

Circuits II Notebook

Dr. Nabeel Tawalbeh

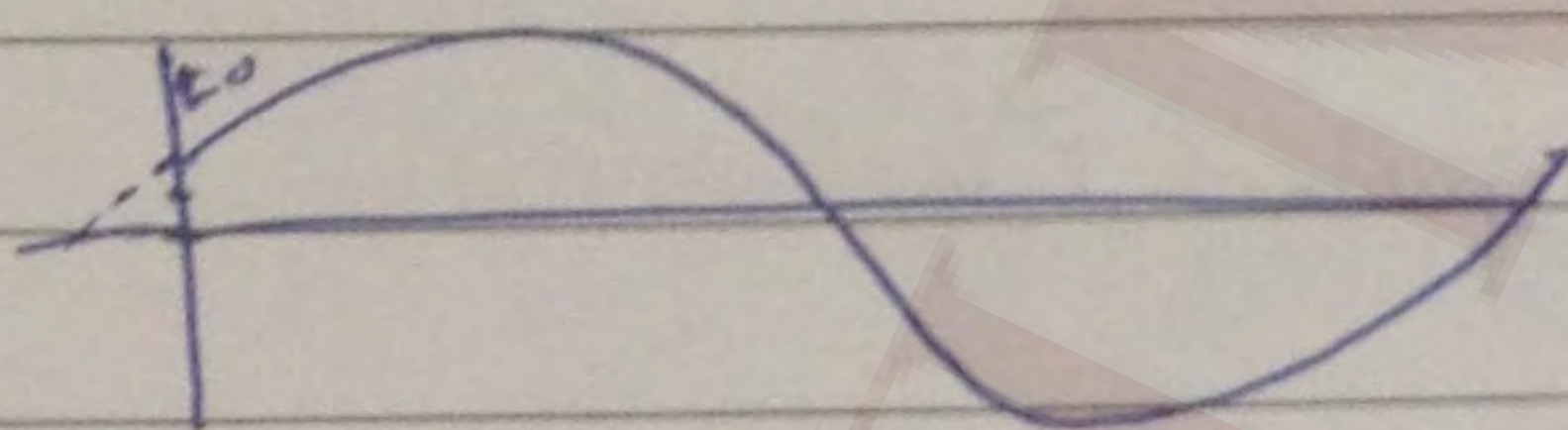
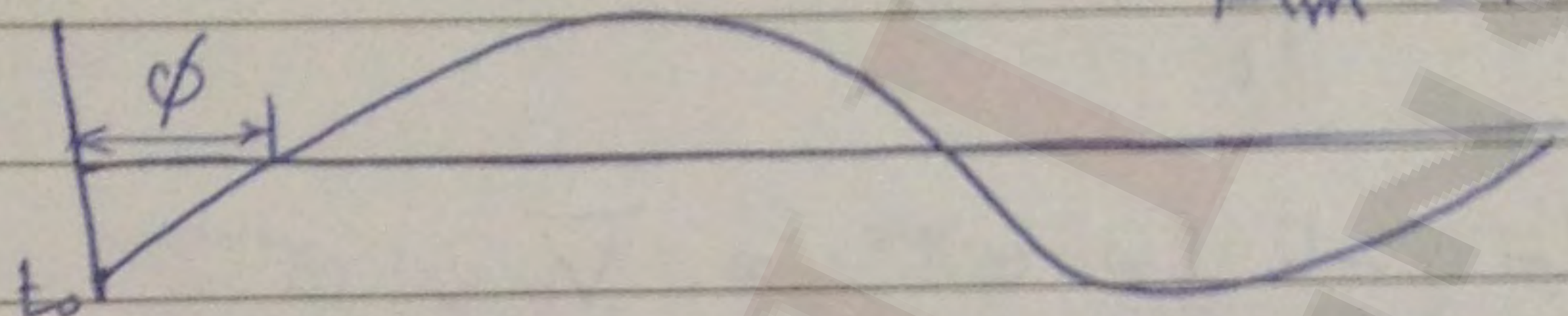
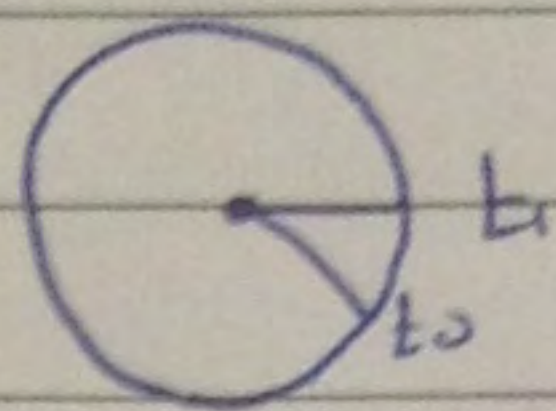
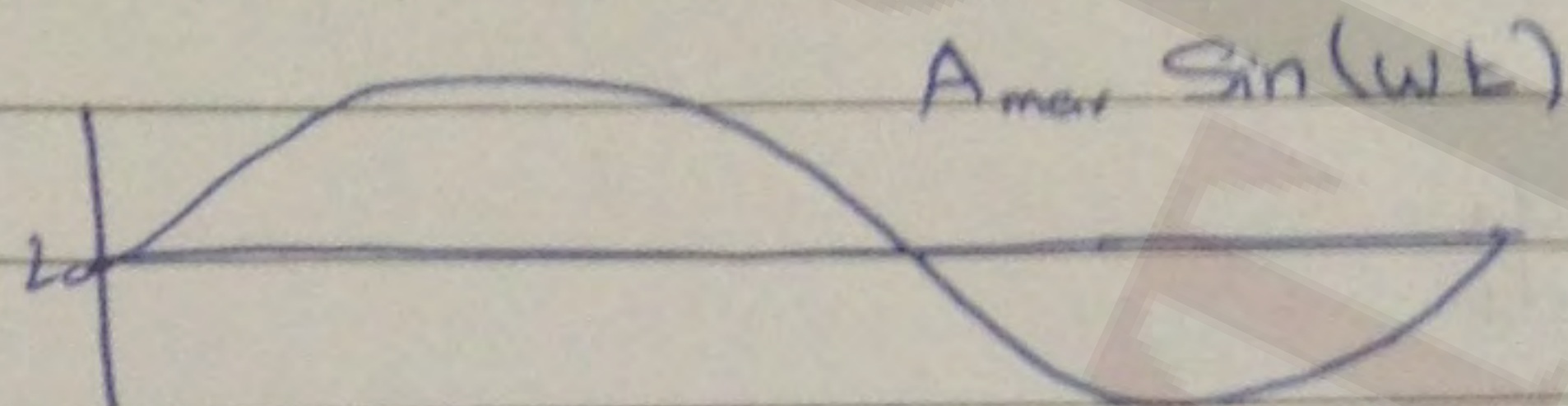
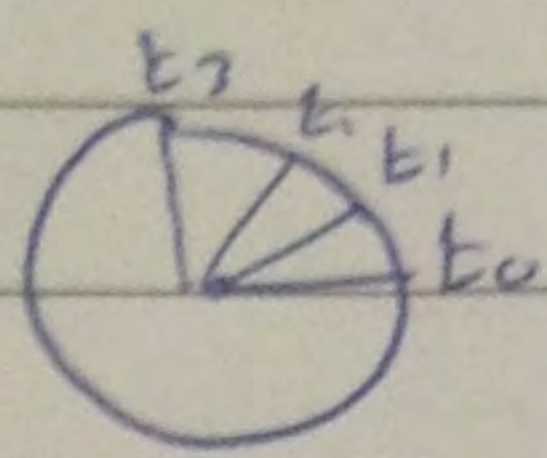
By . Yazan Abawi

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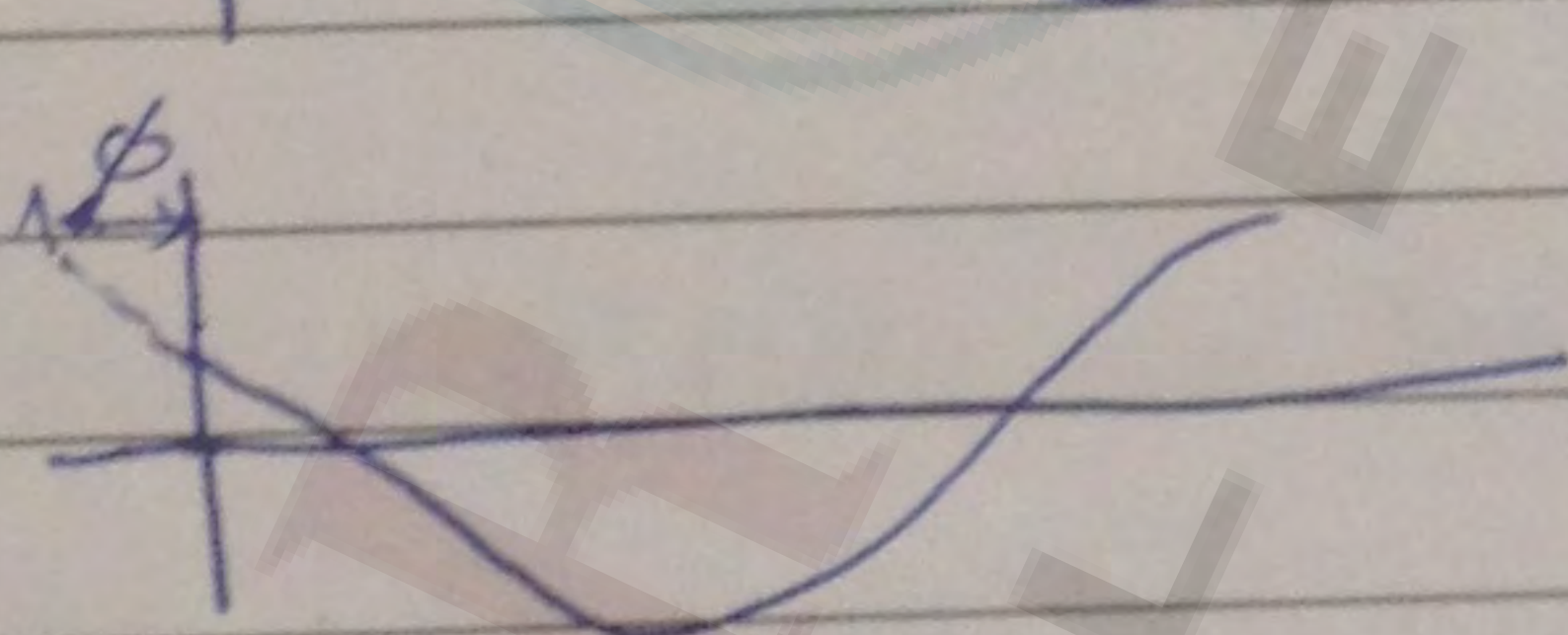
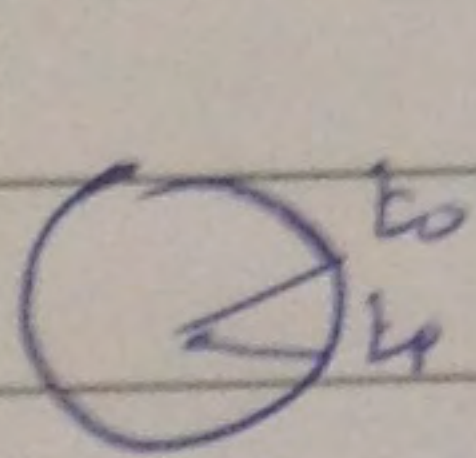
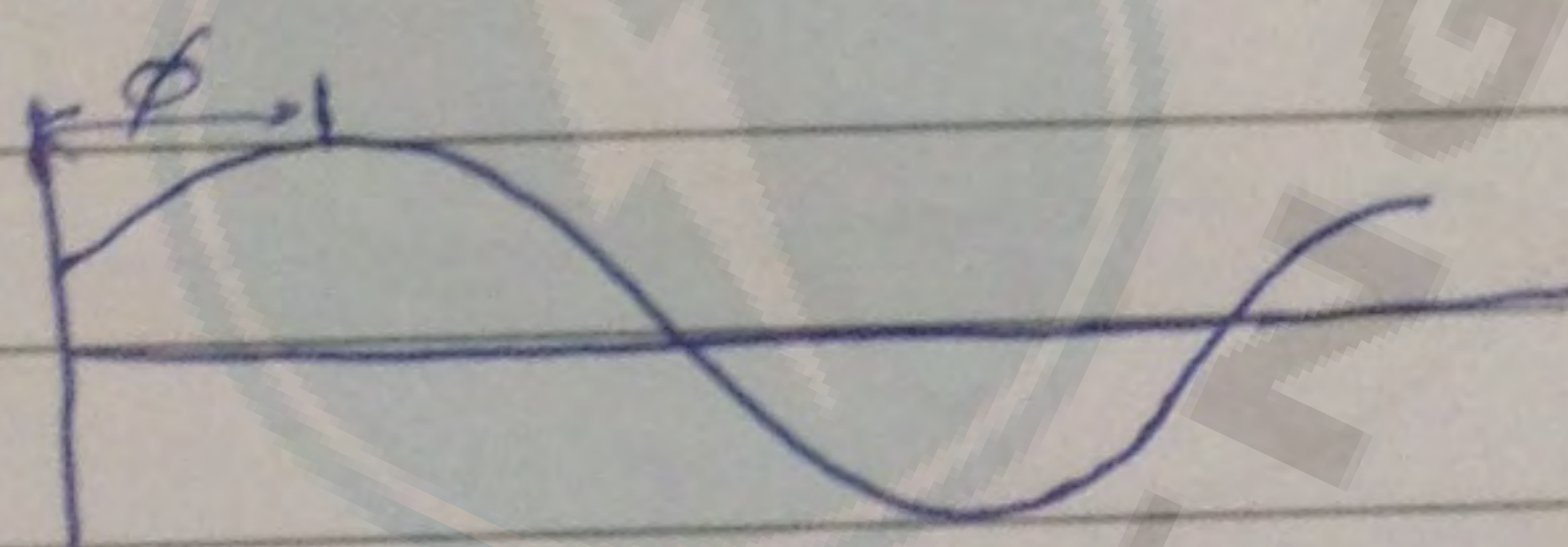
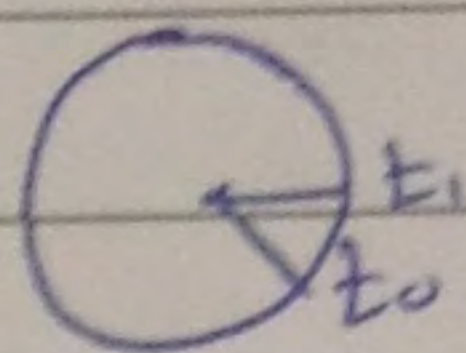
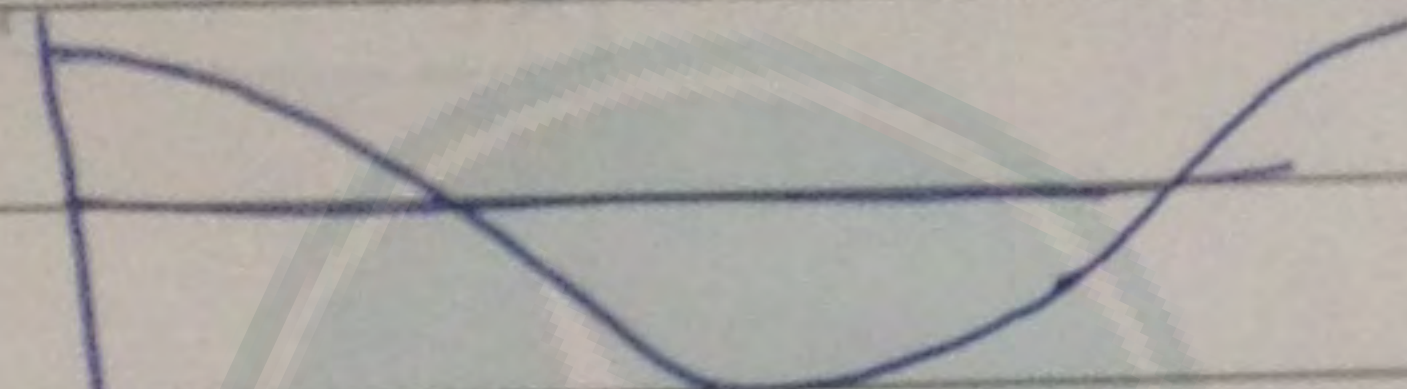
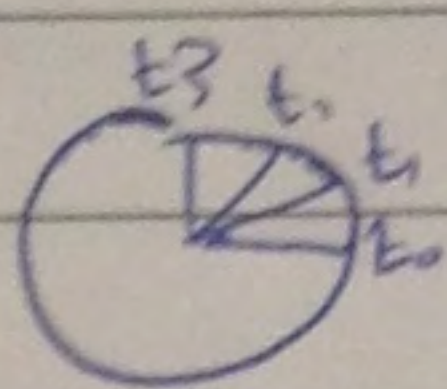
first exam material

Circuits (2)

