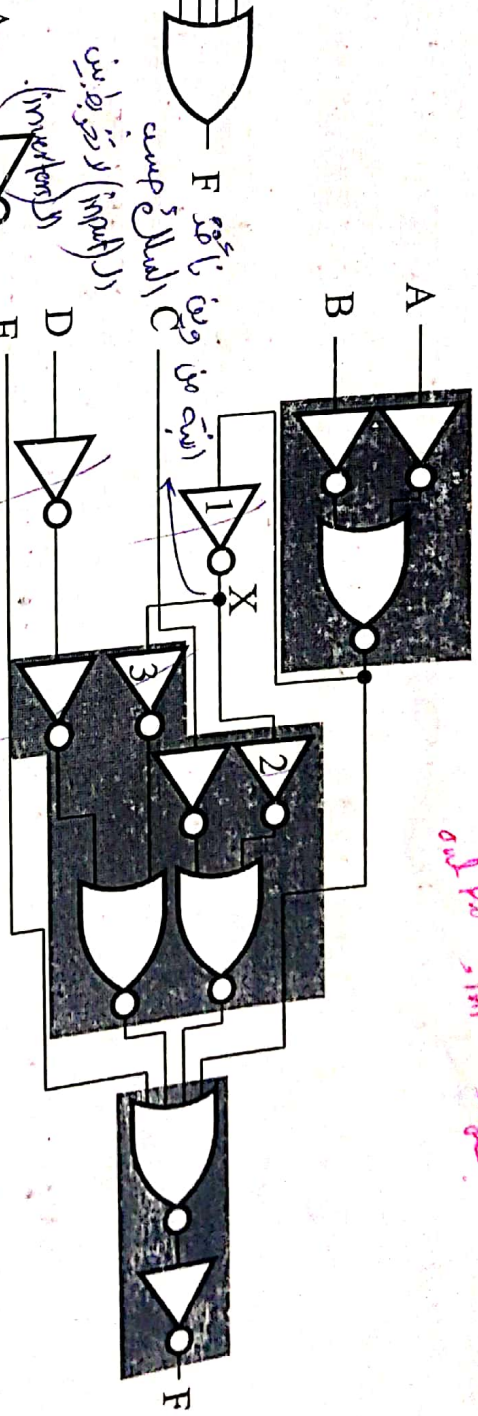
- A circuit input or driving NAND gate output, and
- The attached NAND gate inputs.



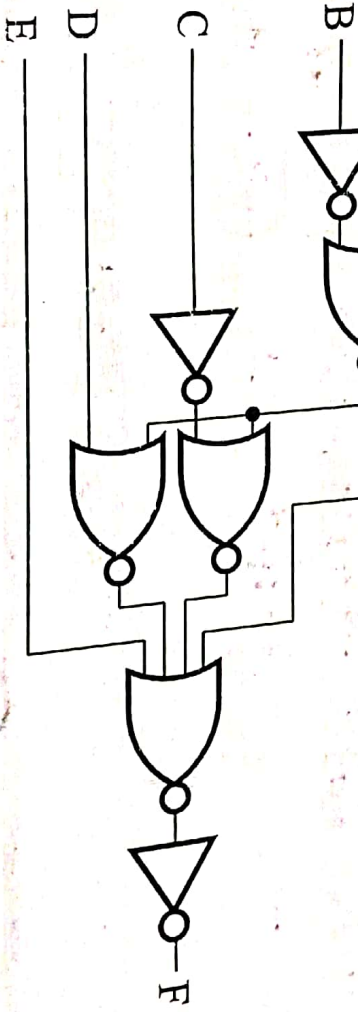
# NOR Mapping Example



(a)



(b)



(c)

مخرج الـ C (inverter) و الـ D و الـ E و الـ F

مخرج الـ F

and other input system

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# Overview

- **Part 2 – Combinational Logic** (without memory),  
input  $\xrightarrow{\text{gives}}$  output.
- Functions and functional blocks
- Rudimentary logic functions
- \* Decoding using Decoders (blocks).
  - Implementing Combinational Functions with Decoders
- \* Encoding using Encoders (blocks)
- Selecting using Multiplexers
  - Implementing Combinational Functions with Multiplexers

# Functions and Functional Blocks

---

- The functions considered are those found to be very useful in design
- Corresponding to each of the functions is a combinational circuit implementation called a **functional block**
- In the past, functional blocks were packaged as small-scale-integrated (SSI), medium-scale integrated (MSI), and large-scale-integrated (LSI) circuits
- Today, they are often simply implemented within a very-large-scale-integrated (VLSI) circuit



# Rudimentary Logic Functions

## Basic Functions :-

- Functions of a single variable  $X$
- Can be used on the inputs to functional blocks to implement other than the block's intended function

Functions of One Variable					
	Value fixing				transferring
$X$	$F = 0$	$F = 1$	$F = X$	$F = \bar{X}$	
0	0	1	0	1	1
1	0	1	1	1	0

**Value fixing:**  $a, b$

$a, b$  (output) و  $a, b$  (input) و  $a, b$  (output) و  $a, b$  (input)

$V_{DD} = 1$  (logic high)

لا تغير قيمة البتة و معيانية حسب المصفوف (الاحالة)

**Transferring:**  $c$  (same data)

the output is as same  $V_{CC}$  or  $V_{DD}$  as the input.

(logic high) (صحيح)

صحيح المصفوف (الاحالة)

**Inverting:**  $d$  (invert the data)

invert the data

$F = X$

**Enabling:** next slide

$F = 0$

$F = 0$

$F = \bar{X}$

(a)  $F = 0$ , (b)  $F = 0$

(c)  $F = X$

(d)  $F = \bar{X}$

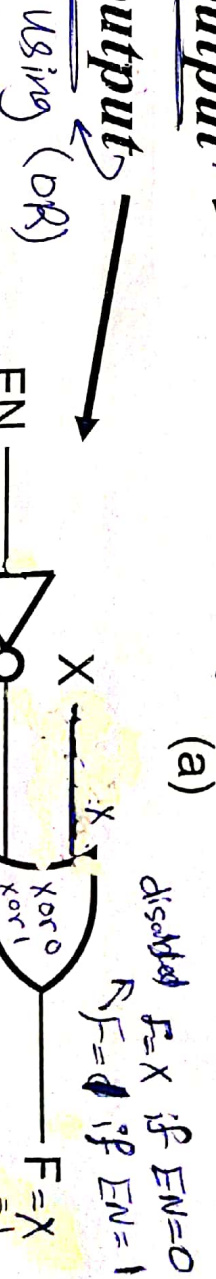
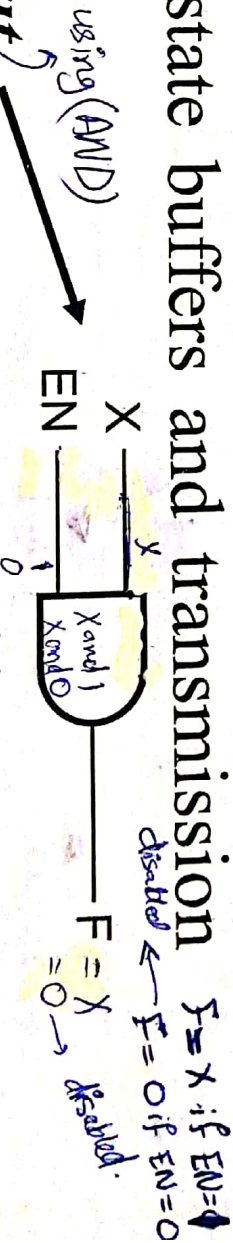
# Enabling Function

- Enabling permits an input signal to pass through to an output
- Disabling blocks an input signal from passing through to an output, replacing it with a fixed value

\* when Enable = 0. Then the function have a fixed value Hi-Z (as for three-state buffers and transmission gates), 0, or 1

\* when Enable = 1 then the function = X. whether X = 0 or X = 1

- When disabled, 0 output
- When disabled, 1 output

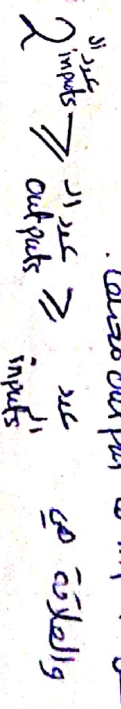


# Decoding

8 conventional logic block.