Sin

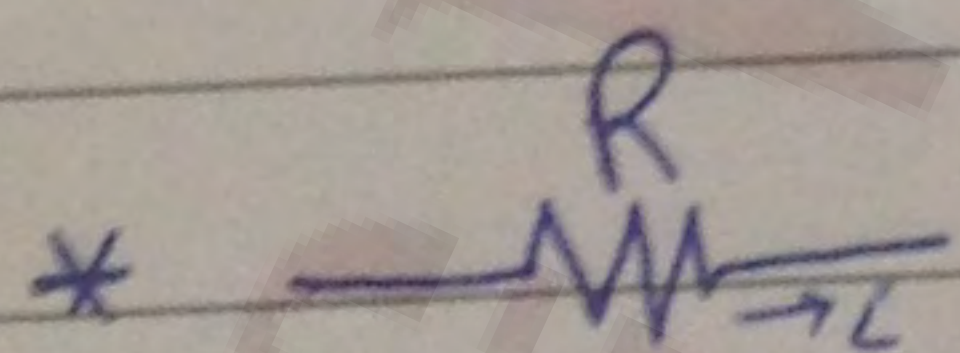


Cos



$$A(t) = A \cos(\omega t + \phi) + j A \sin(\omega t + \phi)$$

$$= A_m e^{j(\omega t + \phi)} = \underline{A} e^{j\phi} e^{j\omega t} = \underline{A} e^{j\omega t}$$



$$i(t) = I_{max} \cos(\omega t + \phi)$$

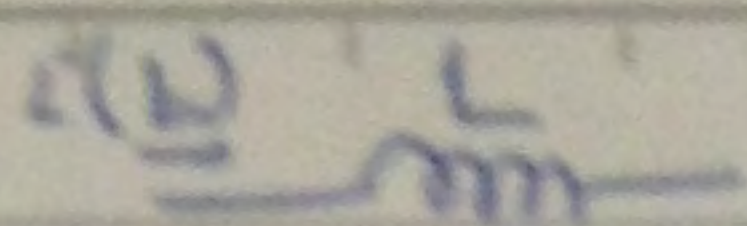
$$V = R i$$

$$V(t) = R I_{max} \cos(\omega t + \phi) = V_{max} \cos(\omega t + \phi)$$

$$i(t) = I_{max} \cos(\omega t + \phi)$$

$$i(t) = I_{max} \cos(\omega t + \phi) + j I_{max} \sin(\omega t + \phi)$$

$$= I_{max} e^{j\phi} e^{j\omega t} = I_{max} e^{j\omega t} = I_{max} \angle \phi$$



$$v = L \frac{di}{dt}$$

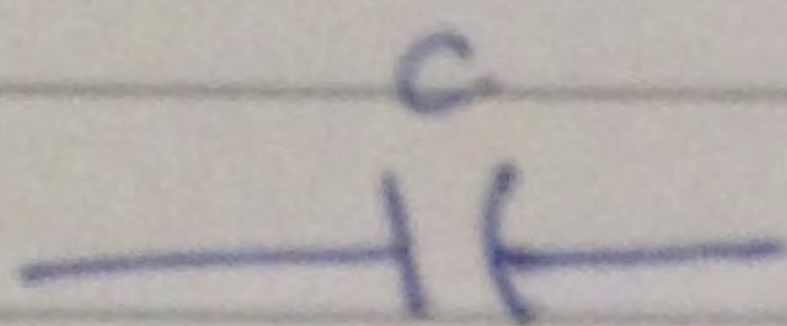
$$v(t) = L \frac{d}{dt} \vec{I}_m e^{j\omega t}$$

$$v(t) = \underline{j\omega L} \vec{I}_m e^{j\omega t} = \vec{V}_{max} e^{j\omega t}$$

$$\therefore \vec{V}_{max} = j\omega L \vec{I}_{max}$$

$$\vec{Z}_L = \frac{\vec{V}_{max}}{\vec{I}_{max}} = \frac{j\omega L \vec{I}_{max}}{\vec{I}_{max}} \Rightarrow \boxed{Z_L = j\omega L}$$

$$\boxed{X_L = \omega L}$$



$$i = C \frac{dv}{dt}$$

$$v(t) = V_{max} \cos(\omega t)$$

$$\vec{v}(t) = \vec{V}_{max} e^{j\omega t}$$

$$i = C \frac{d}{dt} \vec{V}_{max} e^{j\omega t}$$

$$\therefore \vec{I}_{max} = j\omega C \vec{V}_{max}$$

$$\boxed{Z_C = \frac{1}{j\omega C}}$$

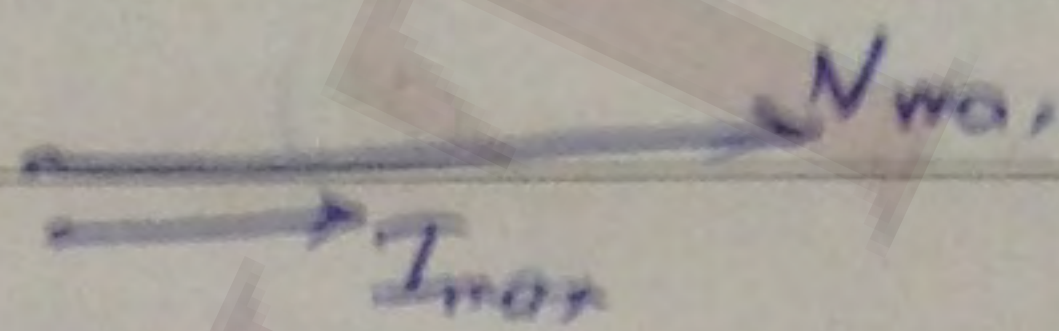
$$\boxed{X_C = \frac{1}{\omega C}}$$

$$\Rightarrow v(t) = V_{max} \cos(\omega t + \phi_v) \xrightarrow{\text{Phase}} \vec{V}_{max} = |V_{max}| e^{j\phi_v}$$

$$i(t) = I_{max} \cos(\omega t + \phi_i) \xrightarrow{\text{Phase}} \vec{I}_{max} = |I_{max}| e^{j\phi_i}$$

$$\underline{Z} = \frac{V_{max}}{I_{max}}$$

$$\boxed{a} \quad R = \frac{V_m}{I_m}$$

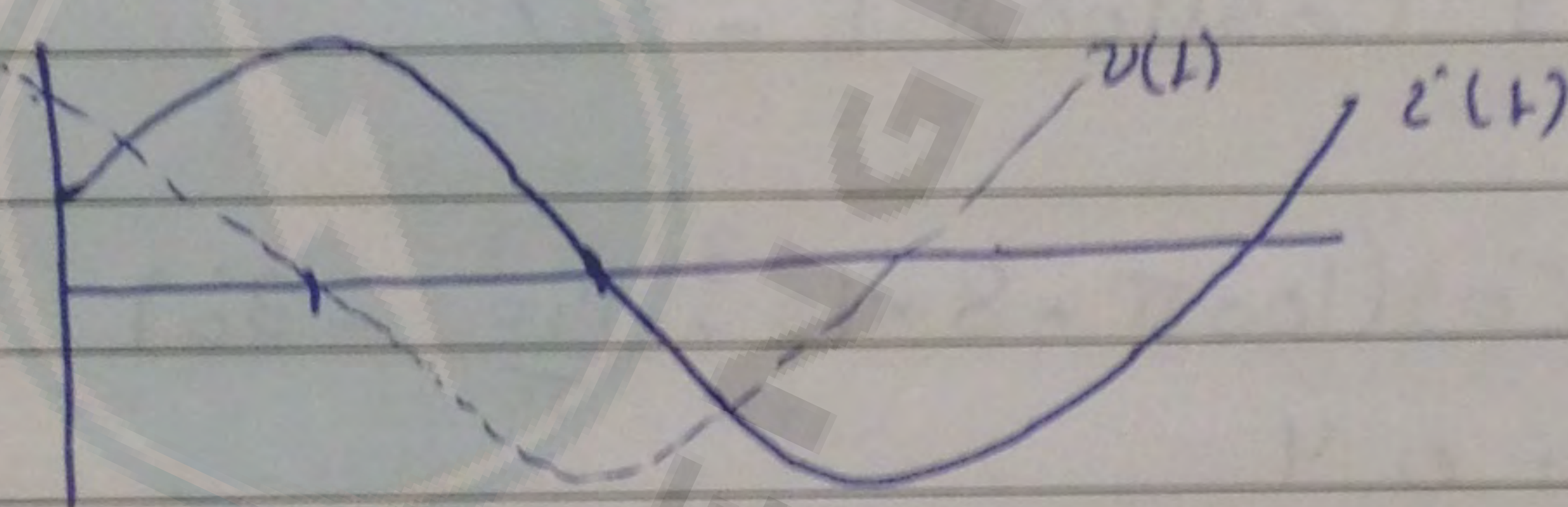


$$\boxed{b} \quad L \rightarrow Z = \frac{V_{max}}{I_{max}} = \frac{|V_m| \angle \phi_v}{|I_m| \angle \phi_i} = Z \angle \phi_v - \phi_i = \boxed{\omega L \angle 90^\circ}$$

$$\boxed{c} \quad C \rightarrow Z = \frac{|V_m| \angle \phi_v}{|I_m| \angle \phi_i} = \frac{1}{j\omega C} = \boxed{Z \angle -90^\circ}$$

exp $v(t) = 100 \cos(\omega t + 30^\circ) = 100 \angle 30^\circ$

$i(t) = 5 \cos(\omega t - 60^\circ) = 5 \angle -60^\circ$

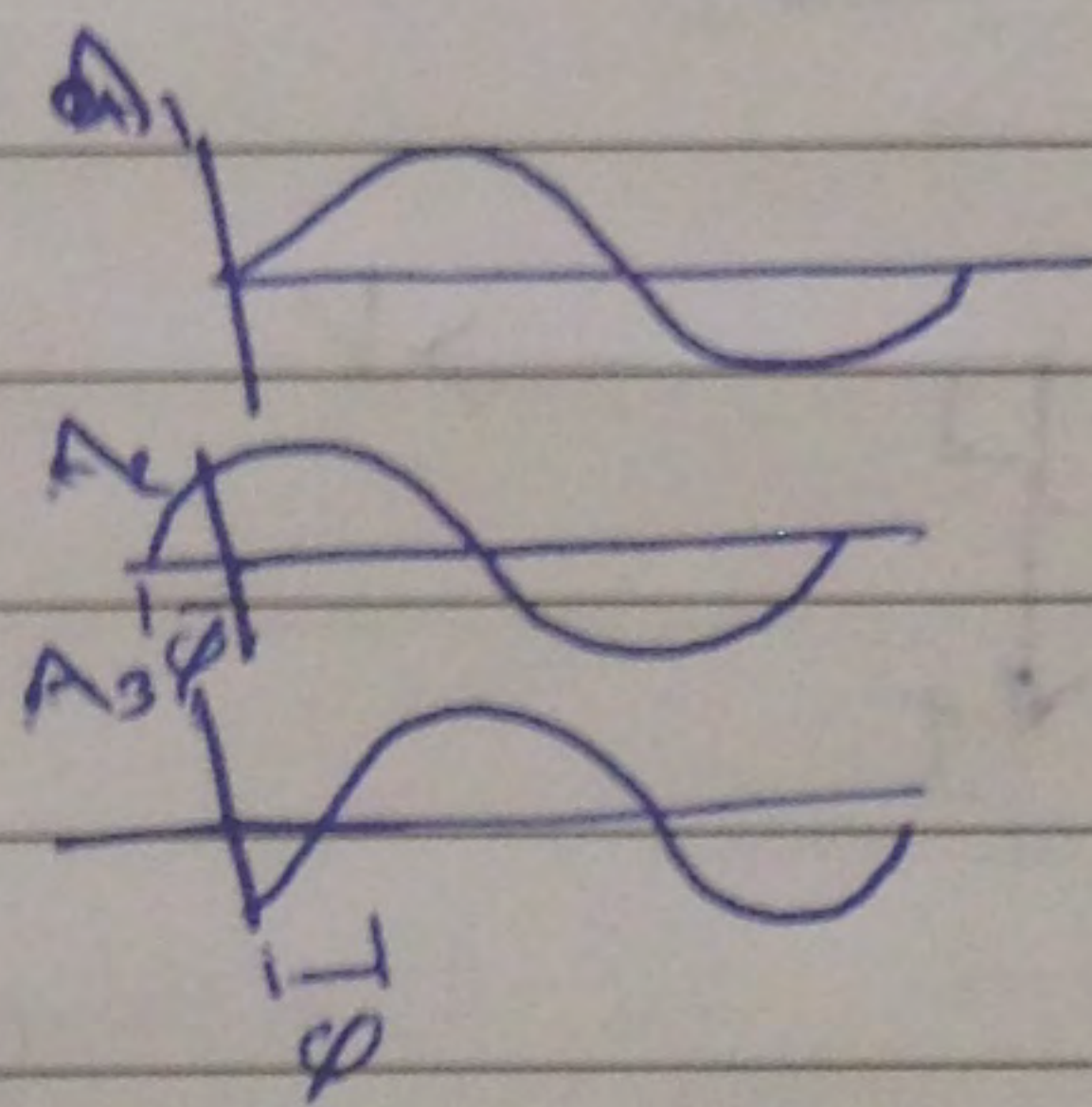


$P(t) = v(t) i(t) \rightarrow$ Instantaneous Power

exp $A_1 = A_m \sin(\omega t)$

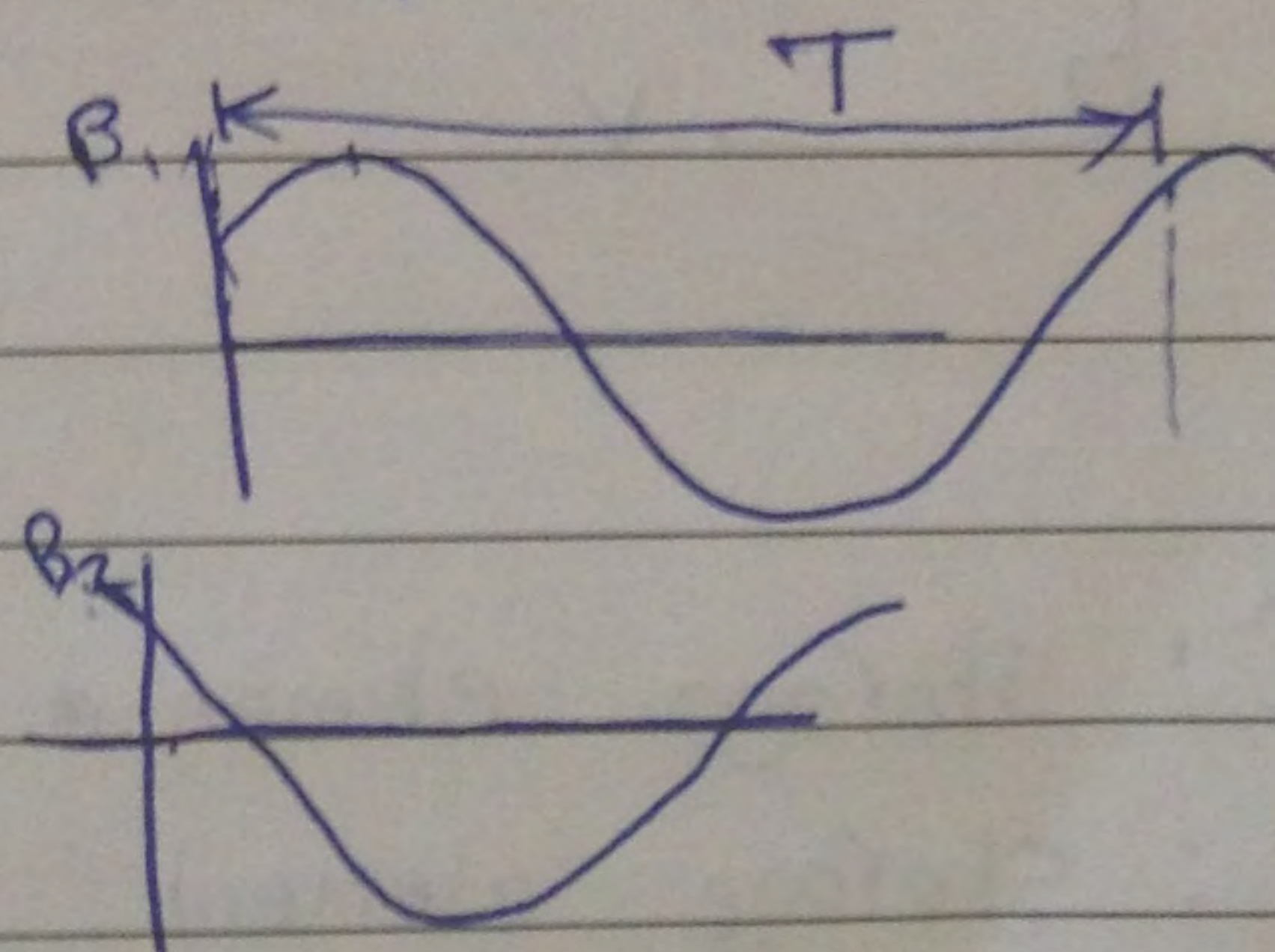
$A_2 = A_m \sin(\omega t + \phi)$

$A_3 = A_m \sin(\omega t - \phi)$



exp $B_1 = B_m \cos(\omega t - \phi)$

$B_2 = B_m \cos(\omega t + \phi)$



$$T = 360^\circ = 2\pi \text{ rad (one Period)}$$

$$\Rightarrow A_1 = B_m \cos(2\pi t - \phi)$$

$$\omega = \frac{2\pi}{T} = 2\pi \rightarrow T = 1 \text{ sec}$$

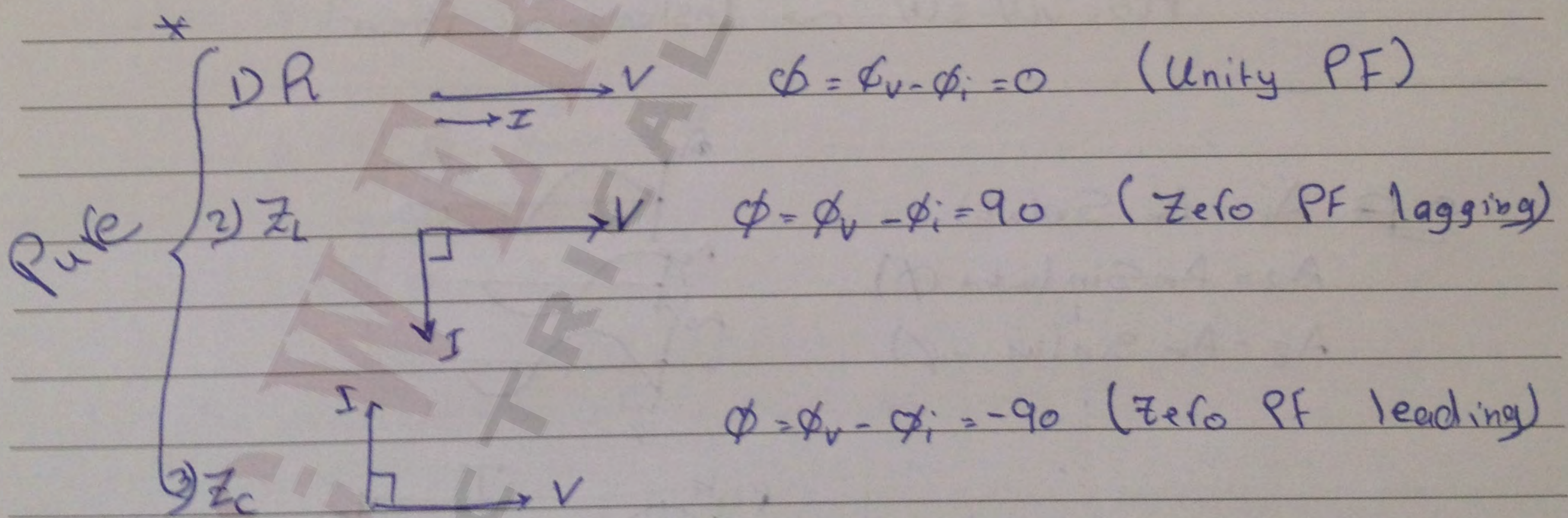
$$\Rightarrow A_2 = B_m \cos(100\pi t - \phi)$$

$$\omega = \frac{2\pi}{100\pi} = \frac{1}{50} \text{ sec} = 20 \text{ msec}$$

$$\Rightarrow A(8\text{ms}) = 1 \cos(100\pi t - 30^\circ)$$

$$= \cos\left(100\pi \times 8 \times 10^{-3} \times \frac{180}{\pi} - 30^\circ\right)$$

$$= -0.4$$

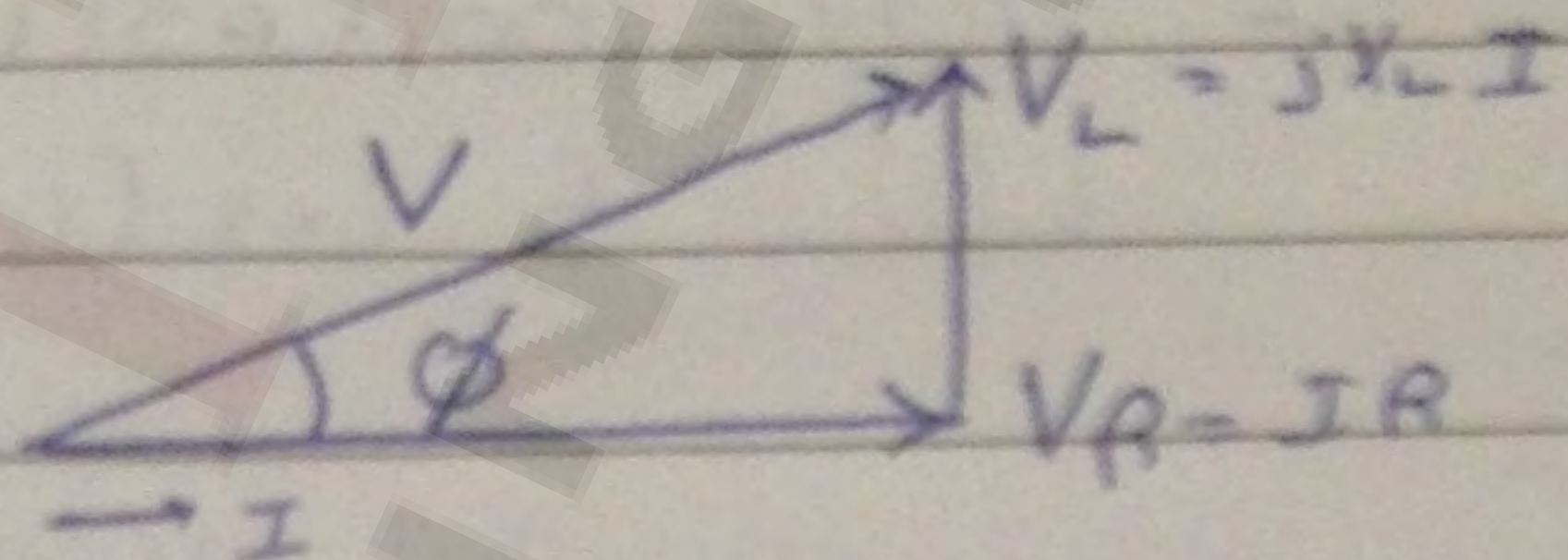
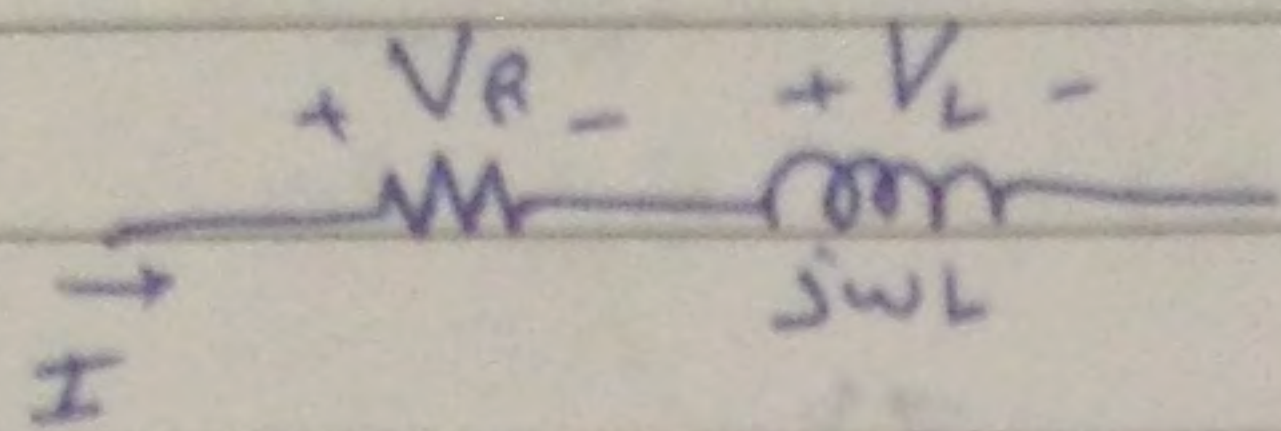


C: Storage element in form of electric field.

L: Storage element in form of magnetic field.

* Practical

1) L: inductive (consists of inductor and R)

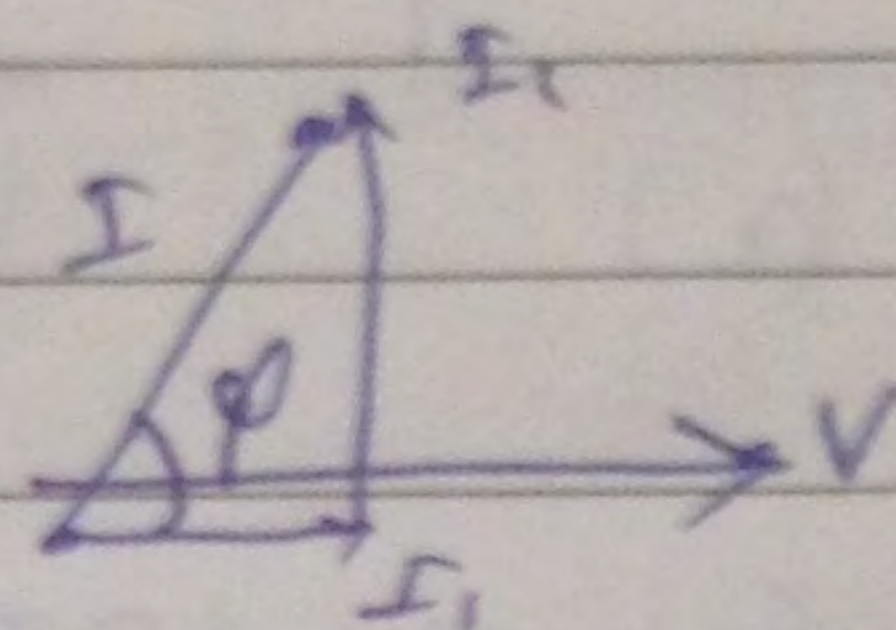
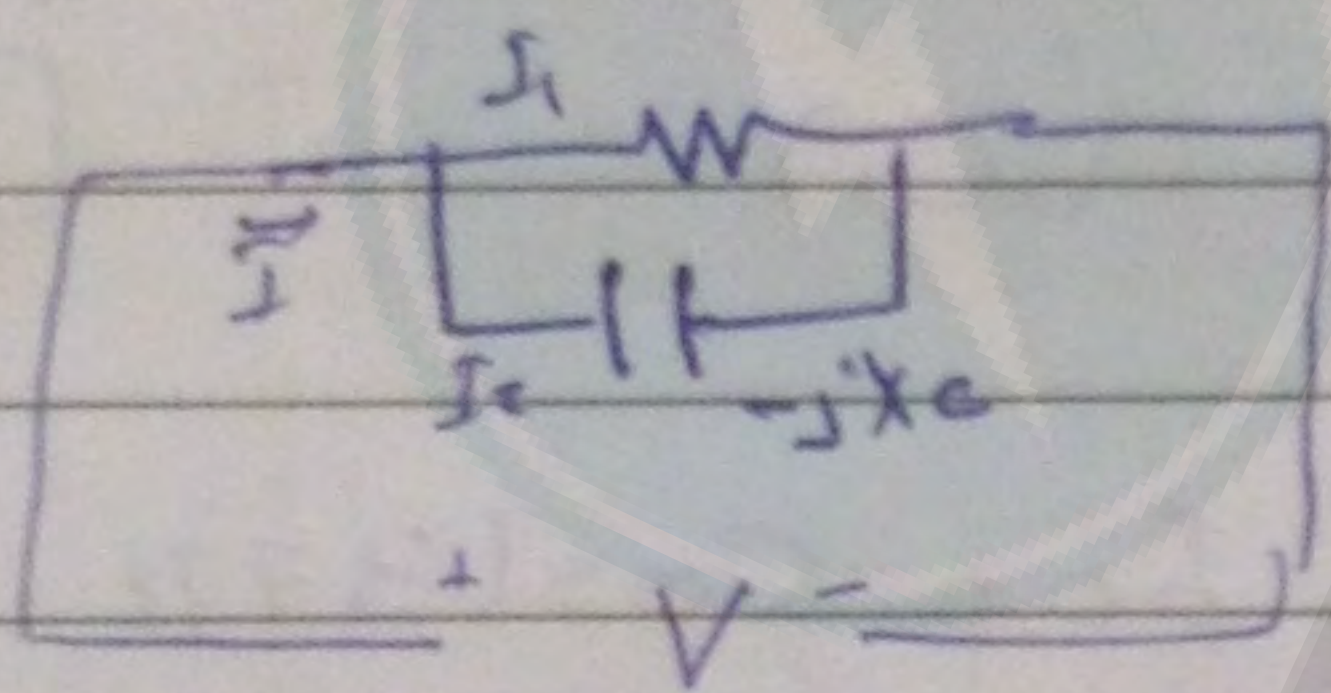


lagging PF

- Once I lags V the ϕ is +ve

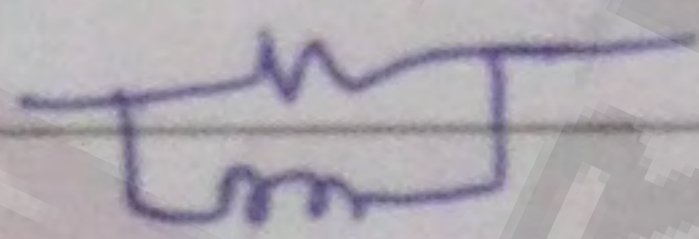
- Once I leads V the ϕ is -ve

2) C: Capacitive

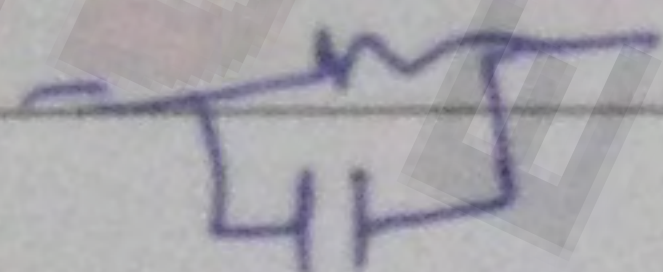
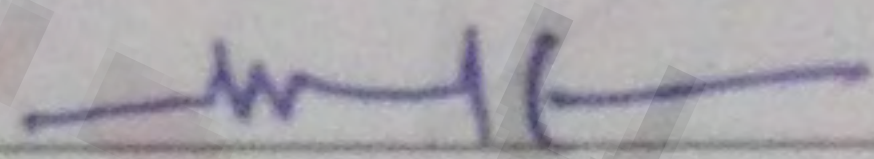