تختلف في بناء (1) و (2) و (3) و (4) و (5) و (6) و (7) و (8) في (m) و (n) و (k) و (l) و (p) و (q) و (r) و (s) و (t) و (u) و (v) و (w) و (x) و (y) و (z) و (aa) و (ab) و (ac) و (ad) و (ae) و (af) و (ag) و (ah) و (ai) و (aj) و (ak) و (al) و (am) و (an) و (ao) و (ap) و (aq) و (ar) و (as) و (at) و (au) و (av) و (aw) و (ax) و (ay) و (az) و (ba) و (bb) و (bc) و (bd) و (be) و (bf) و (bg) و (bh) و (bi) و (bj) و (bk) و (bl) و (bm) و (bn) و (bo) و (bp) و (bq) و (br) و (bs) و (bt) و (bu) و (bv) و (bw) و (bx) و (by) و (bz) و (ca) و (cb) و (cc) و (cd) و (ce) و (cf) و (cg) و (ch) و (ci) و (cj) و (ck) و (cl) و (cm) و (cn) و (co) و (cp) و (cq) و (cr) و (cs) و (ct) و (cu) و (cv) و (cw) و (cx) و (cy) و (cz) و (da) و (db) و (dc) و (dd) و (de) و (df) و (dg) و (dh) و (di) و (dj) و (dk) و (dl) و (dm) و (dn) و (do) و (dp) و (dq) و (dr) و (ds) و (dt) و (du) و (dv) و (dw) و (dx) و (dy) و (dz) و (ea) و (eb) و (ec) و (ed) و (ee) و (ef) و (eg) و (eh) و (ei) و (ej) و (ek) و (el) و (em) و (en) و (eo) و (ep) و (eq) و (er) و (es) و (et) و (eu) و (ev) و (ew) و (ex) و (ey) و (ez) و (fa) و (fb) و (fc) و (fd) و (fe) و (ff) و (fg) و (fh) و (fi) و (fj) و (fk) و (fl) و (fm) و (fn) و (fo) و (fp) و (fq) و (fr) و (fs) و (ft) و (fu) و (fv) و (fw) و (fx) و (fy) و (fz) و (ga) و (gb) و (gc) و (gd) و (ge) و (gf) و (gg) و (gh) و (gi) و (gj) و (gk) و (gl) و (gm) و (gn) و (go) و (gp) و (gq) و (gr) و (gs) و (gt) و (gu) و (gv) و (gw) و (gx) و (gy) و (gz) و (ha) و (hb) و (hc) و (hd) و (he) و (hf) و (hg) و (hh) و (hi) و (hj) و (hk) و (hl) و (hm) و (hn) و (ho) و (hp) و (hq) و (hr) و (hs) و (ht) و (hu) و (hv) و (hw) و (hx) و (hy) و (hz) و (ia) و (ib) و (ic) و (id) و (ie) و (if) و (ig) و (ih) و (ii) و (ij) و (ik) و (il) و (im) و (in) و (io) و (ip) و (iq) و (ir) و (is) و (it) و (iu) و (iv) و (iw) و (ix) و (iy) و (iz) و (ja) و (jb) و (jc) و (jd) و (je) و (jf) و (jg) و (jh) و (ji) و (jj) و (jk) و (jl) و (jm) و (jn) و (jo) و (jp) و (jq) و (jr) و (js) و (jt) و (ju) و (jv) و (jw) و (jx) و (jy) و (jz) و (ka) و (kb) و (kc) و (kd) و (ke) و (kf) و (kg) و (kh) و (ki) و (kj) و (kk) و (kl) و (km) و (kn) و (ko) و (kp) و (kq) و (kr) و (ks) و (kt) و (ku) و (kv) و (kw) و (kx) و (ky) و (kz) و (la) و (lb) و (lc) و (ld) و (le) و (lf) و (lg) و (lh) و (li) و (lj) و (lk) و (ll) و (lm) و (ln) و (lo) و (lp) و (lq) و (lr) و (ls) و (lt) و (lu) و (lv) و (lw) و (lx) و (ly) و (lz) و (ma) و (mb) و (mc) و (md) و (me) و (mf) و (mg) و (mh) و (mi) و (mj) و (mk) و (ml) و (mm) و (mn) و (mo) و (mp) و (mq) و (mr) و (ms) و (mt) و (mu) و (mv) و (mw) و (mx) و (my) و (mz) و (na) و (nb) و (nc) و (nd) و (ne) و (nf) و (ng) و (nh) و (ni) و (nj) و (nk) و (nl) و (nm) و (nn) و (no) و (np) و (nq) و (nr) و (ns) و (nt) و (nu) و (nv) و (nw) و (nx) و (ny) و (nz) و (oa) و (ob) و (oc) و (od) و (oe) و (of) و (og) و (oh) و (oi) و (oj) و (ok) و (ol) و (om) و (on) و (oo) و (op) و (oq) و (or) و (os) و (ot) و (ou) و (ov) و (ow) و (ox) و (oy) و (oz) و (pa) و (pb) و (pc) و (pd) و (pe) و (pf) و (pg) و (ph) و (pi) و (pj) و (pk) و (pl) و (pm) و (pn) و (po) و (pp) و (pq) و (pr) و (ps) و (pt) و (pu) و (pv) و (pw) و (px) و (py) و (pz) و (qa) و (qb) و (qc) و (qd) و (qe) و (qf) و (qg) و (qh) و (qi) و (qj) و (qk) و (ql) و (qm) و (qn) و (qo) و (qp) و (qq) و (qr) و (qs) و (qt) و (qu) و (qv) و (qw) و (qx) و (qy) و (qz) و (ra) و (rb) و (rc) و (rd) و (re) و (rf) و (rg) و (rh) و (ri) و (rj) و (rk) و (rl) و (rm) و (rn) و (ro) و (rp) و (rq) و (rr) و (rs) و (rt) و (ru) و (rv) و (rw) و (rx) و (ry) و (rz) و (sa) و (sb) و (sc) و (sd) و (se) و (sf) و (sg) و (sh) و (si) و (sj) و (sk) و (sl) و (sm) و (sn) و (so) و (sp) و (sq) و (sr) و (ss) و (st) و (su) و (sv) و (sw) و (sx) و (sy) و (sz) و (ta) و (tb) و (tc) و (td) و (te) و (tf) و (tg) و (th) و (ti) و (tj) و (tk) و (tl) و (tm) و (tn) و (to) و (tp) و (tq) و (tr) و (ts) و (tt) و (tu) و (tv) و (tw) و (tx) و (ty) و (tz) و (ua) و (ub) و (uc) و (ud) و (ue) و (uf) و (ug) و (uh) و (ui) و (uj) و (uk) و (ul) و (um) و (un) و (uo) و (up) و (uq) و (ur) و (us) و (ut) و (uu) و (uv) و (uw) و (ux) و (uy) و (uz) و (va) و (vb) و (vc) و (vd) و (ve) و (vf) و (vg) و (vh) و (vi) و (vj) و (vk) و (vl) و (vm) و (vn) و (vo) و (vp) و (vq) و (vr) و (vs) و (vt) و (vu) و (vv) و (vw) و (vx) و (vy) و (vz) و (wa) و (wb) و (wc) و (wd) و (we) و (wf) و (wg) و (wh) و (wi) و (wj) و (wk) و (wl) و (wm) و (wn) و (wo) و (wp) و (wq) و (wr) و (ws) و (wt) و (wu) و (wv) و (ww) و (wx) و (wy) و (wz) و (xa) و (xb) و (xc) و (xd) و (xe) و (xf) و (xg) و (xh) و (xi) و (xj) و (xk) و (xl) و (xm) و (xn) و (xo) و (xp) و (xq) و (xr) و (xs) و (xt) و (xu) و (xv) و (xw) و (xx) و (xy) و (xz) و (ya) و (yb) و (yc) و (yd) و (ye) و (yf) و (yg) و (yh) و (yi) و (yj) و (yk) و (yl) و (ym) و (yn) و (yo) و (yp) و (yq) و (yr) و (ys) و (yt) و (yu) و (yv) و (yw) و (yx) و (yy) و (yz) و (za) و (zb) و (zc) و (zd) و (ze) و (zf) و (zg) و (zh) و (zi) و (zj) و (zk) و (zl) و (zm) و (zn) و (zo) و (zp) و (zq) و (zr) و (zs) و (zt) و (zu) و (zv) و (zw) و (zx) و (zy) و (zz)

- Decoding: the conversion of an  $n$ -bit input code to an  $m$ -bit output code with  $n \leq m \leq 2^n$  such that each valid code word produces a unique output code

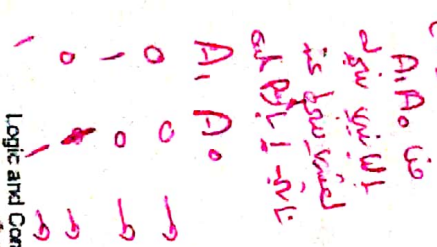


- Circuits that perform decoding are called **decoders**
- Functional blocks for decoding are
  - called  **$n$ -to- $m$  line decoders** where  $m \leq 2^n$ , and
  - generate  $2^n$  (or fewer) minterms for the  $n$  input variables