leading PF

* $0 < \phi < 90 \Rightarrow$ inductive

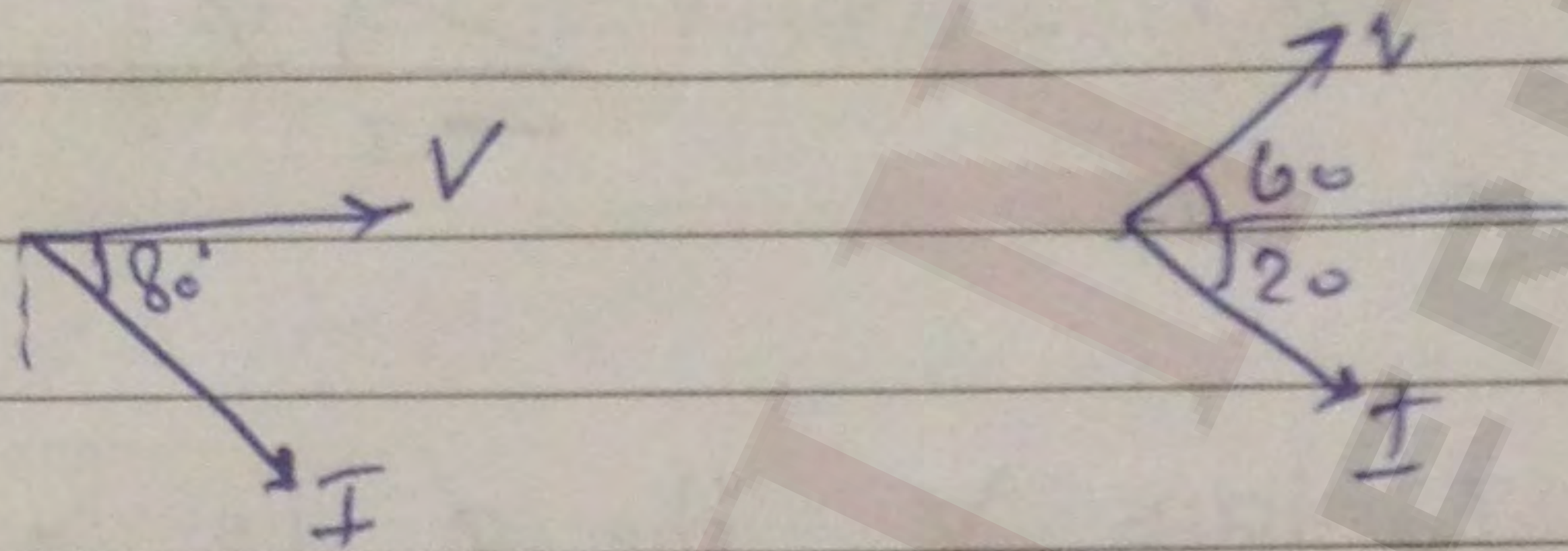


* $90 < \phi < 0 \Rightarrow$ Capacitive

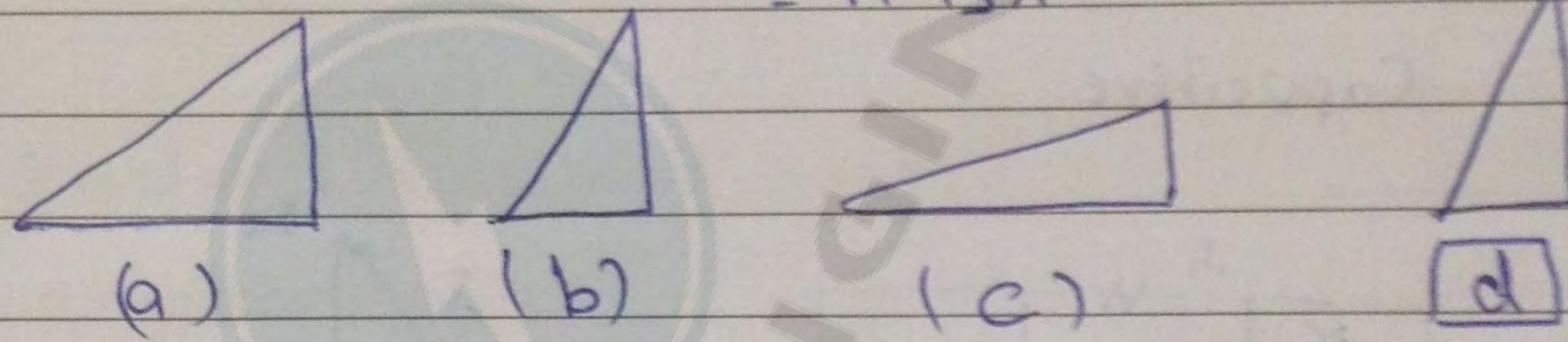


exp $v(t) = V_m \cos(\omega t + \phi_v)$
 $i(t) = I_m \cos(\omega t + \phi_i)$

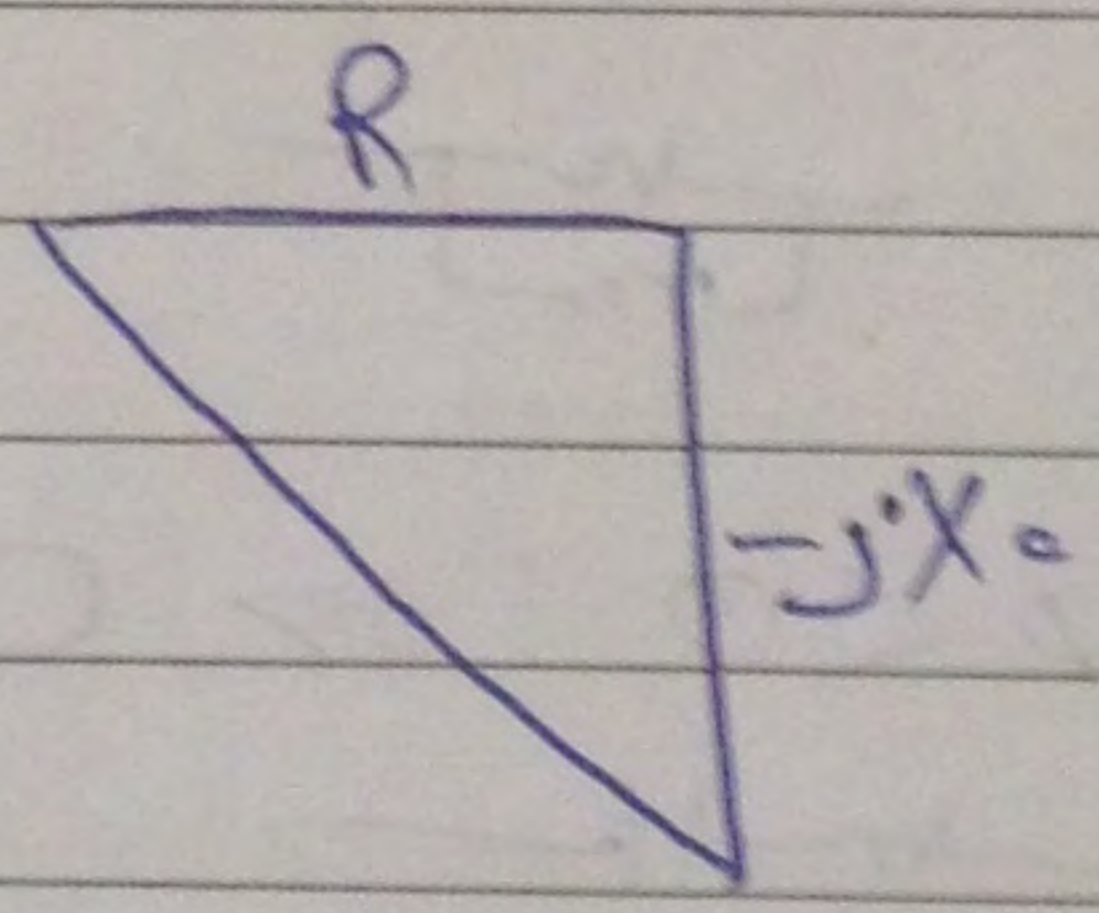
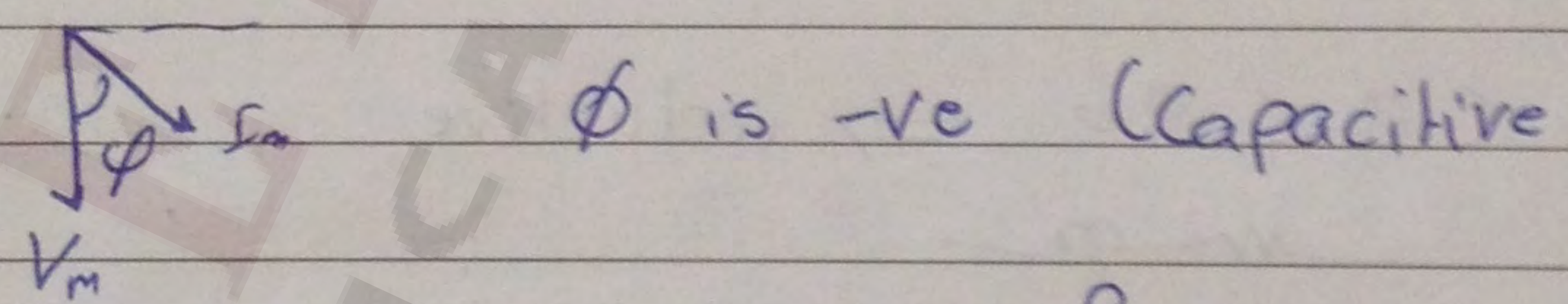
- $v(t) = 100 \cos(100\pi t + 60) = 100 \angle 60$
 $i(t) = 5 \cos(100\pi t - 20) = 5 \angle -20$

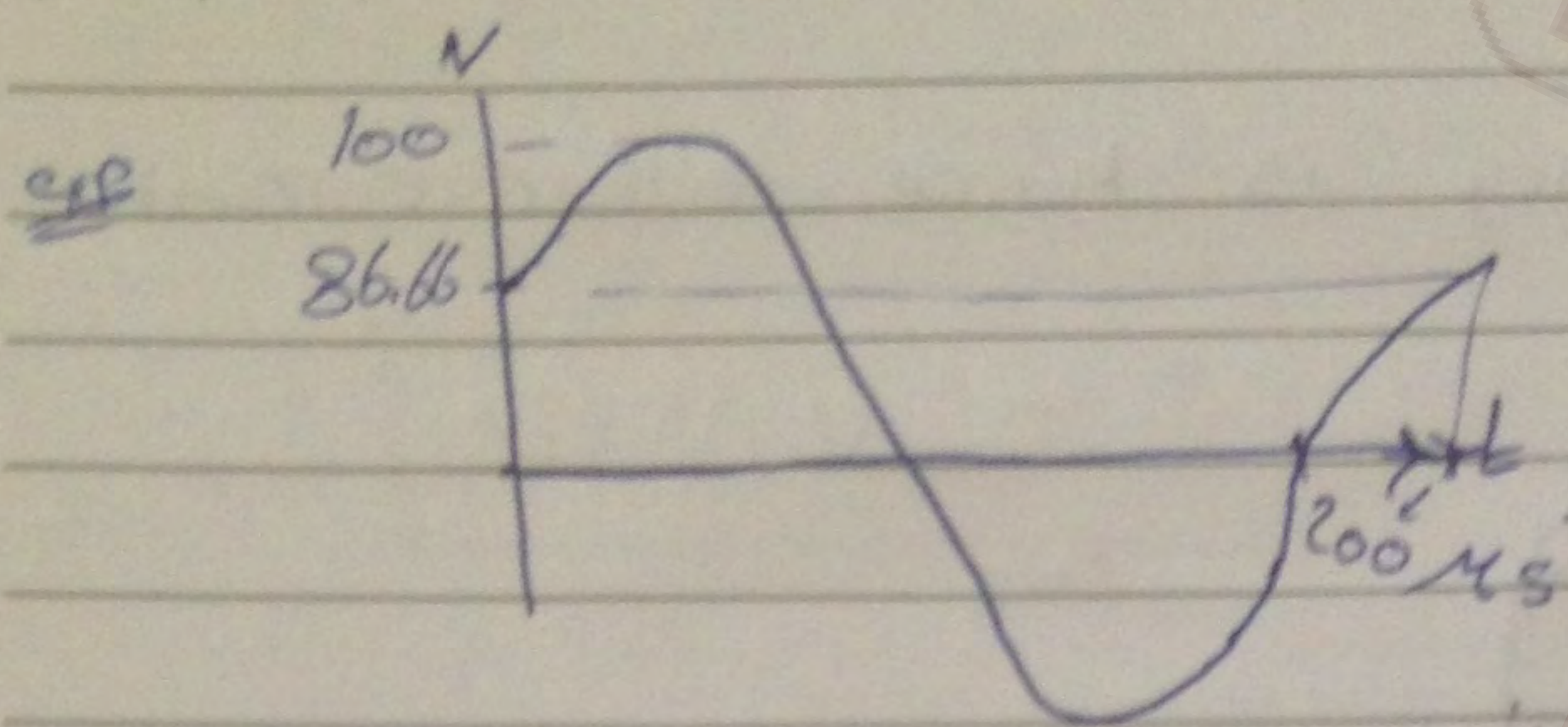


$Z = \frac{100 \angle 60 + 20}{5} = 20 \angle 80$ - impedance triangle
 $= R + jX$



- $v(t) = 500 \sin(377t) = 500 \cos(377t - 90) = 500 \angle -90$
 $i(t) = 10 \cos(377t - 30) = 10 \angle -30$





$$T = \frac{2\pi}{200} = \frac{\pi}{100}$$

$$v(t) = 100 \cos(\omega t - 30^\circ) \rightarrow v(0) = 86.6 = 100 \cos \phi \rightarrow \phi = 30^\circ$$

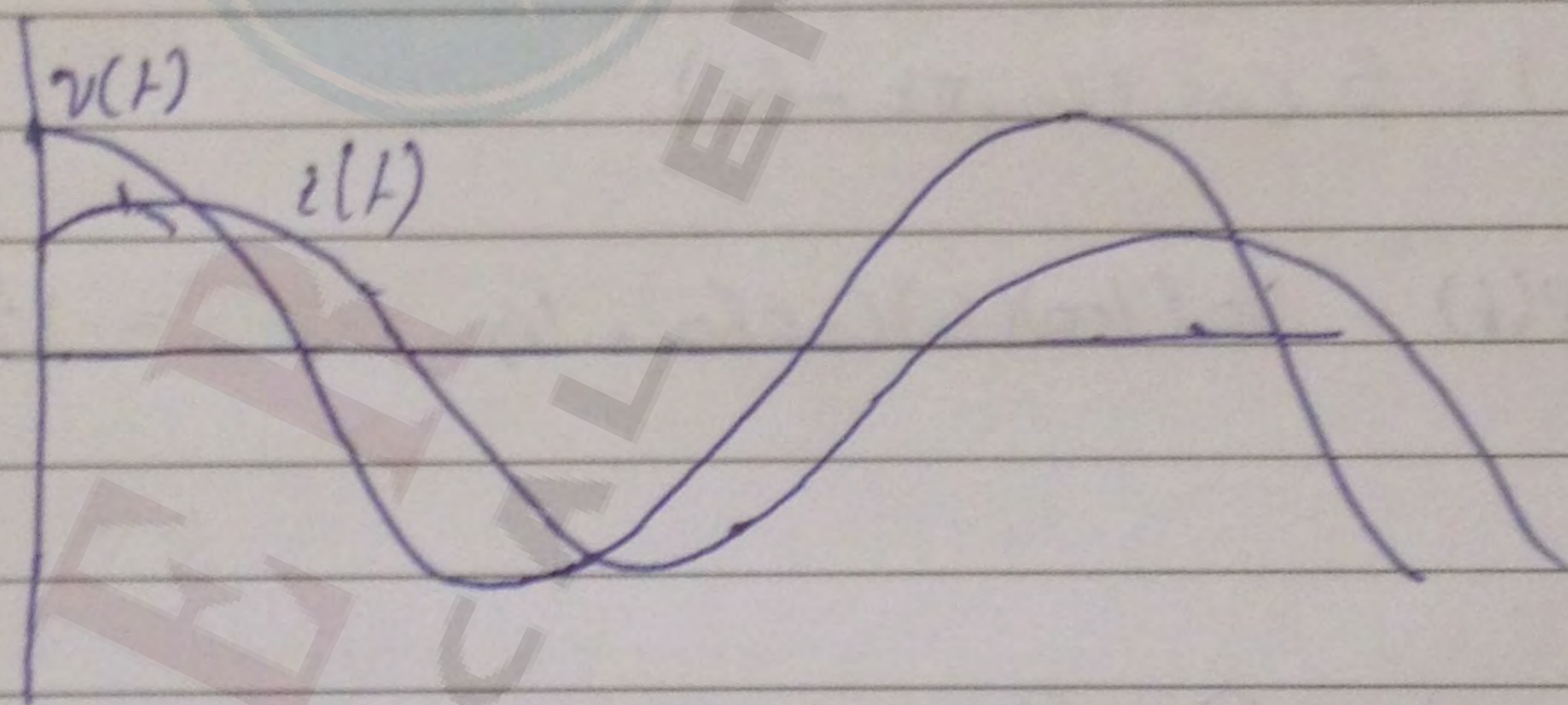
$$v(t) = 100 \sin(10\pi t + \phi) \rightarrow v(0) = 86.6 = 100 \sin \phi \rightarrow \phi = 60^\circ$$

$$= 100 \sin(10\pi t + 60^\circ)$$

exp $v(t) = V_m \cos(\omega t)$

$$i(t) = I_m \cos(\omega t - \phi)$$

lag



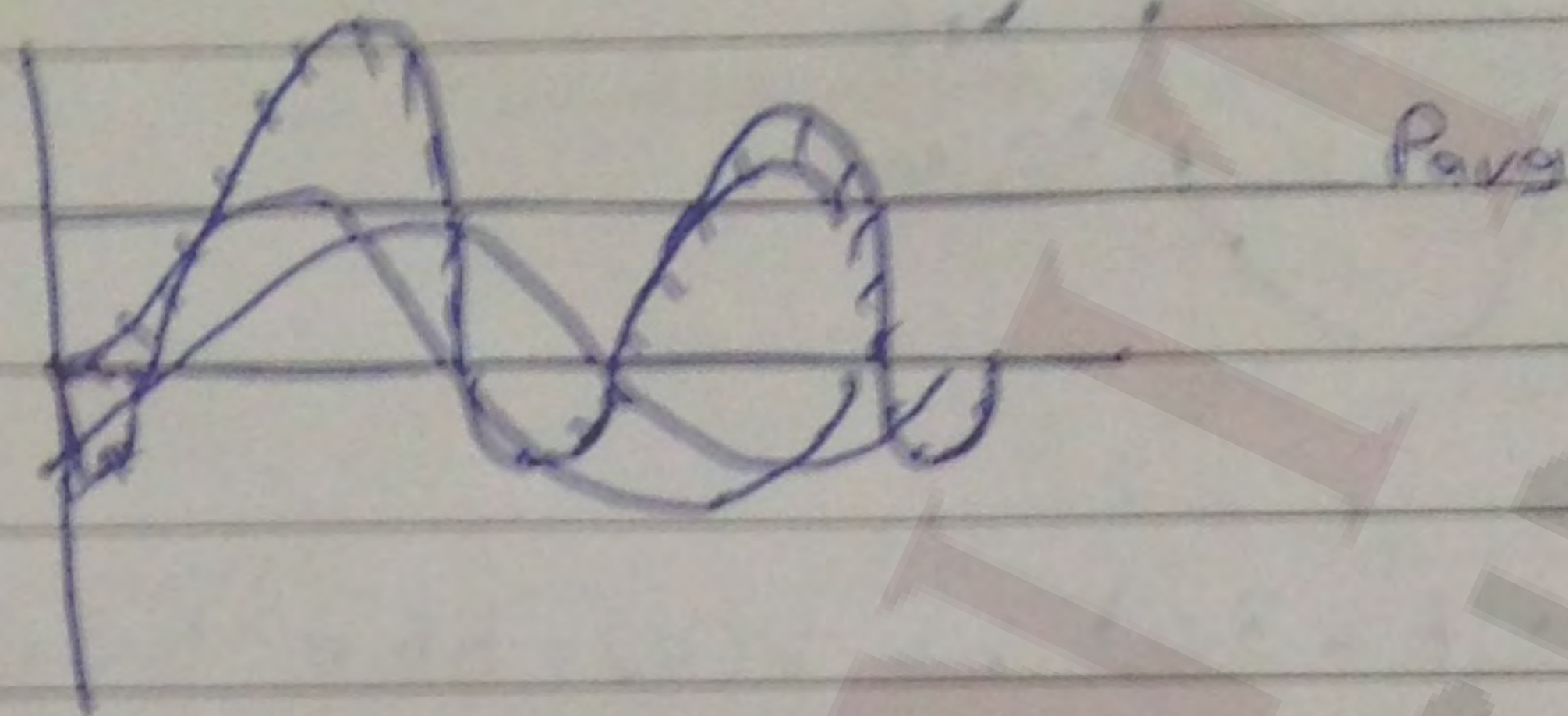
$$P(t) = \frac{1}{2} V_m I_m [\cos \phi + \cos(2\omega t - \phi)]$$

$$= \frac{1}{2} V_m I_m \cos \phi + \frac{1}{2} V_m I_m \cos(2\omega t - \phi)$$

$$= \frac{1}{2} V_m I_m \cos \phi + \frac{1}{2} V_m I_m \cos \phi + \frac{1}{2} V_m I_m \sin 2\omega t \sin \phi$$

$$\cos 2\omega t$$

⇒ Instant. Power frequency is twice either current or voltage freq.



$$P_{avg} = P_{max} - \frac{P_{max} - P_{min}}{2}$$

$$P_{avg} = P_{min} + \frac{P_{max} - P_{min}}{2}$$

exp $v(t) = 100 \cos 100\pi t$

$i(t) = 5 \cos (100\pi t - 60)$

$$P(t) = \frac{1}{2}(100)(5) \cos 60 + \frac{1}{2} \dots = 125$$

$$P(t) = P + P \cos 2\omega t + Q \sin 2\omega t$$

$$P = \frac{1}{2} V_m I_m \cos(\phi_v - \phi_i)$$

$$Q = \frac{1}{2} V_m I_m \sin(\phi_v - \phi_i)$$

$$\vec{S} = P + jQ \quad (\text{Complex Power})$$

$$= \frac{1}{2} V_m I_m (\cos \phi + j \sin \phi)$$

$$= \frac{1}{2} V_m I_m e^{j\phi} = \frac{1}{2} V_m I_m e^{j(\phi_v - \phi_i)}$$

$$S = \frac{1}{2} V_m e^{j\phi_v} \cdot I_m e^{-j\phi_i}$$

$$\boxed{\bar{S} = \frac{1}{2} \bar{V}_{max} \bar{I}_{max}^*}$$

$$\boxed{|S| = AP = \frac{1}{2} |V_m| |I_m|}$$

$p(t) \rightarrow$ Watt

$P_{avg} \rightarrow$ W

$Q \rightarrow$ VAR

$AP \rightarrow$ VA

$$\underline{V}_{max} = 100 \angle 0$$

$$\underline{I}_{max} = 5 \angle -60$$

$$\underline{S} = \frac{1}{2} V_m I_m^* = \frac{1}{2} (100 \angle 0) (5 \angle 60) = 250 \cos 60 + j 250 \sin 60$$

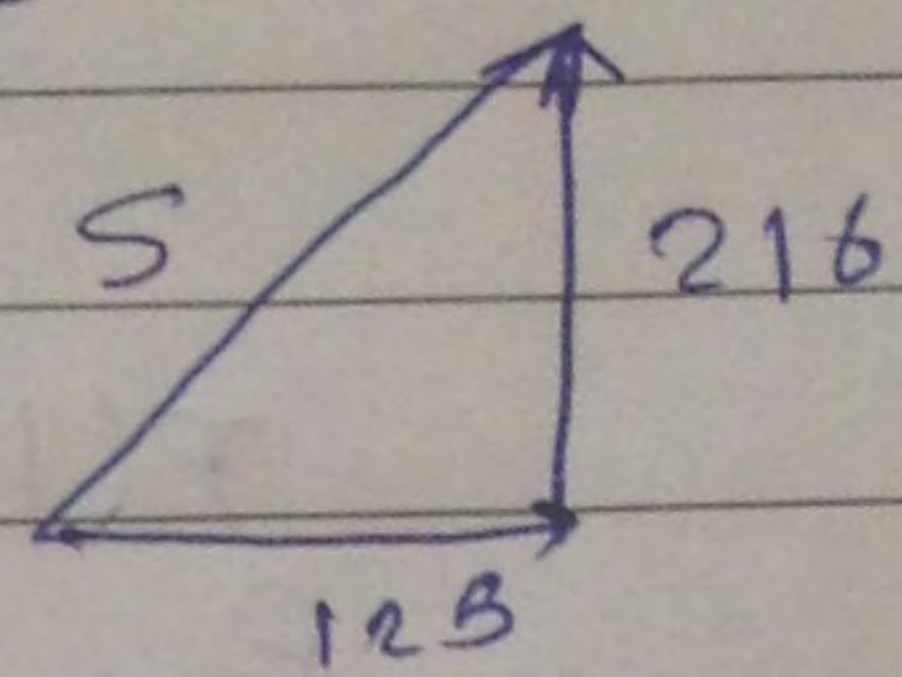
$$= 125 + j 250 * 0.866$$

$$P_{avg} = 125 \text{ W}$$

$$Q = 216 \text{ VAR}$$

$$|S| = 250$$

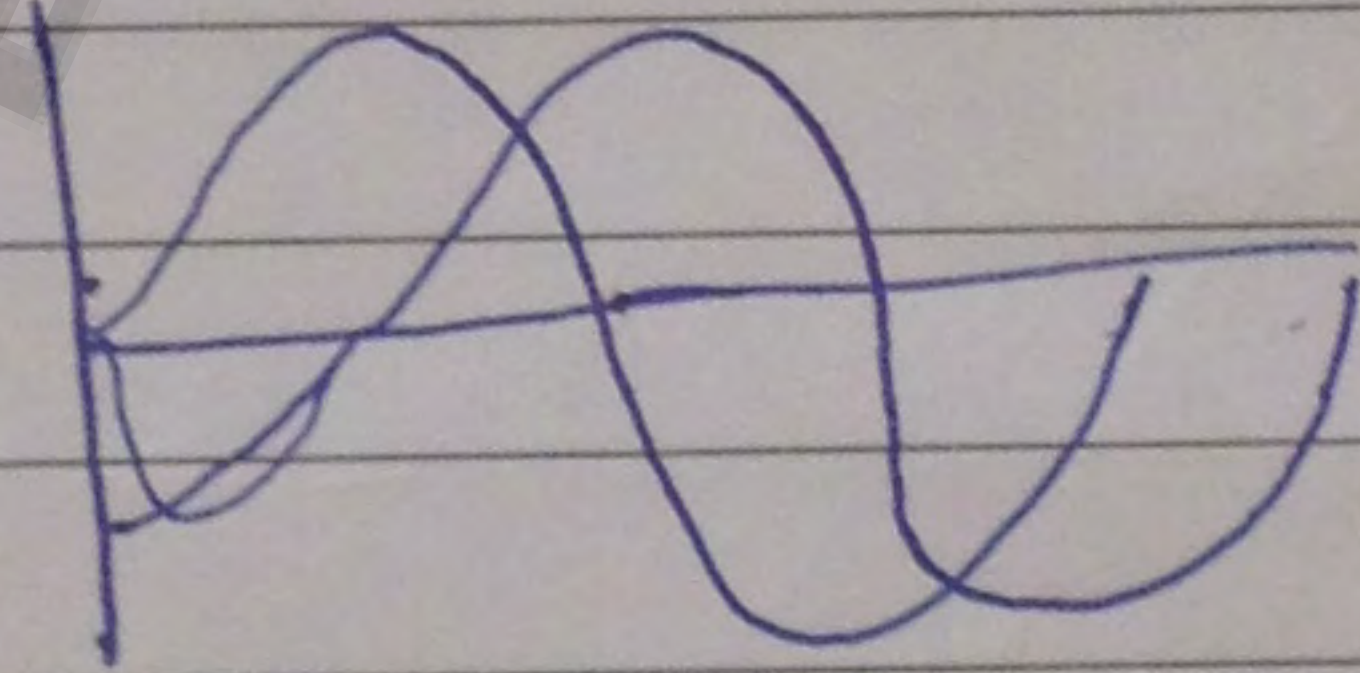
Power triangle



$$P(t) = 125 + 125 \cos 200\pi t + 216 \sin 200\pi t$$

$$= 125 + 250 \cos(200\pi t - 60)$$

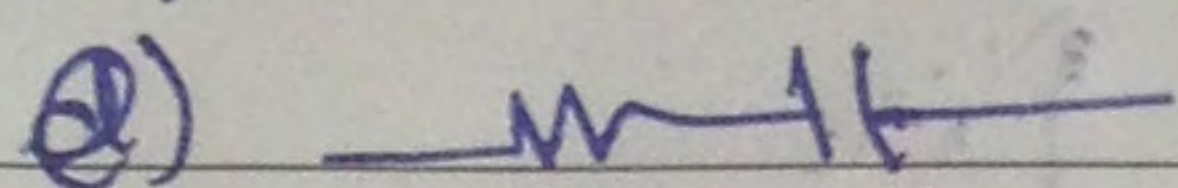
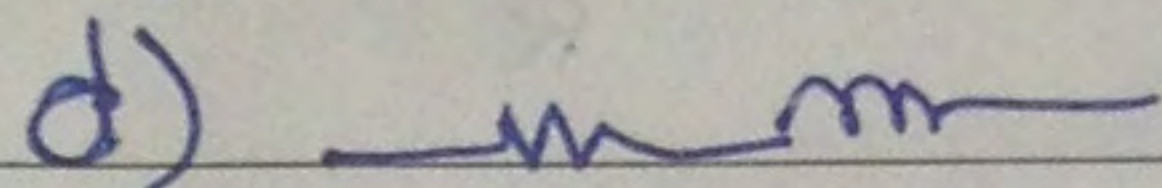
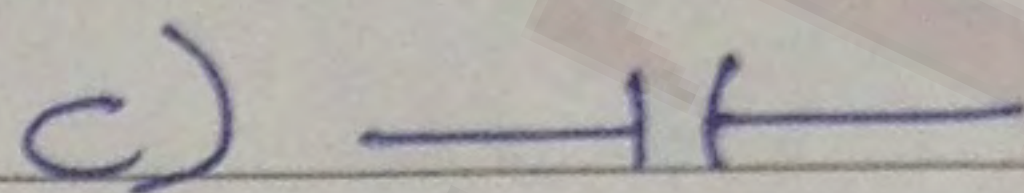
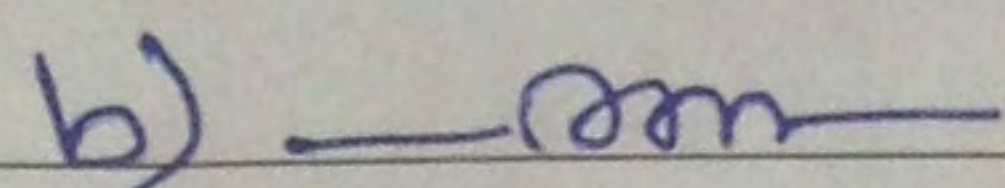
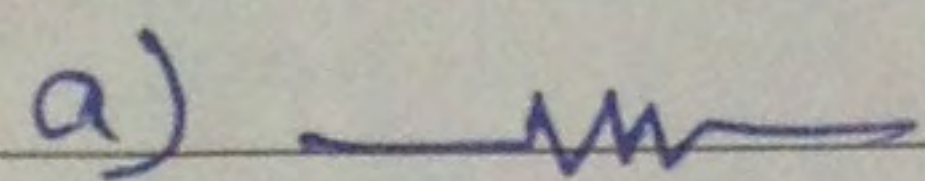
$$P_{avg L} = 0$$



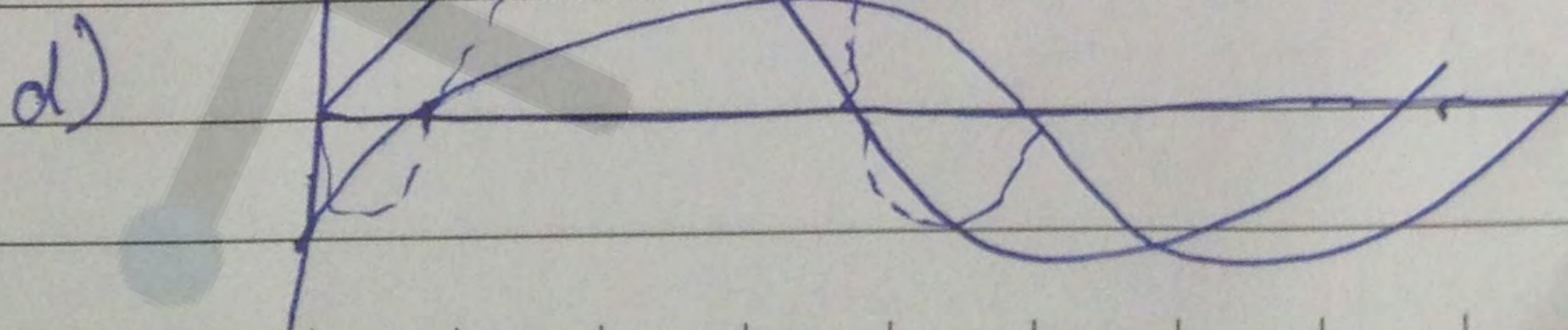
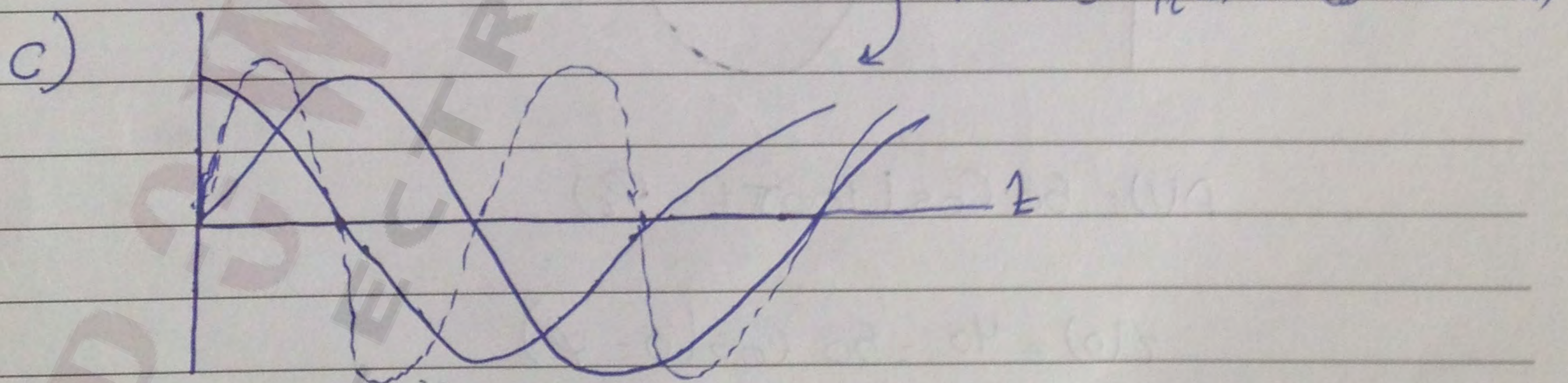
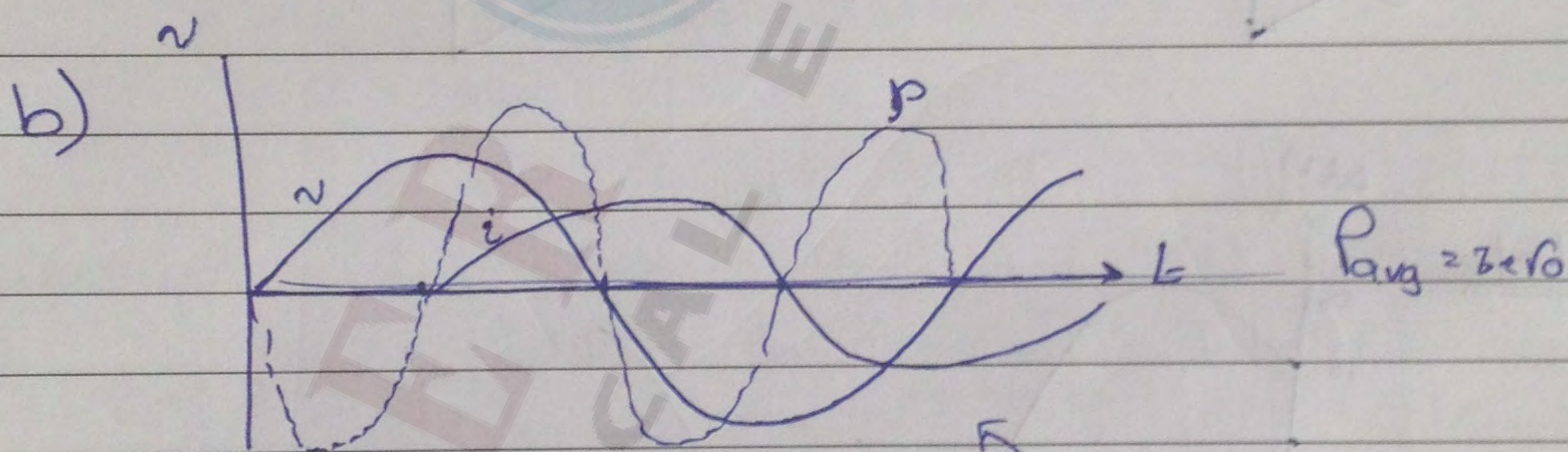
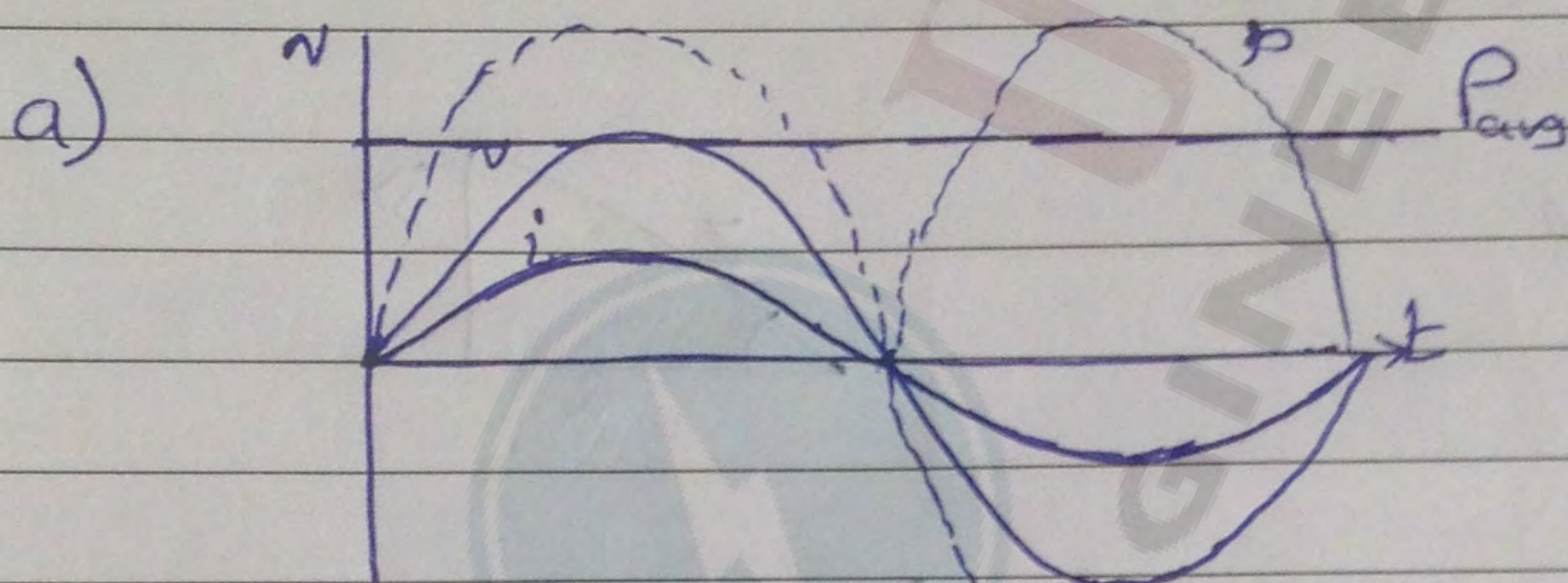
$$P(t) = 0 + 250 \underbrace{\cos(200\pi t + 90)}_{= \phi} = P_{max}$$