\* Decoder: -



\* Relation:  $n \leq m \leq 2^n$



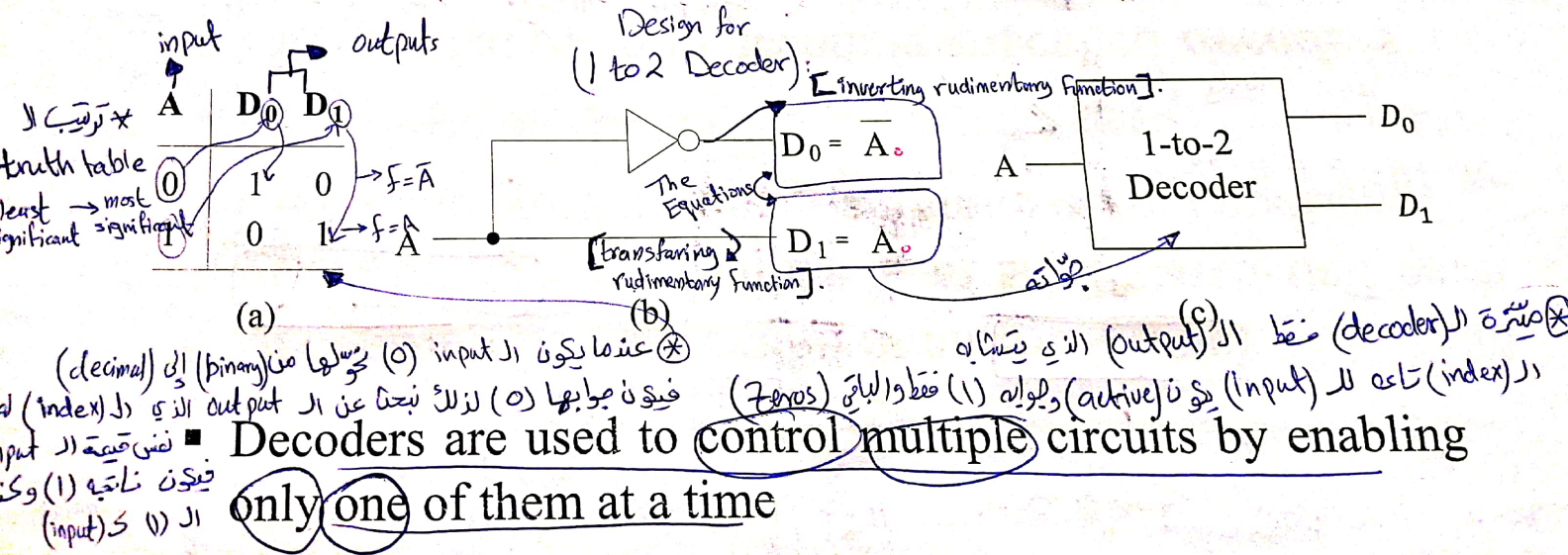
# 1-to-2 Line Decoder when $n=1$

\* A: input: Address  
 \* D: output: Data.

When the decimal value of A equals the subscript of  $D_i$ , that  $D_i$  will be 1 and all others will be 0's

قائمة (A) بال (decimal) هي ال (index) ال output الذي قيمته (1) \* قائمة ال (decoder) : أنه عند (output) واحد هو (active) في وقت معين .

**Only one output is active at a time**





# Decoder Expansion (less cost).

---

- General procedure given in book for any decoder with  $n$  inputs and  $2^n$  outputs (المُدخلات والمُخرجات)
- This procedure builds a decoder backward from the outputs using

1. Let  $k = n$

2. We need  $2^k$  2-input AND gates driven as follows:

- If  $k$  is even, drive the gates using two  $k/2$ -to- $2^{k/2}$  decoders
- If  $k$  is odd, drive the gates using one  $(k+1)/2$ -to- $2^{(k+1)/2}$  decoder and one  $(k-1)/2$ -to- $2^{(k-1)/2}$  decoder

3. For each decoder resulting from step 2, repeat step 2 until  $k = 1$ . For  $k = 1$ , use 1-to-2 decoder

(loop)  
reach the simplest decoder

# Decoder Expansion - Example 1

## 3-to-8-line decoder

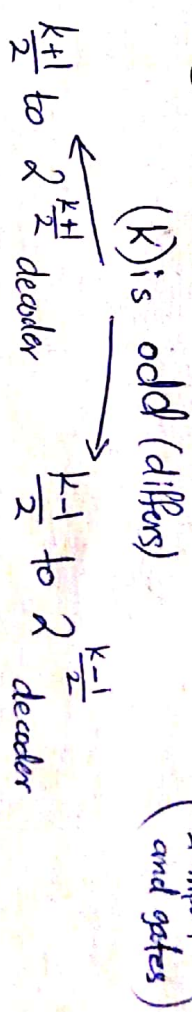
نبدأ مع الرتبة 3 للبيانات (البيانات)

\*  $k = n = 3 = \text{number of inputs}$

معلوم

\* We need  $2^3$  (8) 2-input AND gates driven as follows:  $2^n = \text{عدد البوابات (2-input and gates)}$

\*  $k$  is odd, so split to:



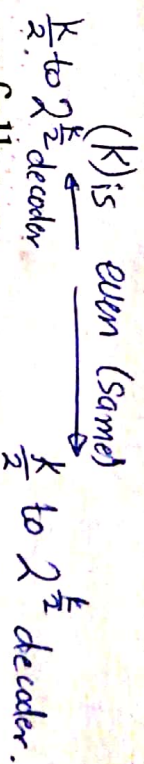
- 2-to-4-line decoder
- 1-to-2-line decoder

\* 2-to-4-line decoder  $\rightarrow k = n = 2$

- We need  $2^2$  (4) 2-input AND gates driven as follows:

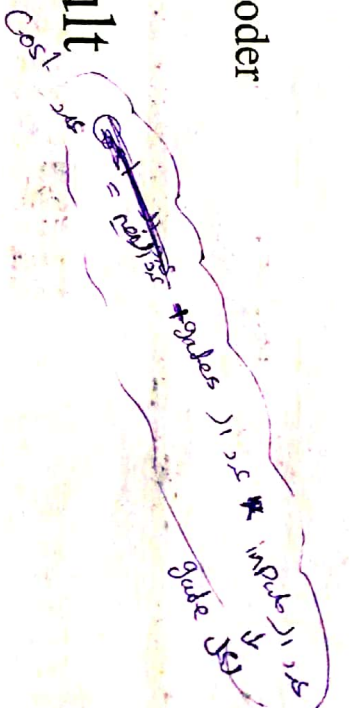
- $k$  is even, so split to:

- Two 1-to-2-line decoder



System expanded manually

See next slide for result

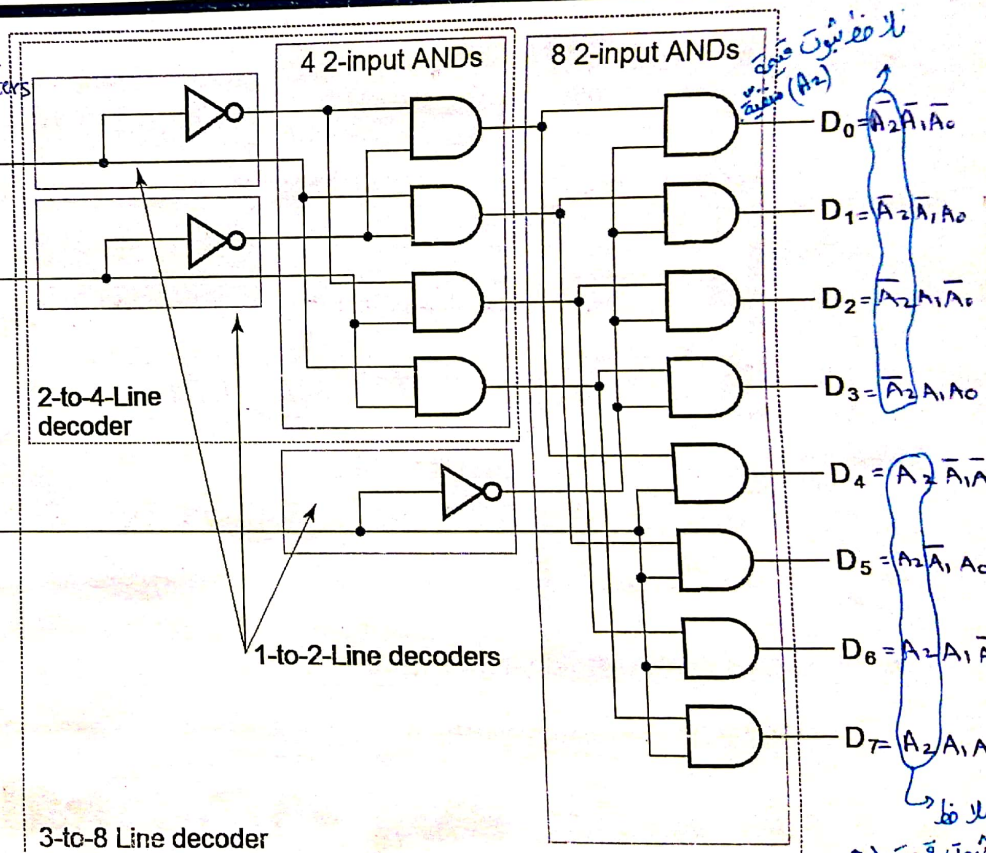


# Decoder Expansion - Example 1

- AND gates. AND gates
- GN =  $8 \times 2 + 4 \times 2 + 3$
- GN = 27

**Straight forward design has the same GN cost**

من ناحية (cost) متساويان ولكن من ناحية (delay) مختلفا.  
 \* لكي نسهل الرسم بنبدأ من اليمين ثم نبدأ من اليسار  
 نبدأ من اليمين لكل (and gate) من اليمين  
 $D_0 = \bar{A}_2 \bar{A}_1 \bar{A}_0$  وهكذا نكمل للخارج  
 \* يجوز تبديل ( $A_0$  بـ  $A_2$ ) أو غيرها  
 أي أن أي ترتيب للبيوت والبيوت بعد  
 أن ترتيب ال outputs يكون عليه  
 $0 \equiv \bar{A}_2 \bar{A}_1 \bar{A}_0$   
 $1 \equiv \bar{A}_2 \bar{A}_1 A_0$





# Decoder Expansion - Example 2

6-to-64-line decoder =  $1 \times 2^2$

$k = n = 6 \rightarrow$  The number of inputs.

We need  $2^6$  (64) 2-input AND gates driven as follows: ( $2^n =$  number of AND gates)

$k$  is even, so split to:

Two 3-to-8-line decoders

Each 3-to-8-line decoder is designed as shown in Example 1

OR continue the loop!

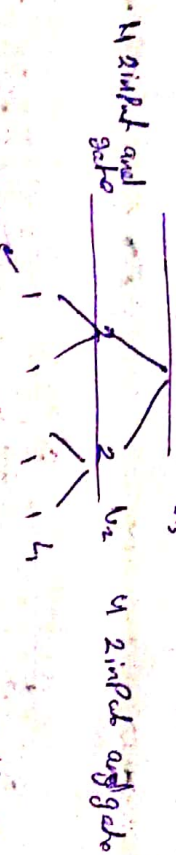
$k = n = 3 \rightarrow$  odd  $\rightarrow$  (1 to 2) decoder.

(2 to 4) decoder.

4-to-16 line decoder

continue the loop!  $\rightarrow n = k = 2$  even

1 to 2 decoder.  $\rightarrow$  1 to 2 decoder.

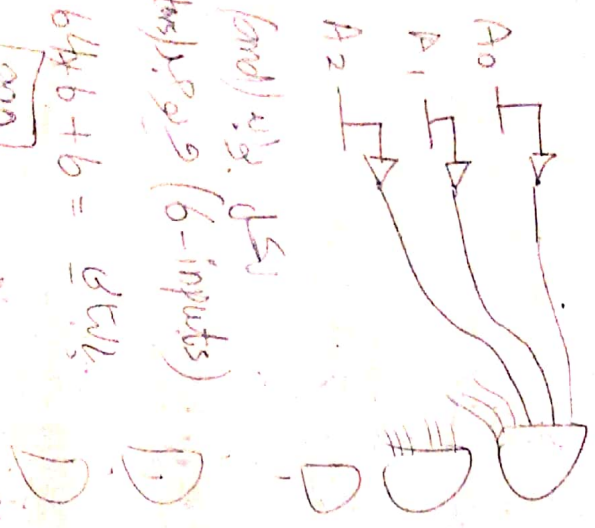


24 saved gates, 25  
52 = Cost

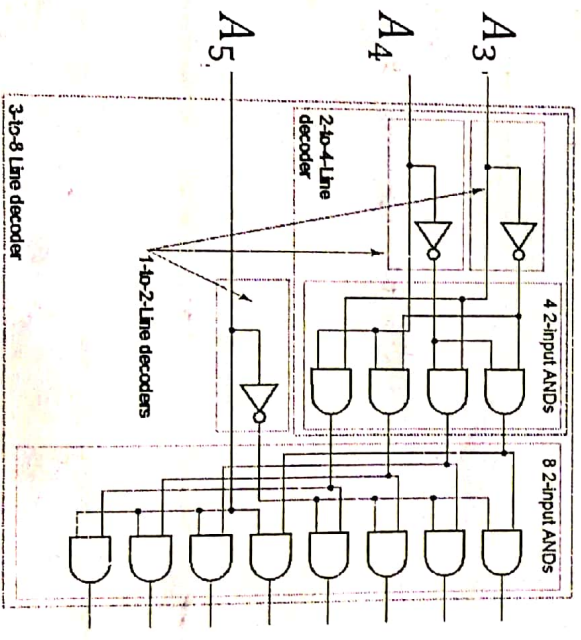
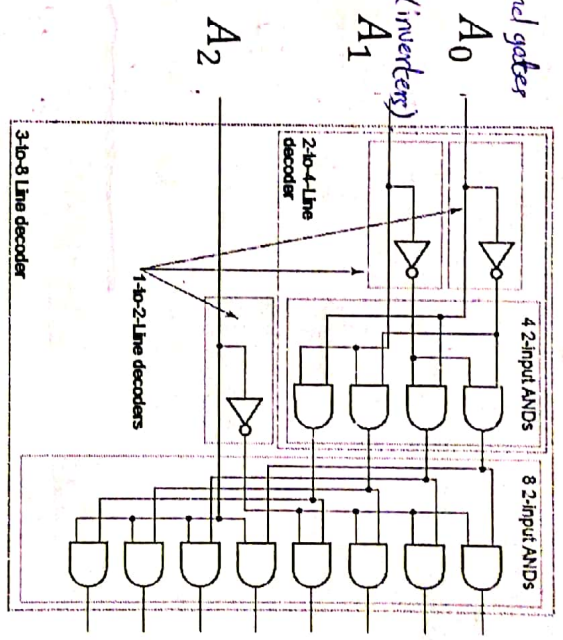
# Decoder Expansion - Example 2

- GN =  $64 \times 2 + 16 \times 2 + 8 \times 2 + 6$   $\rightarrow (2^6)$  and gates.  $\rightarrow (2^2)$  and gates
- GN = 182  $\rightarrow (2^4)$  and gates.
- Straight forward design has GN cost of 390

Traditional  $\downarrow$



6 and 2's. (6-inputs)  
 $64 \times 6 + 6 = 390$



$$D_0 = \overline{A_5} \overline{A_4} \overline{A_3} \overline{A_2} \overline{A_1} A_0$$

$$D_1 = \overline{A_5} \overline{A_4} \overline{A_3} \overline{A_2} \overline{A_1} \overline{A_0}$$

$$D_{62} = \overline{A_5} \overline{A_4} \overline{A_3} \overline{A_2} \overline{A_1} \overline{A_0}$$

$$D_{63} = \overline{A_5} \overline{A_4} \overline{A_3} \overline{A_2} \overline{A_1} A_0$$

For the cost decoder Expansion is better. but for the delay the straight forward design is better because it has one level implementation



using Enables.