$$= -1 \quad P_{min}$$

ex Draw the current, Voltage and instantaneous Power in the following elements:-

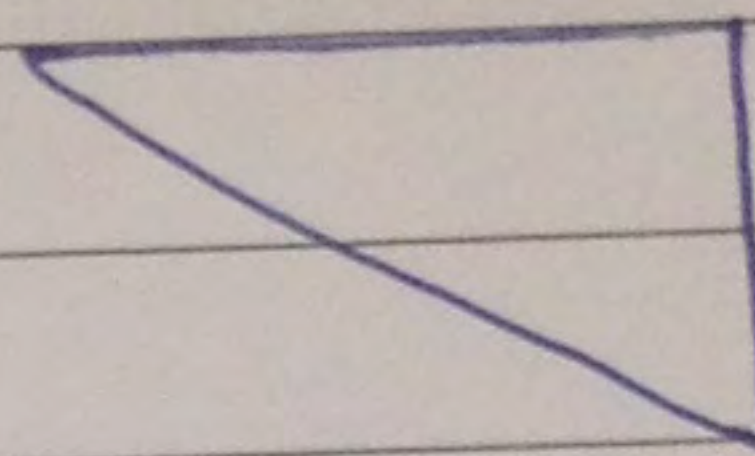
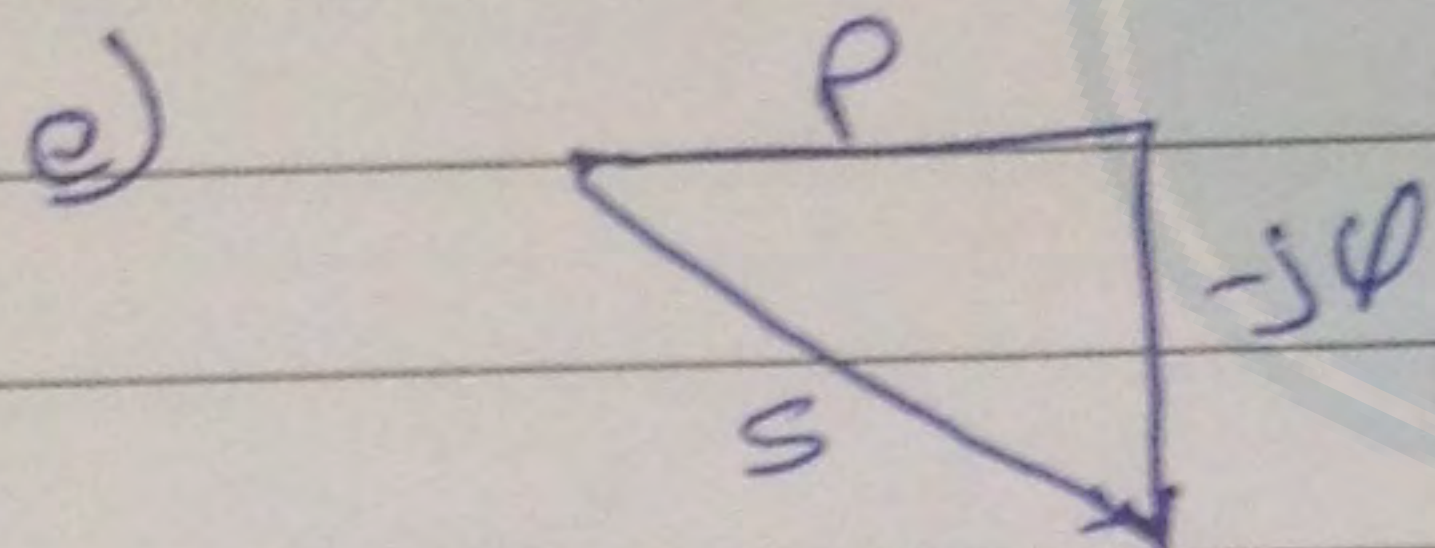
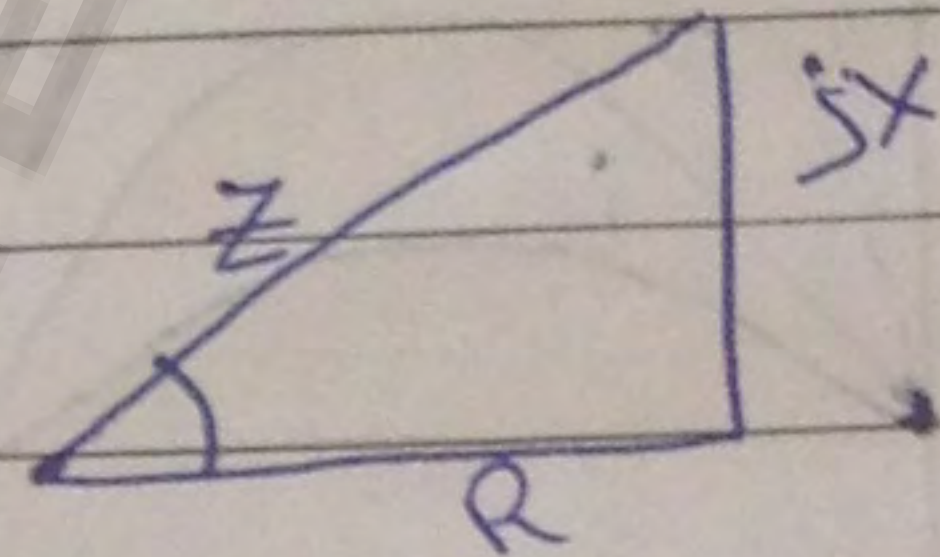
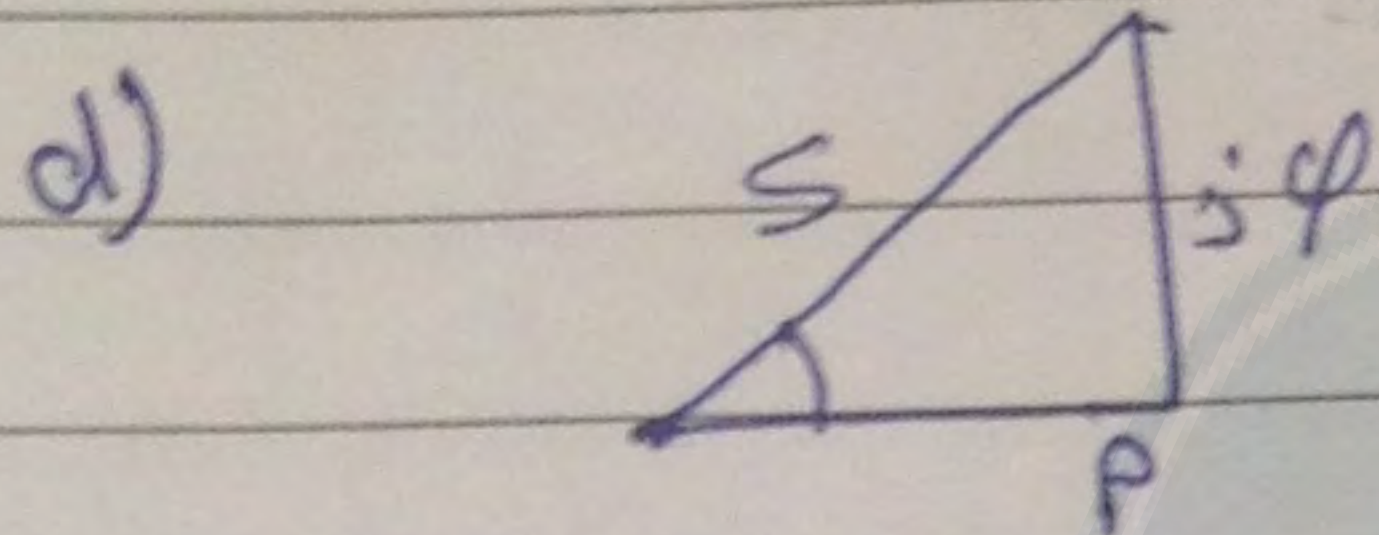
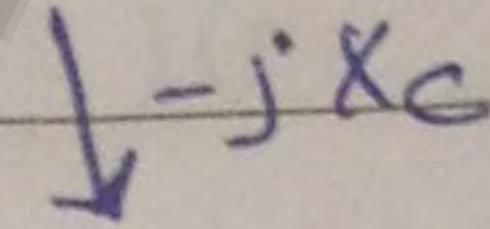
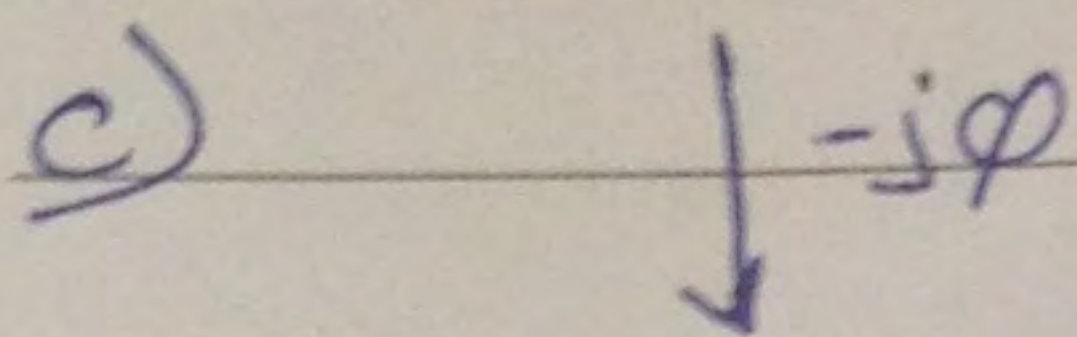
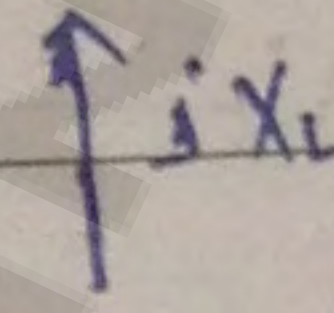
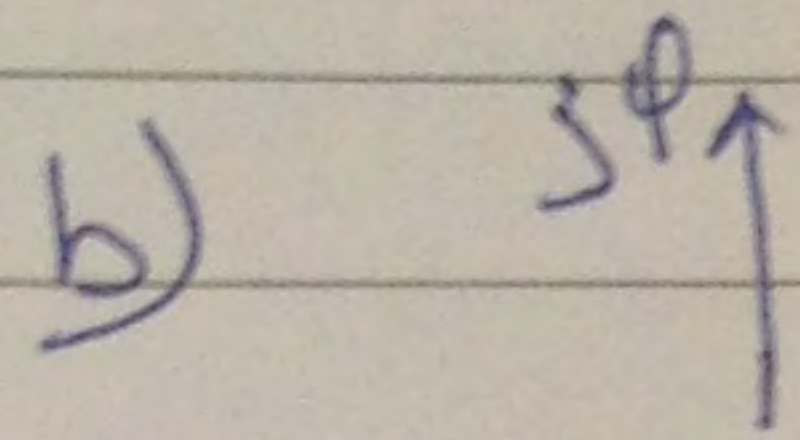
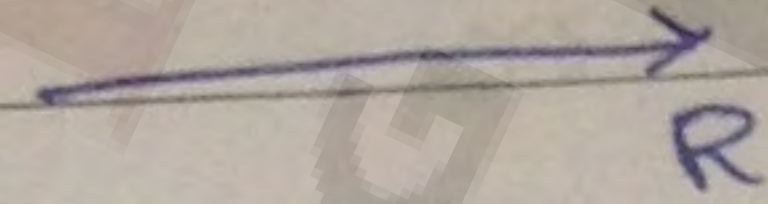
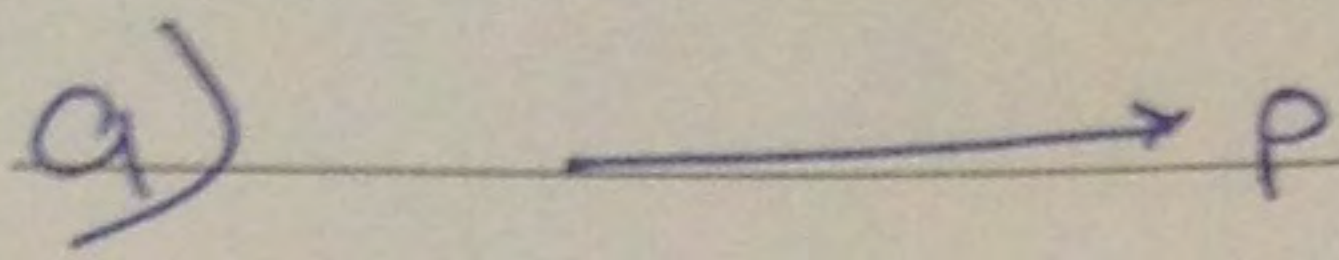


Suppose $v(t) = V_m \sin \omega t$

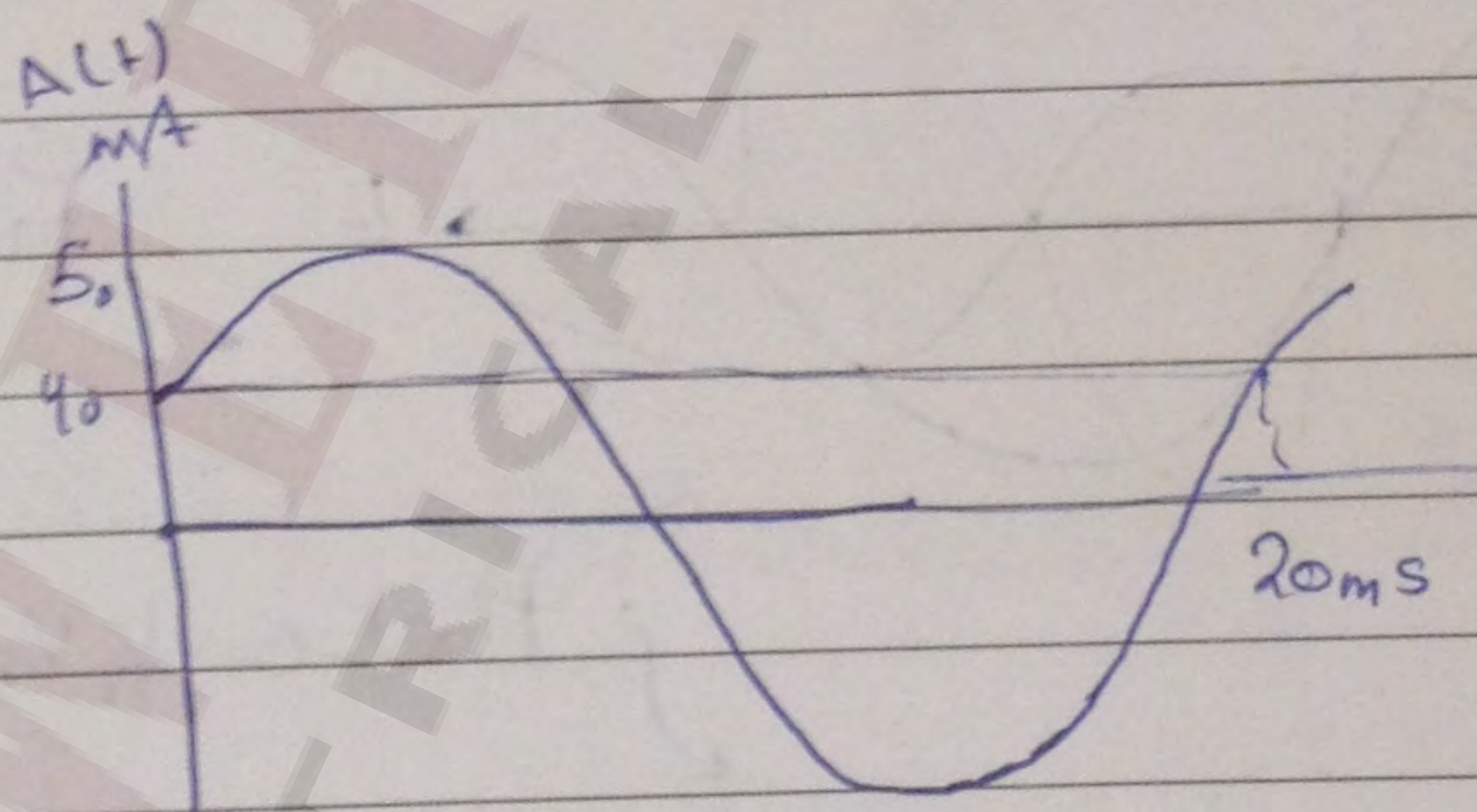


Power triangle

Imp. Triangle



Note

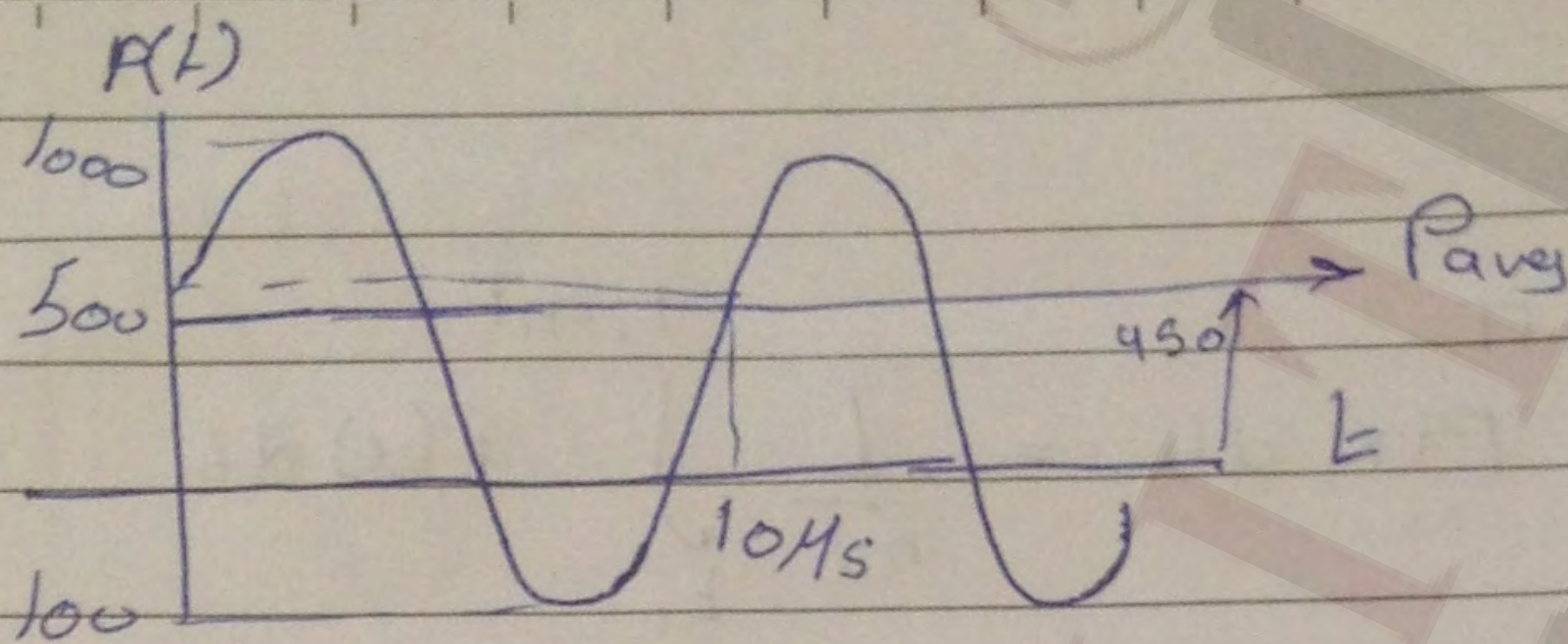


$$A(t) = 50 \cos(100\pi t - \phi)$$

$$i(0) = 40 = 50 \cos(0 - \phi)$$

$$\phi = \cos^{-1} 0.8 = 36.87^\circ$$

Exam question



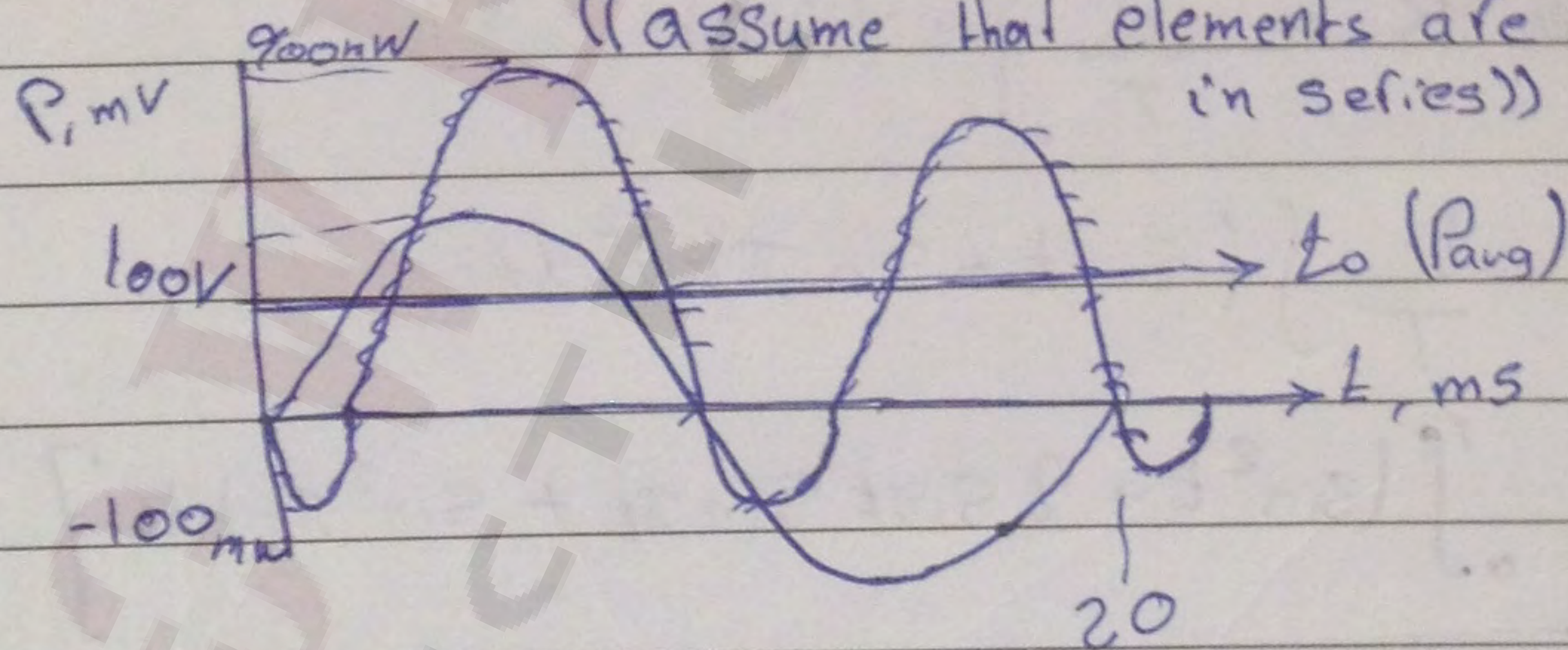
$$P_{av} = P_{max} - \frac{P_{max} - P_{min}}{2}$$
$$= 1000 - \frac{1000 - 100}{2} = 1000 - 550 = 450$$