# Building Larger Decoders

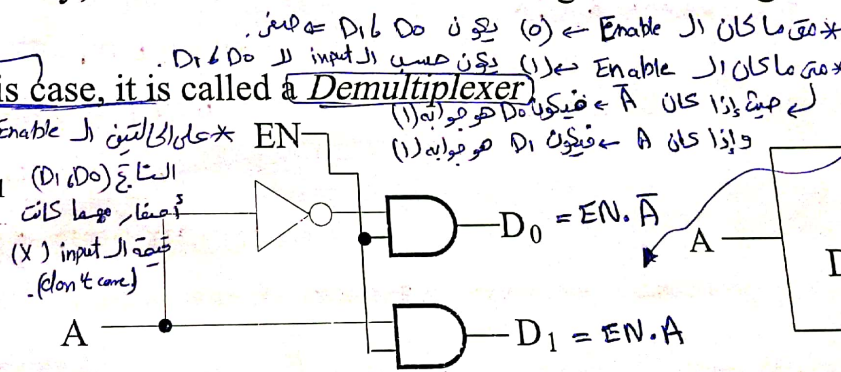
- Method\_1: Decoder Expansion (الطريقة السابقة)
- Method\_2: Using Small Decoders with Enable input (الطريقة الحالية):
- Example: 1-to-2 line decoder with enable
  - In general, attach *m-enabling* circuits to the outputs
  - See truth table below for function
    - Note use of X's to denote both 0 and 1
    - Combination containing two X's represent two binary combinations
- Alternatively, can be viewed as distributing value of signal EN to 1 of 2 outputs

$D_{mux} = (\text{decoder} + \text{Enable})$

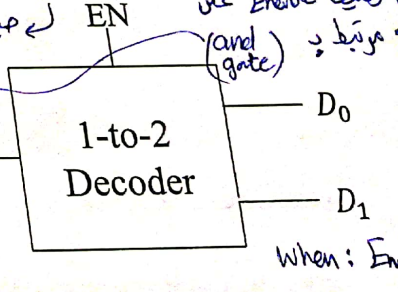
لا يشغل  
لا يشغل

EN	A	D <sub>0</sub>	D <sub>1</sub>
0	X	0	0
1	0	1	0
1	1	0	1

(a)



(b)

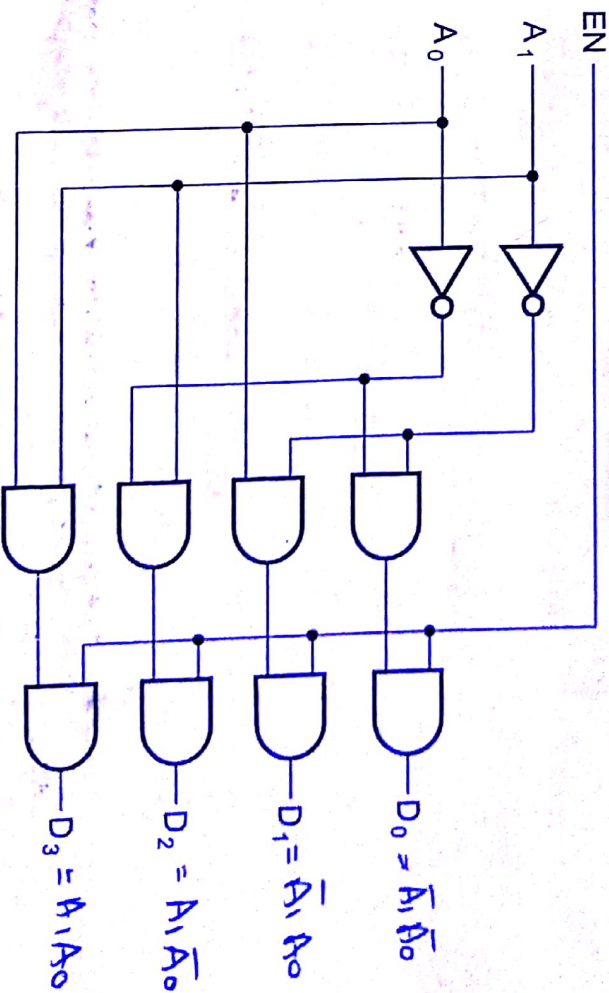
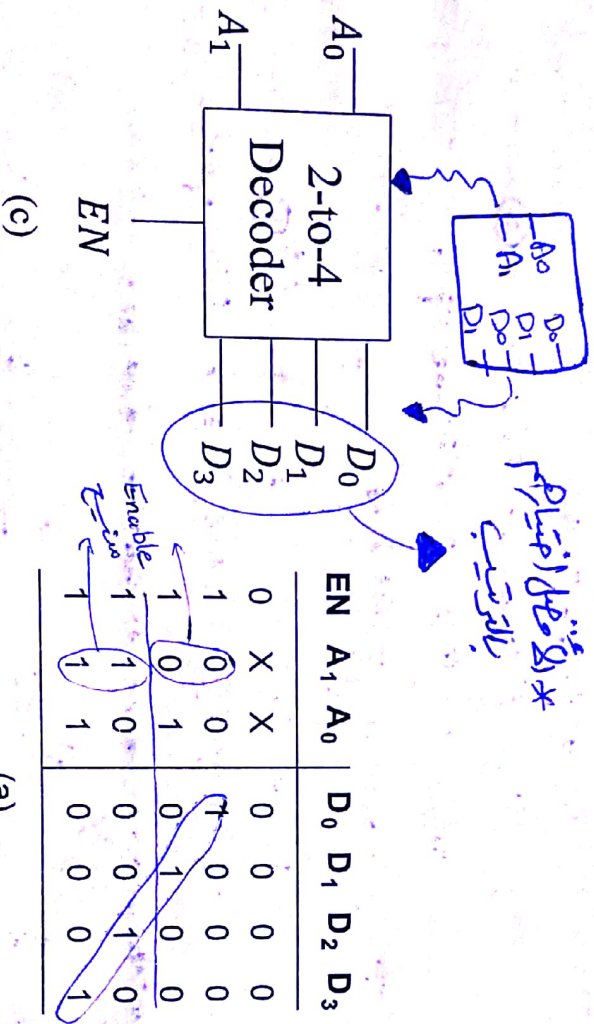


(c)

# 2-to-4 Line Decoder with Enable

- Attach **4-enabling** circuits to the outputs by *an and gates*
- See truth table below for function
  - Combination containing two X's represent four binary combinations *Don't cares represent 4 حالات*
- Alternatively, can be viewed as distributing value of signal  $\overline{EN}$  to 1 of 4 outputs

• In this case, it is called a **Demultiplexer**



(a)

(b) *في حالة (البيانات الوسيطة) لن يكون*

*(most significant digit) هو (Enable) غالباً*

output 1  
 كيف يمكن ان تتغير  
 inputs 2-2-to-4 Decoder using  
 وتعمل توصيل 2-2-to-4

## 2-to-4 Decoder using 1-to-2 Decoders and Inverters

$n = 2 = \text{inputs}$   
 $m = 2^n = \text{outputs} = \text{number of 2-input AND gates.}$

two 1-2 decoder and one inverter

1-2 Decoder

$A_1$	0	1
$A_0$	0	1

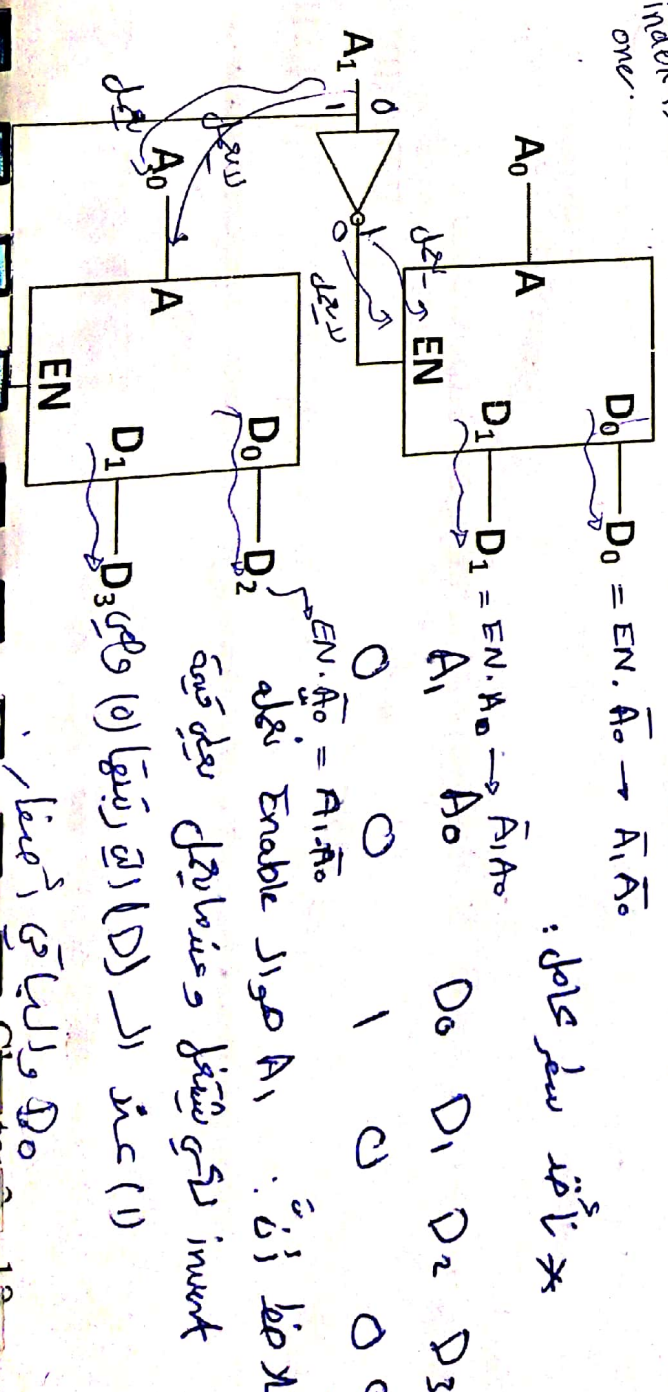
1st 1-to-2 Decoder

Decoder	$D_0$	$D_1$
(0)	1	0
(1)	0	1
(2)	0	0
(3)	0	0

2nd 1-to-2 Decoder

$D_2$	$D_3$
0	0
0	0
1	0
0	1

most significant digit  
 Enable  
 (1) (0) أي إشارة وإشارة  
 لا يعمل لكي لا يطلع قسم  
 غالباً ال (Enable) هو ال  
 (most significant digit.)

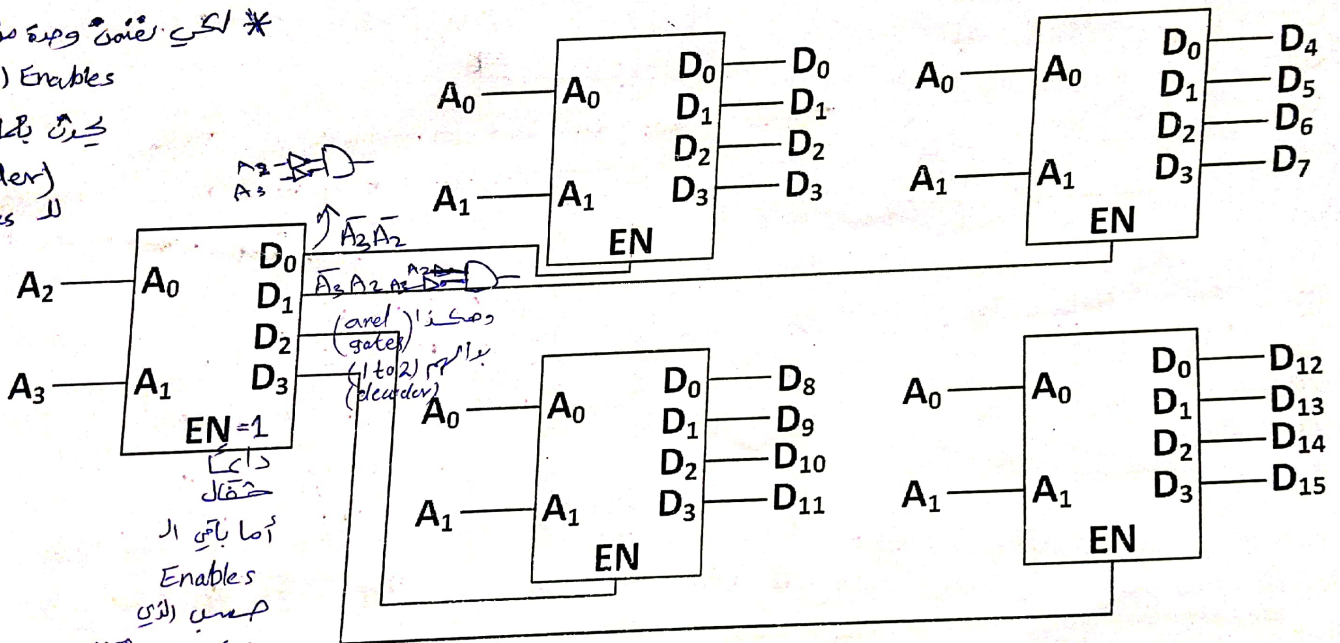




$n = 4 = \text{inputs}$   
 $m = 16 = \text{outputs} = 2^n = \text{number of AND gates}$   
**4-to-16 Decoder using Only 2-to-4 Decoders**

عدد 2 to 4 decoders (X) عدد 4 \* X = 16 → X = 4 number of decoders.

\* لكي نكوّن 16 مخرجات من 4 مدخلات  
 (1) Enables  
 يجب علينا  
 (decoder)  
 Enables 4



لقد  
 جعلنا  
 4 enables  
 من باقي ال  
 من (1) enables  
 وطلع ال outputs  
 من 4



# Combinational Logic Implementation

## - Decoder and OR Gates

- Implement  $m$  functions of  $n$  variables with:
  - $\rightarrow f/g/... (number\ of\ functions).$

- Sum-of-minterms expressions
- One  $n$ -to- $2^n$  line decoder
- $m$  OR gates, one for each function
- For each function, the OR gate has  $k$  inputs, where  $k$  is the number of minterms in the function

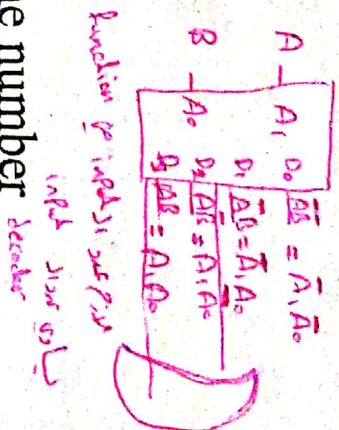
### Approach 1:

- Find the truth table for the functions
- Make a connection to the corresponding OR from the corresponding decoder output wherever a 1 appears in the truth table

### Approach 2

- Find the minterms for each output function
- OR the minterms together

Ex:  $F(A, B) = \sum m_{1,3}$



# Example 1

- Implement function  $f$  using decoder and OR gate:

$f(x, y, z) = x\bar{z} + \bar{x}y$  → from Boolean Equation to some of minterms.

- $m=1$  number of functions.
- $n=3$  variables → **3-to-8 decoder**

or using truth table.  
 $x\bar{z} \rightarrow x y \bar{z}$  or  $x \bar{y} \bar{z}$   
 110      100

- One function → **One OR gate**

- Solution: Convert  $f$  to SOM format

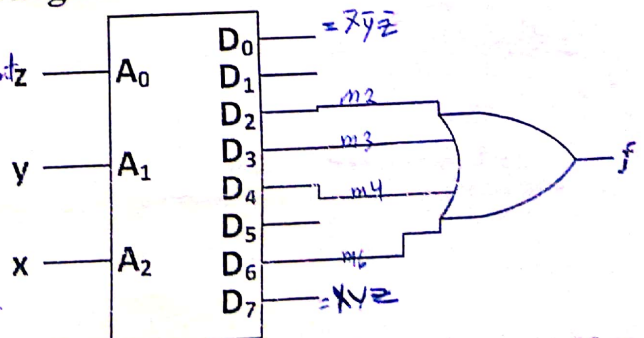
$f = x\bar{z}(y + \bar{y}) + \bar{x}y(z + \bar{z}) = xy\bar{z} + x\bar{y}\bar{z} + \bar{x}yz + \bar{x}y\bar{z}$

$f(x, y, z) = \sum m(2, 3, 4, 6) \rightarrow$  4-input OR gate

**Decoder is a Minterm Generator**

کس کی ترتیب

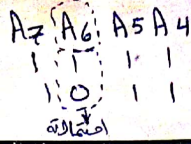
least significant digit  
 most significant







# Example 4



Implement the following set of odd parity functions of

$(A_7, A_6, A_5, A_4)$

\*  $P_1 = A_7 \oplus A_5 \oplus A_4$

\*  $P_2 = A_7 \oplus A_6 \oplus A_4$

\*  $P_3 = A_7 \oplus A_6 \oplus A_5$

Finding sum of minterms expressions

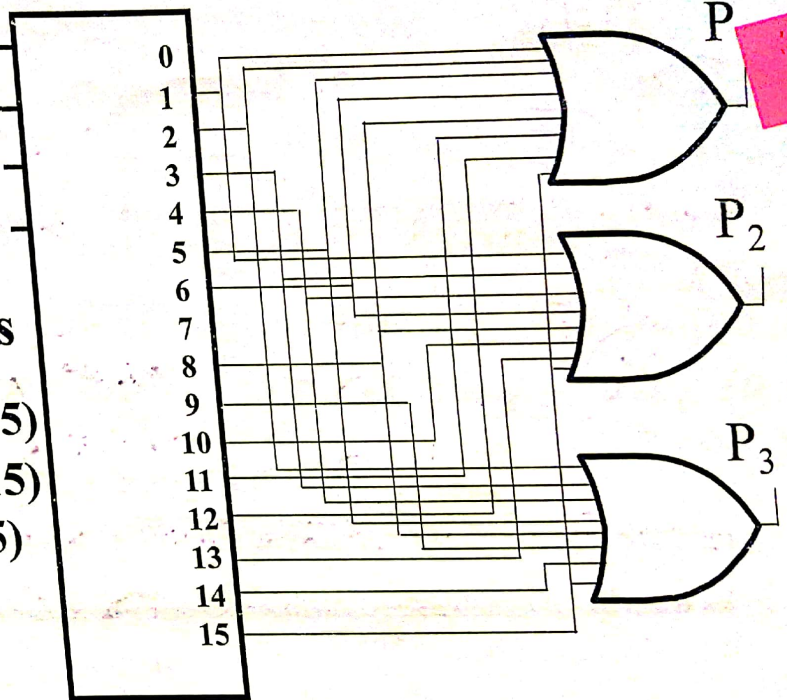
$P_1 = \sum_m(1,2,5,6,8,11,12,15)$

$P_2 = \sum_m(1,3,4,6,8,10,13,15)$

$P_3 = \sum_m(2,3,4,5,8,9,14,15)$

Find circuit

Is this a good idea?