$$P(t) = P_{max} \cos(\omega t - \phi_0)$$

$$500 = P(0) = 550 \cos(\omega t - \phi)$$

$$\phi = \cos^{-1} \frac{500}{550} = 84^\circ$$

→ Suppose that given Voltage and $P(t)$ find Current
(assume that elements are connected ② P_{avg}
in series)



③ Complex P

④ R, L or RC

$$P_{DC} = I^2 R$$

$$P_{DC} = \frac{1}{T} \int_{t_1}^{t_1+T} P(x) dt = \frac{1}{nT} \int_{t_1}^{t_1+nT} P(x) dt$$

$$= \lim_{n \rightarrow \infty} \frac{1}{nT} \int_{-\frac{nT}{2}}^{\frac{nT}{2}} P(t) dt$$

$$P_{DC} = I^2 R = \frac{1}{T} \int_0^T R i^2(t) dt$$

$$I = \sqrt{\frac{1}{T} \int_0^T i^2(t) dt} \quad \text{effective current (RMS)}$$

$$P = I_{eff}^2 R$$

$$\Rightarrow i(t) = \sin t + \sin \pi t$$

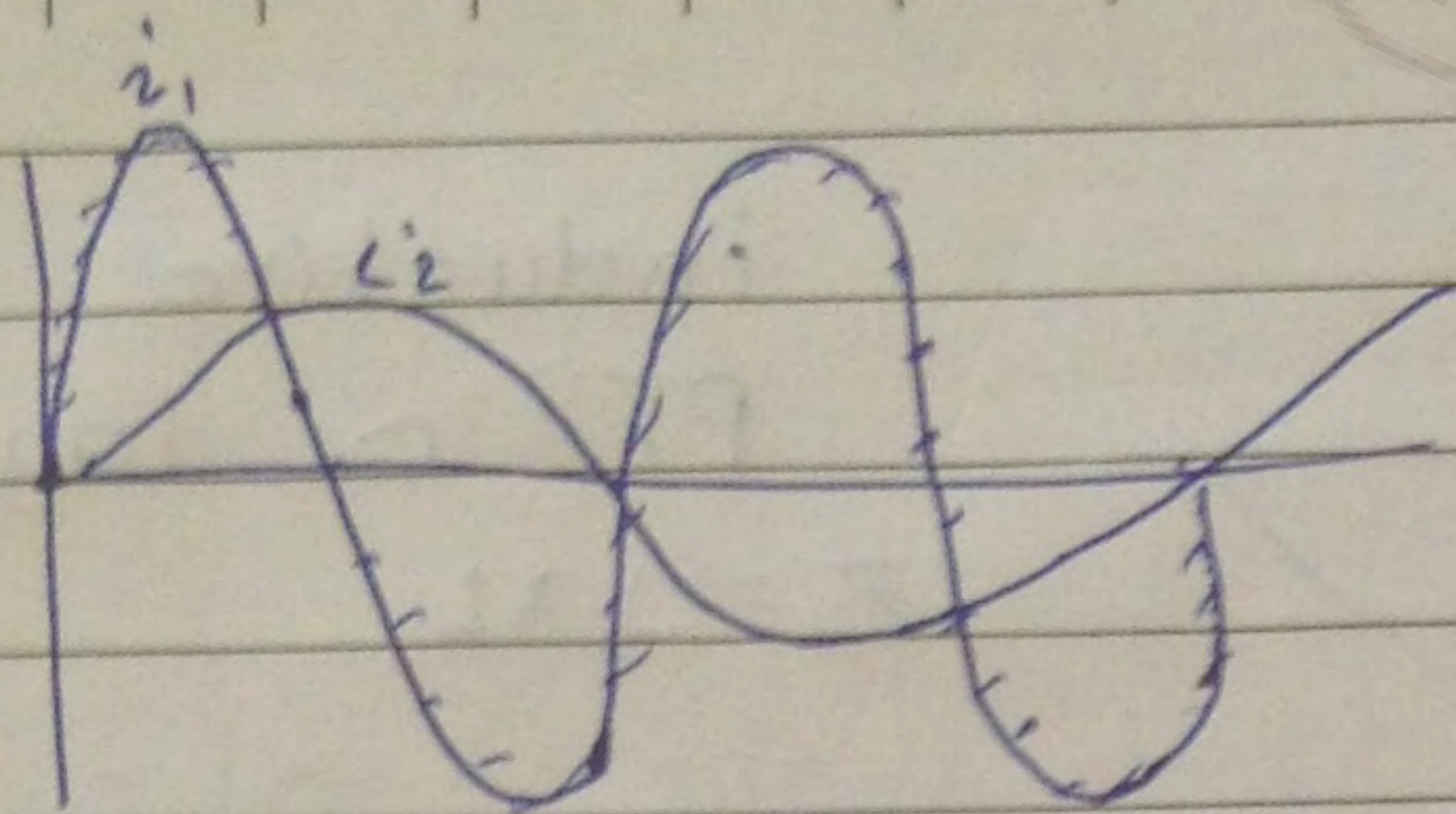
$$P = I_{eff}^2 R = \frac{R}{T} \int_0^T i^2(x) dx$$

$$I_{eff}^2 = \frac{1}{T} \int_0^T (\sin t + \sin \pi t)^2 dt$$

$$= \frac{1}{T} \left[\int_0^T (\sin^2 t + 2 \sin t \sin \pi t + \sin^2 \pi t) dt \right]$$

$$= \int_0^T \left[\frac{1}{2} (1 - \cos 2t) + \frac{1}{2} (1 - \cos 2\pi t) \right] dt$$

$$I_{rms} = \sqrt{I_{eff}^2 + I_{eff}^2}$$



$$i(t) = i_1(t) + i_2(t)$$

$$I_{\text{eff}} = \frac{1}{T} \int_0^T (I_m \sin(\omega t))^2 dt$$

$$= \frac{I_m^2}{T} \int_0^T \frac{1}{2} (1 - \cos 2\omega t) dt = \sqrt{\frac{I_m^2}{2}} = \boxed{\frac{I_{\text{max}}}{\sqrt{2}}}$$

$$\boxed{I_{\text{rms}} = \frac{I_{\text{max}}}{\sqrt{2}}}$$

* Since i_1 and i_2 have same frequency then they have to find the equivalent.

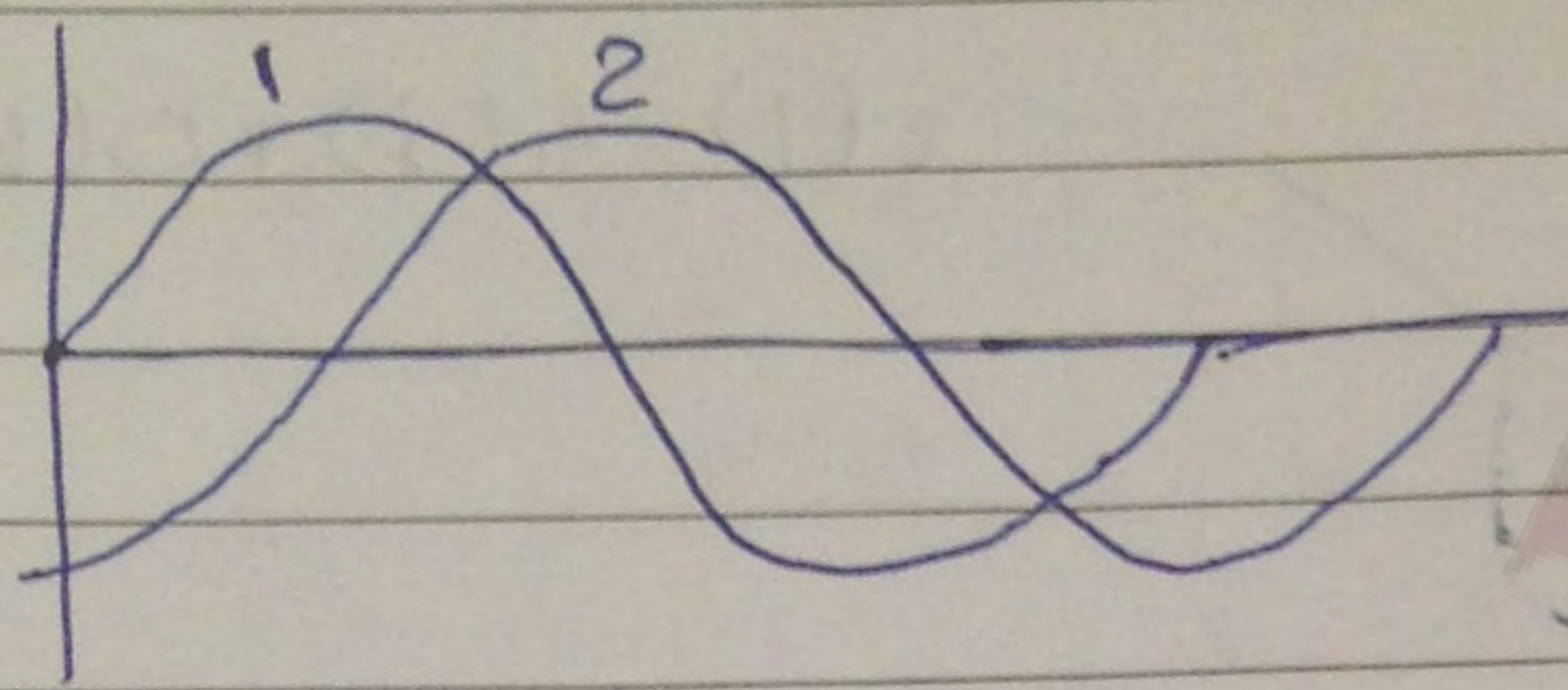
$$i_1 = I_{m1} \cos(2\omega t + \phi_1)$$

$$i_2 = I_{m2} \cos(\omega t + \phi_2) \rightarrow I_{m2} / \phi_2$$

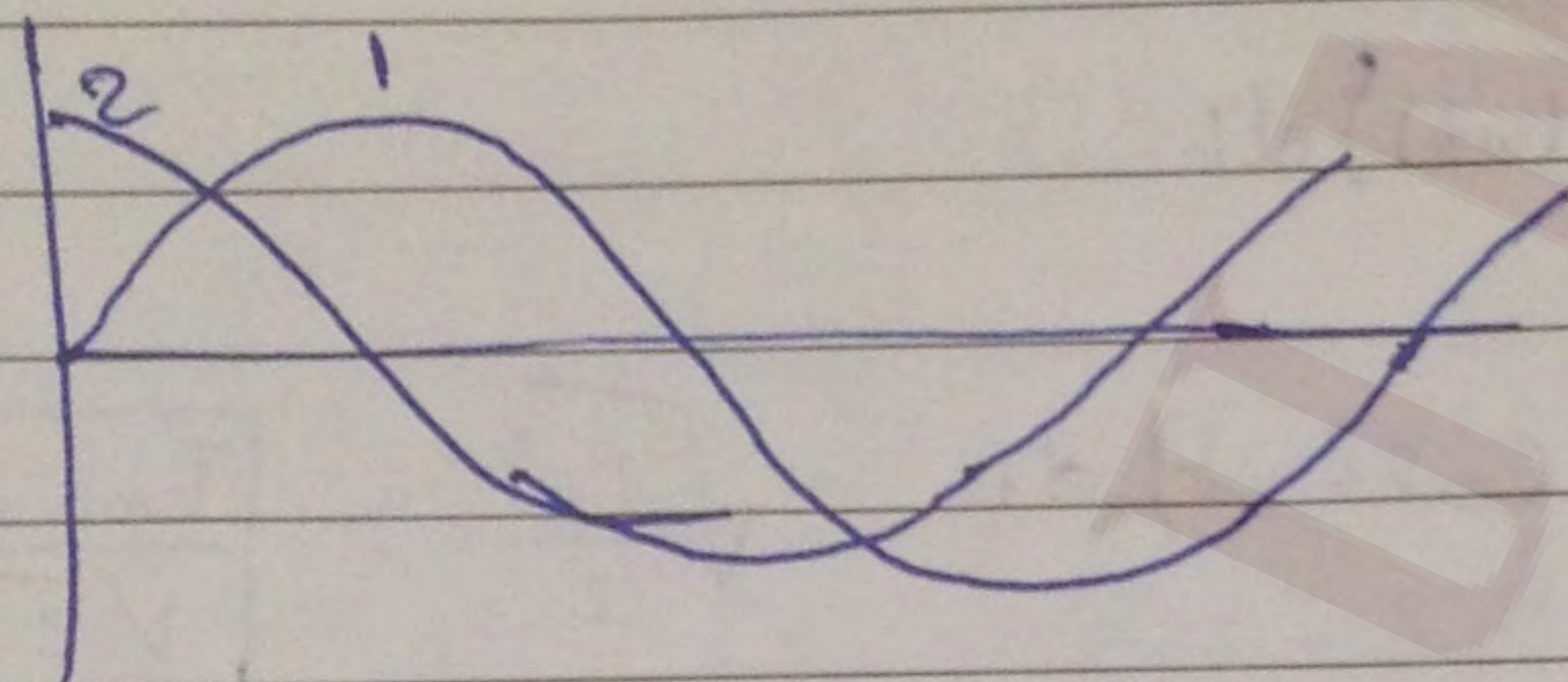
$$i_3 = I_{m3} \cos(\omega t + \phi_3) \rightarrow I_{m3} / \phi_3$$

$$I_{23} = i_2 + i_3 = I_{23(\text{max})} \cos(\omega t + \phi_{23})$$

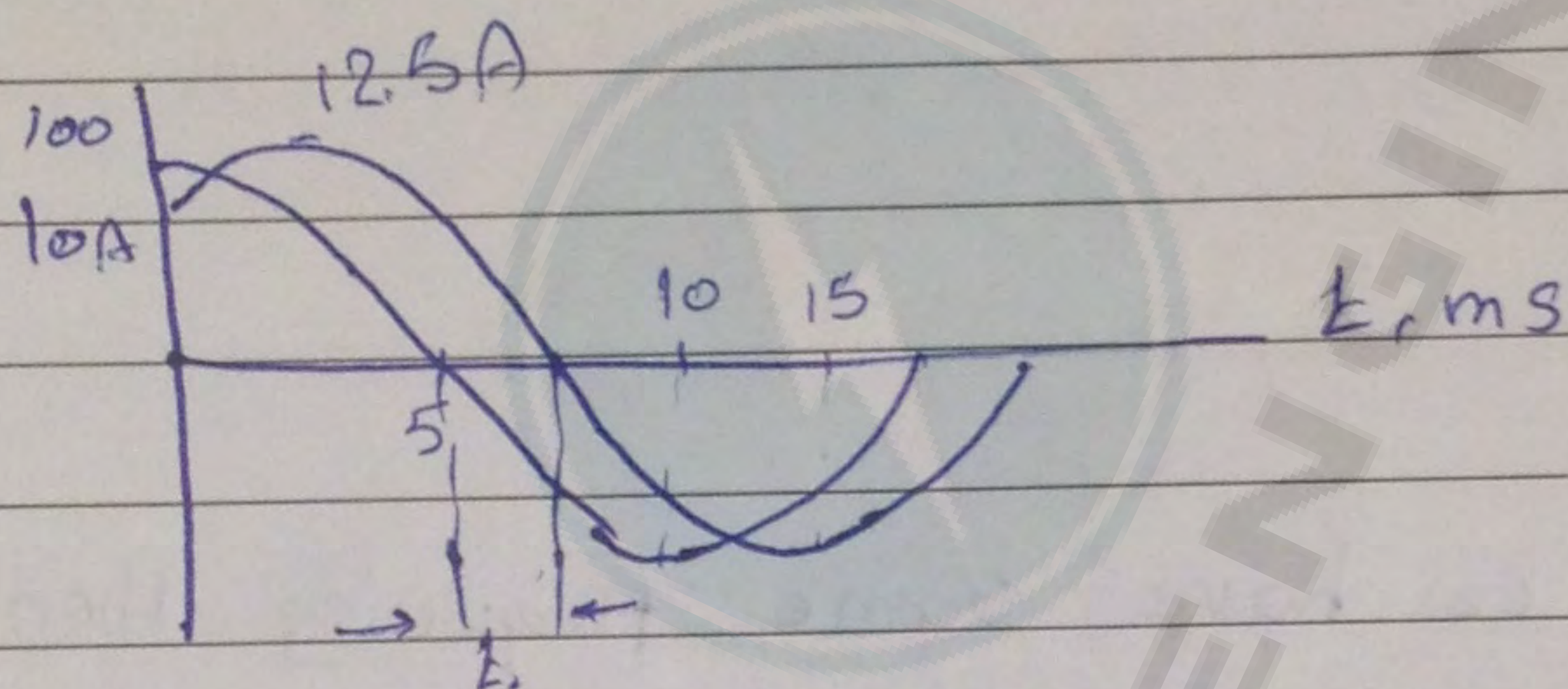
1 → V
2 → I



inductive
PF = 0 lagging
 $Z_L = j \times L$



inductive (same as previous)