عدد ال (ones) فردی و ذلك ب:   
 $A_4 = 1$    
 $A_5 = 1$    
 $A_7 = 1$    
 و عدد ال (ones) فردی و ذلك ب:   
 $A_4 = 1$    
 $A_6 = 1$    
 $A_7 = 1$    
 و عدد ال (ones) فردی و ذلك ب:   
 $A_5 = 1$    
 $A_6 = 1$    
 $A_7 = 1$

every one needs 3-8 decoder. but when we include  $A_7$  in  $A_5, A_6$  in  $P_1, P_2, P_3$  it will use 4-to-16 decoder.

# Examples

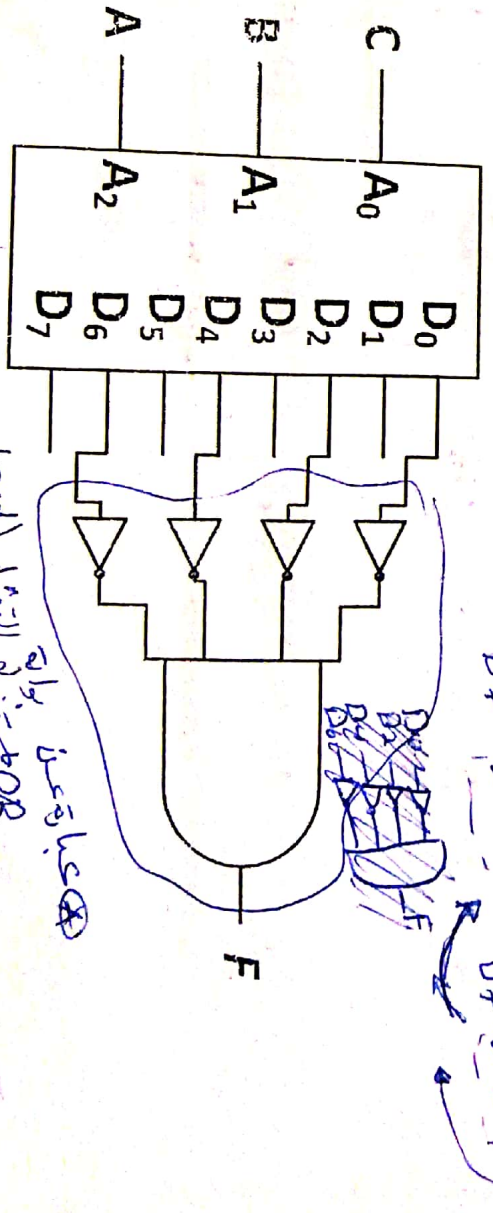
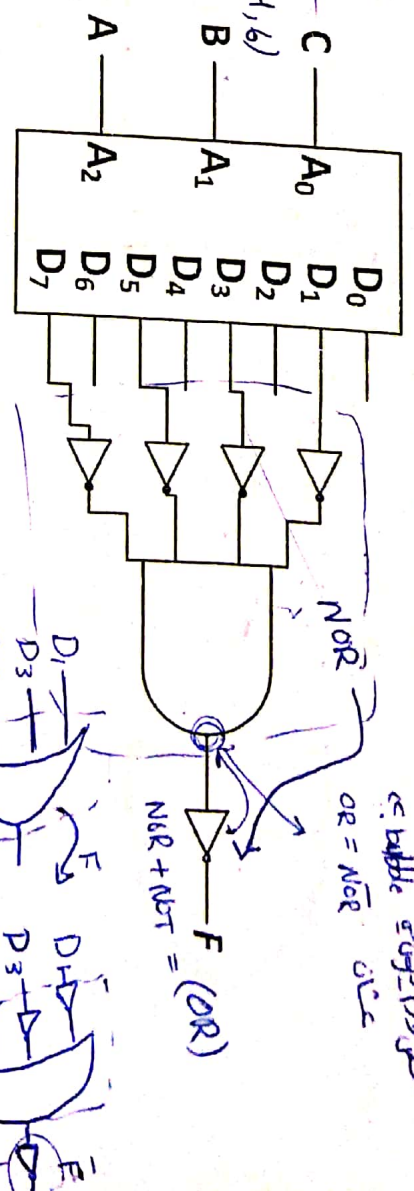
- Implement function  $F$  using 3-to-8 decoder, AND gate and (4) inverters:  $F(A, B, C) = \sum m(1, 3, 5, 7)$   $\rightarrow$  4 minterms. + 4 maxterms.

## Solution with 5 inverters

$\sum m F = \sum_{i=0}^7 \pi_{M_i} F$   
 $\sum m F = \pi_{M_i} \bar{F}$   
 $\pi_{M_i} \bar{F} = \sum m_i \bar{F}$   
 $(\bar{F}) = \sum m(0, 2, 4, 6)$   
*as sum of minterms*

## Solution with 4 inverters

- $F(A, B, C) = \Pi_M(0, 2, 4, 6)$   
 $F \rightarrow \pi_M(0, 2, 4, 6)$



# Encoding (one input = 1)

$m \rightarrow$  inputs /  $n \rightarrow$  outputs.

decoder (one output = 1)  
 $n \rightarrow$  inputs /  $m \rightarrow$  outputs.

- Encoding:** the opposite of decoding - the conversion of an  $m$ -bit input code to a  $n$ -bit output code with  $n \leq m \leq 2^n$  such that each valid code word produces a unique output code
- Circuits that perform encoding are called **encoders**
- An encoder has  $2^n$  (or fewer) input lines and  $n$  output lines which **generate the binary code corresponding to the input values**
- Typically, an encoder converts a code containing exactly one bit that is 1 to a binary code corresponding to the position in which the 1 appears

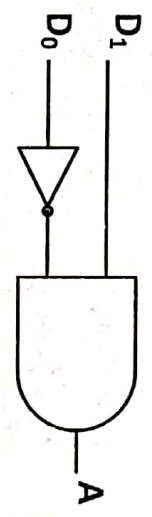
inputs  $\rightarrow D \rightarrow$  Data.  
 outputs  $\rightarrow A \rightarrow$  Address.

(12/20)

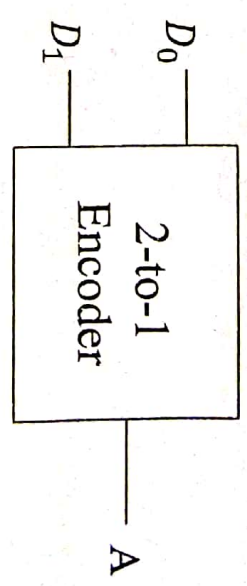
مُدخل ال (input) الذي يحتوي على (1)  
 رقم (index) وترتيبه (ق) في المخرجات (outputs)

# 2-to-1 Encoder & 4-to-2 Encoder

$D_1$	$D_0$	$A$
0	0	Invalid Input
0	1	0 (decimal)
1	0	1 (decimal)
1	1	Invalid Input



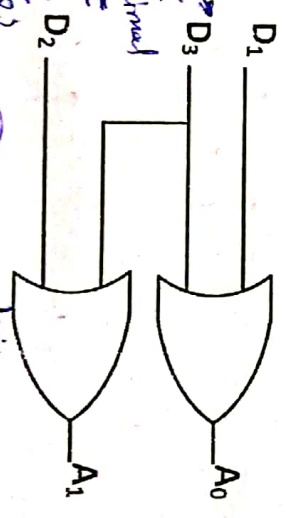
decimal  $A = D_1 \cdot \overline{D_0}$



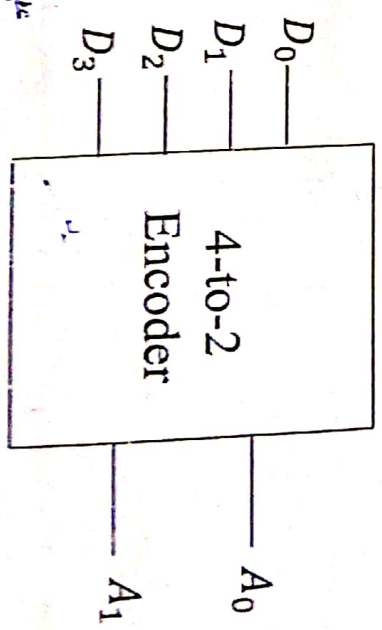
دعوى output  $D_3$  decimal

$D_3$	$D_2$	$D_1$	$D_0$	$A_1$	$A_0$
0	0	0	1	0	0
0	0	1	0	0	1
0	1	0	0	1	0
1	0	0	0	1	1

(a)



(b)



(c)

دعوى output  $A_1$  decimal

$A_0 \equiv D_1 + D_3$

$A_1 \equiv D_2 + D_0$

by default (0)

(a)

(b)

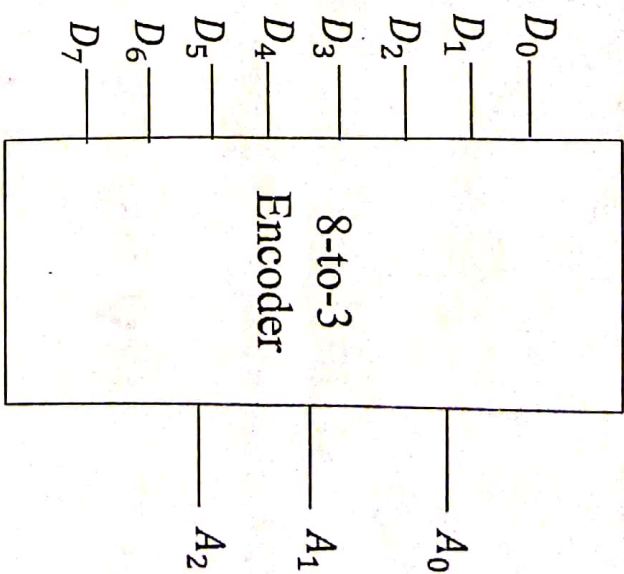
(c)



# 8-to-3 Encoder (Octal-to-Binary Encoder)

binary ← Octal من جدول في ك

$D_7$	$D_6$	$D_5$	$D_4$	$D_3$	$D_2$	$D_1$	$D_0$	$A_2$	$A_1$	$A_0$
0	0	0	0	0	0	0	1	0	0	0
0	0	0	0	0	0	1	0	0	0	1
0	0	0	0	0	1	0	0	0	1	0
0	0	0	0	1	0	0	0	0	1	1
0	0	0	1	0	0	0	0	1	0	0
0	0	1	0	0	0	0	0	1	0	1
0	1	0	0	0	0	0	0	1	1	0
1	0	0	0	0	0	0	0	1	1	1

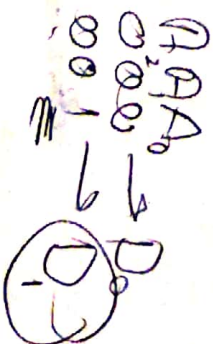


(a)

index  
 \* لكل bit ترتيبا نفسي ال  
 يمكن صونا

$$\begin{aligned}
 A_0 &= D_0 + D_1 + D_2 + D_3 + D_4 + D_5 + D_6 + D_7 \\
 A_1 &= D_2 + D_3 + D_6 + D_7 \\
 A_2 &= D_4 + D_5 + D_6 + D_7
 \end{aligned}$$

(c)



10 - inputs, ... 4 outputs (digits)  $2^4 \rightarrow 16 \checkmark$  (يفي (10) لا يفي (10)  $2^3 \rightarrow 8 \times$

## Decimal-to-BCD Encoder

- **Inputs:** 10 bits corresponding to decimal digits 0 through 9, ( $D_0, \dots, D_9$ )
- **Outputs:** 4 bits with BCD codes ( $A_3, A_2, A_1, A_0$ )
- **Function:** If input bit  $D_i$  is a 1, then the output is the BCD code for  $i$
- The truth table could be formed, but alternatively, the equations for each of the four outputs can be obtained directly

# Decimal-to-BCD Encoder Cont.

- Input  $D_i$  is a term in equation  $A_j$  if bit  $A_j$  is 1 in the binary value for  $i$

Equations:

$$A_3 = D_8 + D_9$$

$$A_2 = D_4 + D_5 + D_6 + D_7$$

$$A_1 = D_2 + D_3 + D_6 + D_7$$

$$A_0 = D_1 + D_3 + D_5 + D_7 + D_9$$

- What happens if two inputs are high simultaneously?
  - For example if  $D_3$  and  $D_6$  are high, then the output is 0111 which indicates that only  $D_7$  is high ???

Solution: Establish input priority

اولی ورودی  
اولی ورودی  
اولی ورودی

# Priority Encoder

\* If more than one input value is 1, then the encoder just designed does not work

One encoder that can accept all possible combinations of input values and produce a meaningful result is a priority encoder

يعطي الأولوية لترقيم شروط معينة

Among the 1s that appear, it selects the most significant input position (or the least significant input position) containing a 1 and responds with the corresponding binary code for that position

- High priority encoder gives priority for the input whose value is 1 and has the highest subscript  
يعطي الأولوية لترقيم الأخرى عند صلا بال
- low priority encoder gives priority for the input whose value is 1 and has the lowest subscript  
يعطي الأولوية لترقيم عند (1) بال input

▪ If all inputs are 0's, what happens?

قطع لعمارة إذا ما كان  
Valid bit  
وظيفة ال

• Define an output (V) to encode whether the input is valid or not

• When all inputs are 0's, V is set to 0 indicating that the input is invalid,

otherwise V is set to 1

و إذا لا تعطى  
وحدة V = input

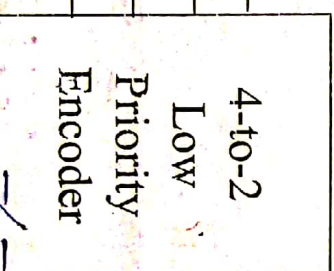
# 4-to-2 Low Priority Encoder

#_of Minterms/ ROWS	D <sub>3</sub>	D <sub>2</sub>	D <sub>1</sub>	D <sub>0</sub>	A <sub>1</sub>	A <sub>0</sub>	V
2 <sup>0</sup> = 1	0	0	0	1	X	X	Valid
2 <sup>1</sup> = 2	X	1	0	0	0	1	Valid
2 <sup>2</sup> = 4	X	X	1	0	1	0	Valid
2 <sup>3</sup> = 8	X	X	X	1	0	0	Valid
2 <sup>0</sup> = 1	1	0	0	0	1	0	Valid

Handwritten notes: 2<sup>0</sup> = 1, 2<sup>1</sup> = 2, 2<sup>2</sup> = 4, 2<sup>3</sup> = 8. Also includes Arabic notes: "عدد المداخل", "عدد المخرجات", "عدد المخرجات = 2<sup>n</sup>".



\* valid bit = 0 . when all inputs are (0)  
 \* valid bit = 1 . when one or two or more than input = (1)



Handwritten equations for the encoder outputs:

$$A_2 = D_2 \bar{D}_1 \bar{D}_0 + \bar{D}_3 D_2 \bar{D}_1 \bar{D}_0$$

$$A_1 = D_2 \bar{D}_1 \bar{D}_0 + \bar{D}_3 D_2 \bar{D}_1 \bar{D}_0$$

Handwritten equations for the encoder outputs:

$$A_0 = D_1 \bar{D}_0 + D_3 \bar{D}_2 \bar{D}_1 \bar{D}_0$$

$$A_0 = \bar{D}_0 (D_1 + D_3 \bar{D}_2 \bar{D}_1)$$

$$A_0 = \bar{D}_0 (D_1 + D_3 \bar{D}_2)$$

$$A_0 = D_1 \bar{D}_0 + D_3 \bar{D}_2 \bar{D}_0$$

Handwritten equations for the encoder outputs:

$$A_1 = D_2 \bar{D}_1 \bar{D}_0 + D_3 \bar{D}_2 \bar{D}_1 \bar{D}_0$$

$$A_1 = \bar{D}_1 \bar{D}_0 (D_2 + D_3 \bar{D}_2)$$

$$A_1 = \bar{D}_1 \bar{D}_0 (D_2 + D_3)$$

$$A_1 = D_2 \bar{D}_1 \bar{D}_0 + D_3 \bar{D}_1 \bar{D}_0$$

Handwritten equation for the valid bit output:

$$V = D_3 + D_2 + D_1 + D_0$$