$$10 \text{ cm} = 100 \text{ V} \rightarrow k_V = 10 \text{ V/cm}$$

$$10 \text{ cm} = 12.5 \text{ V} \rightarrow k_I = 1.25 \text{ A/cm}$$

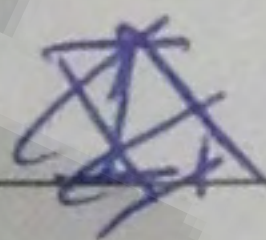
$$v(t) = 100 \cos(100\pi t)$$

$$i(t) = 12.5 \cos(100\pi t - \cos^{-1}(\frac{0}{12.5}))$$

$$T = 20 \text{ ms} = 360^\circ = 20 \text{ cm}$$

$$? = 2.5 \text{ cm}$$

$$\phi_0 = \frac{2.5 \times 360^\circ}{20 \text{ cm}} = 45^\circ$$



$$I_1 = 100 \angle 0$$

$$I_2 = 100 \angle -36.87$$

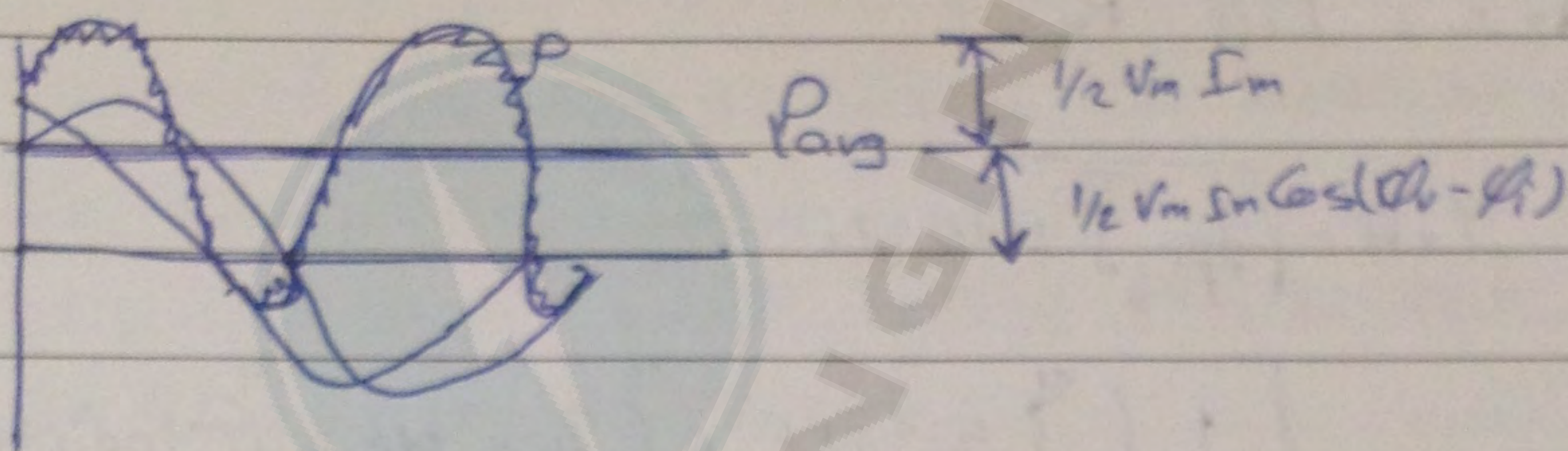
$$I_{12} = 100 \angle 0 + 100 \angle -36.87$$

$$= 100 + 80 - j60$$

$$= 180 - j60 = \sqrt{180^2 + 60^2} \angle -\tan^{-1}\left(\frac{60}{180}\right) = 190 \angle -18$$

$$P = \frac{1}{2} V_m I_m \cos 36.87$$

$$= \frac{1}{2} (100)(12.5) 0.8 = 500 \text{ W}$$

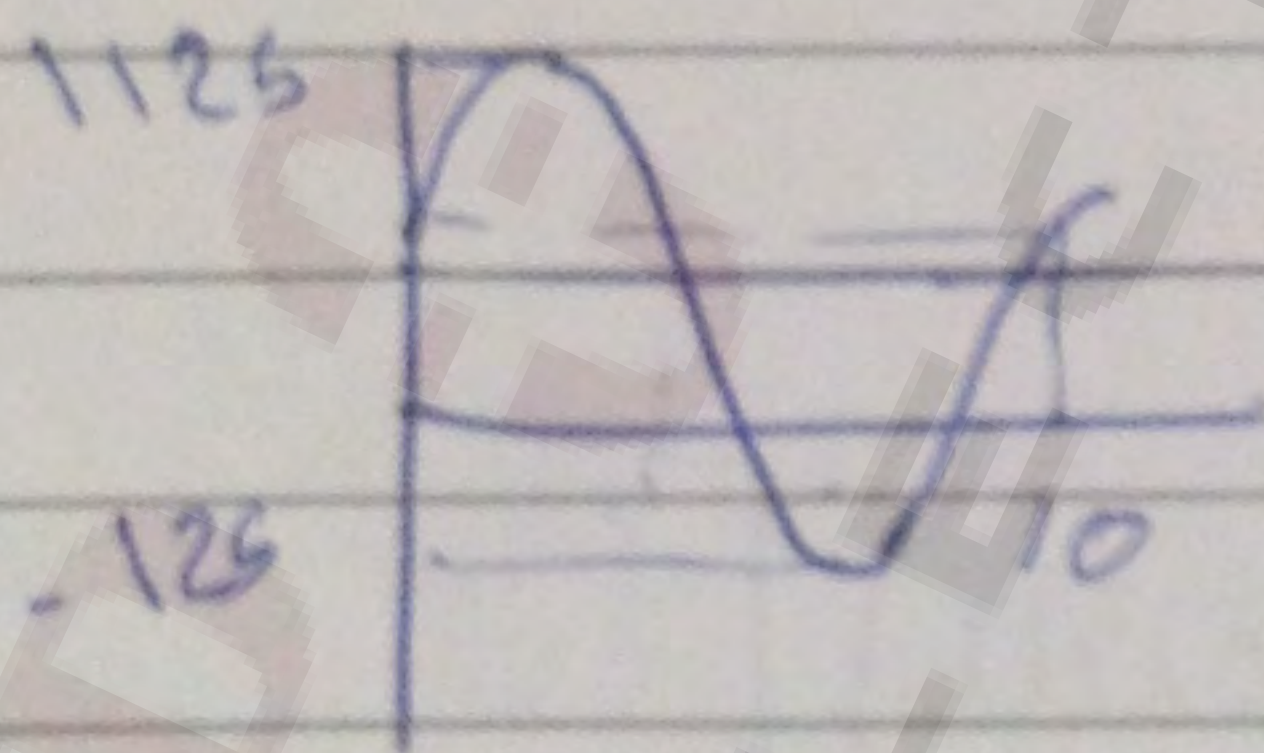


$$P(t) = 500 + 625 \cos(200\pi t - 36.87)$$

$$= \frac{1}{2} V_m I_m \cos \phi + \frac{1}{2} V_m I_m (\cos 2\omega t + \phi)$$

$$\rightarrow P_{max} = \frac{1}{2} V_m I_m \cos \phi + \frac{1}{2} V_m I_m = \frac{1}{2} V_m I_m (PF + 1)$$

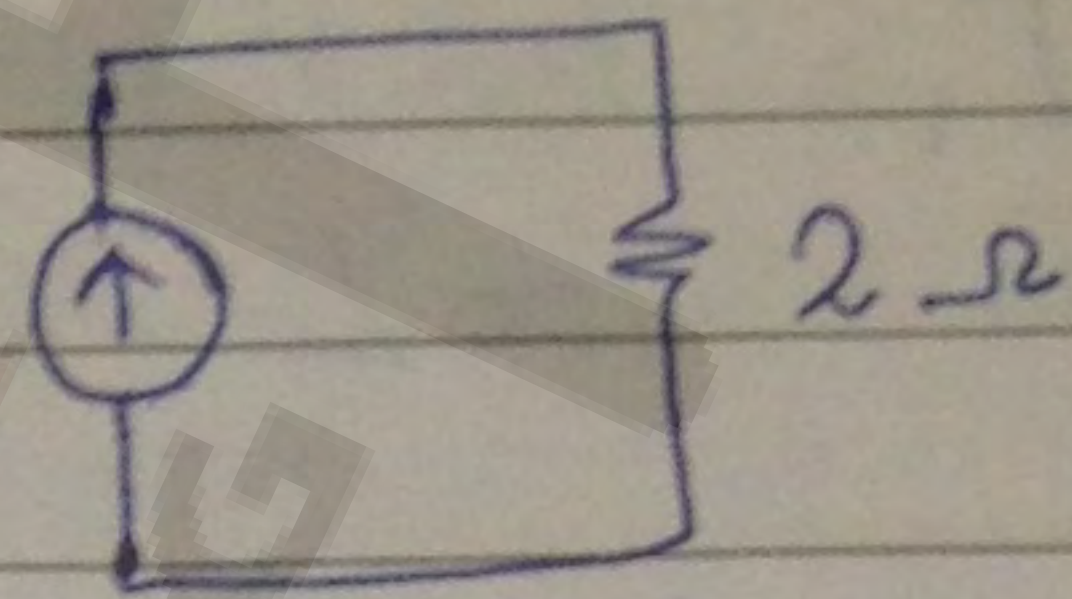
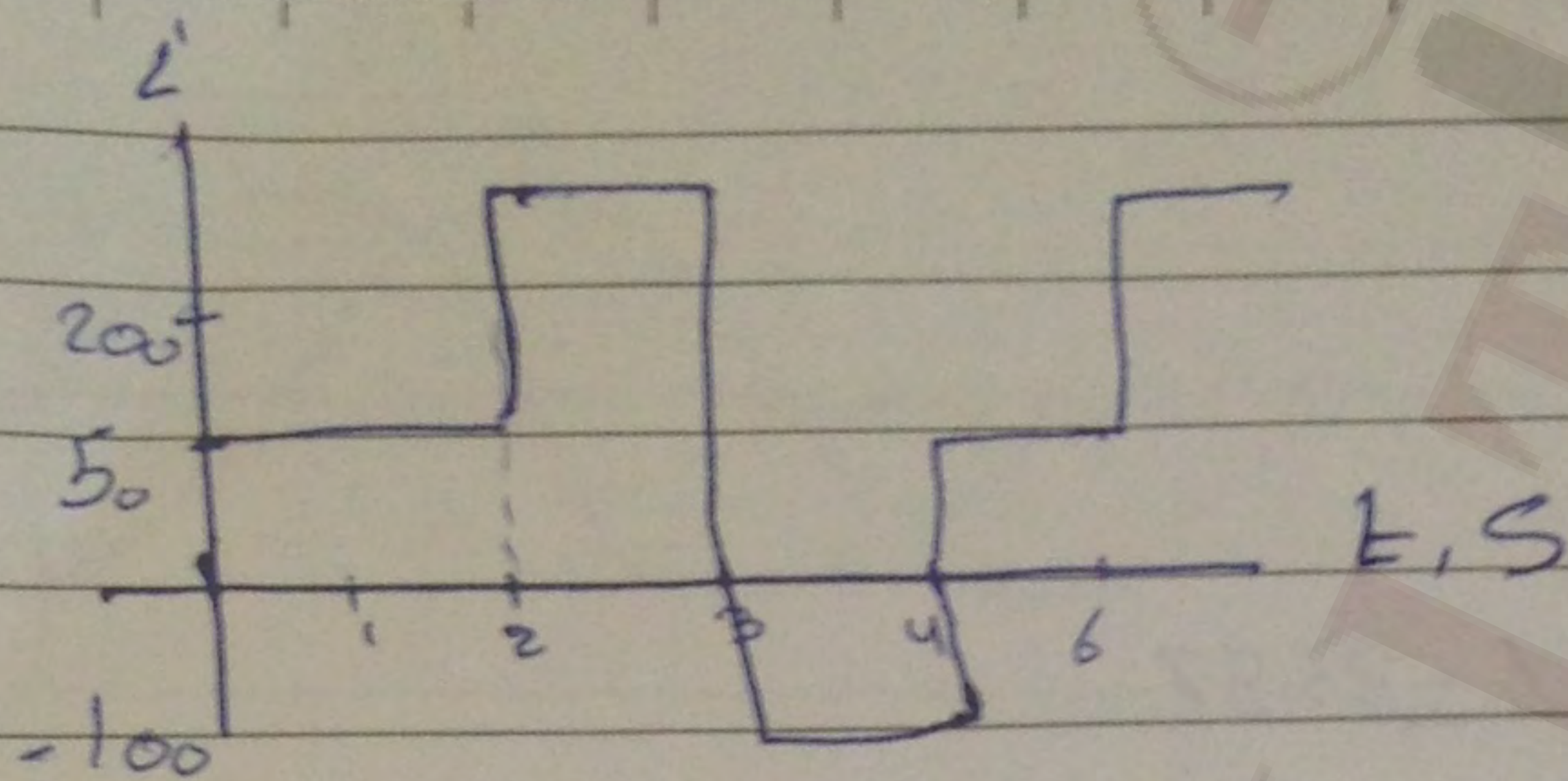
$$\rightarrow P_{min} = \frac{1}{2} V_m I_m \cos \phi - \frac{1}{2} V_m I_m = \frac{1}{2} V_m I_m (PF - 1)$$



$$P_{av} = P_{max} - \frac{P_{min}}{2} = 500 \text{ W}$$

$$P_{max} = S(PF + 1) \rightarrow 1125 = S(PF + 1)$$

exp



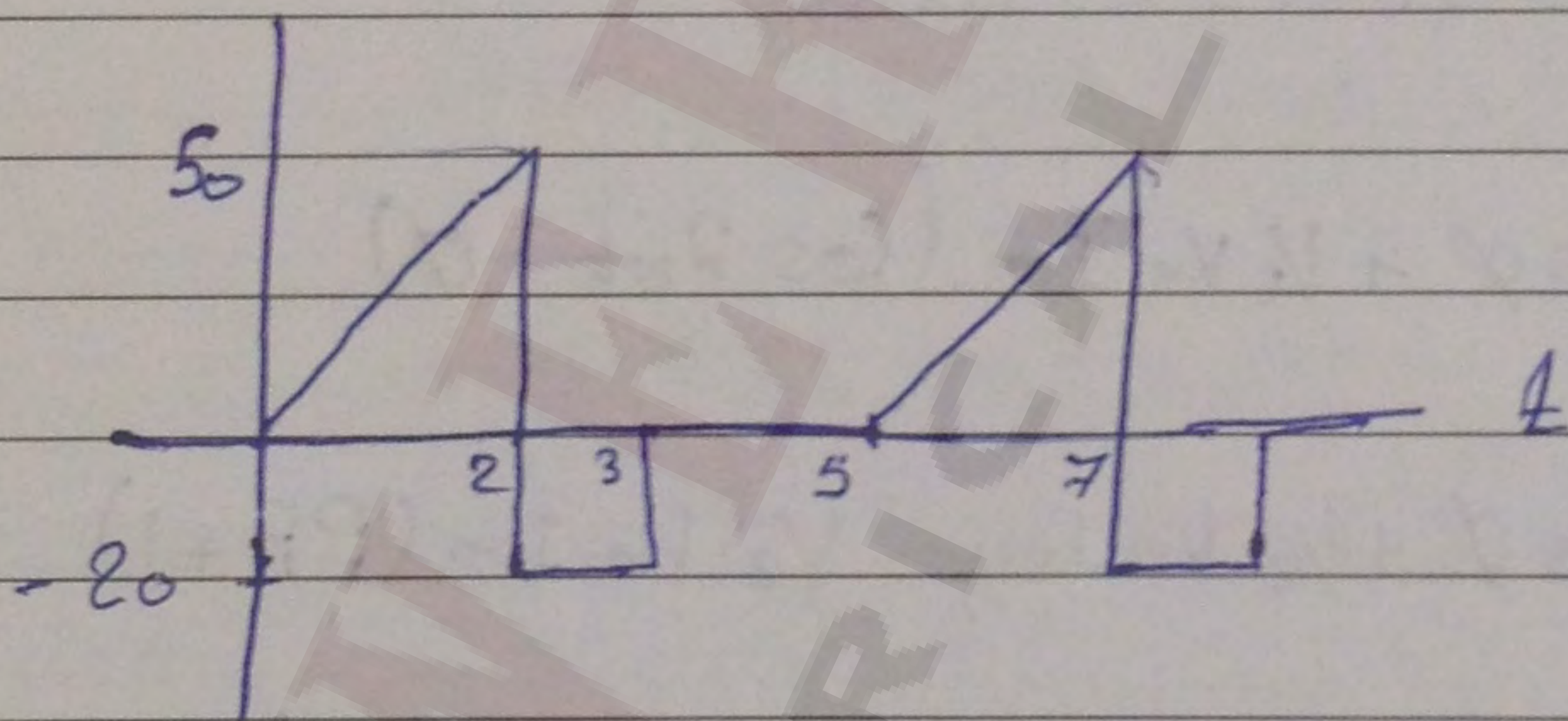
$$i(t) = \begin{cases} 50 & 0 < t < 2 \\ 200 & 2 < t < 3 \\ -100 & 3 < t < 4 \end{cases}$$

$$P_{2\Omega} = I_{\text{eff}}^2 R$$

$$I_{\text{eff}} = \sqrt{\frac{1}{T} \int_0^T i^2 dt}$$

$$= \left[\frac{1}{4} \left(\int_0^2 50^2 dt + \int_2^3 200^2 dt + \int_3^4 (-100)^2 dt \right) \right]^{1/2}$$

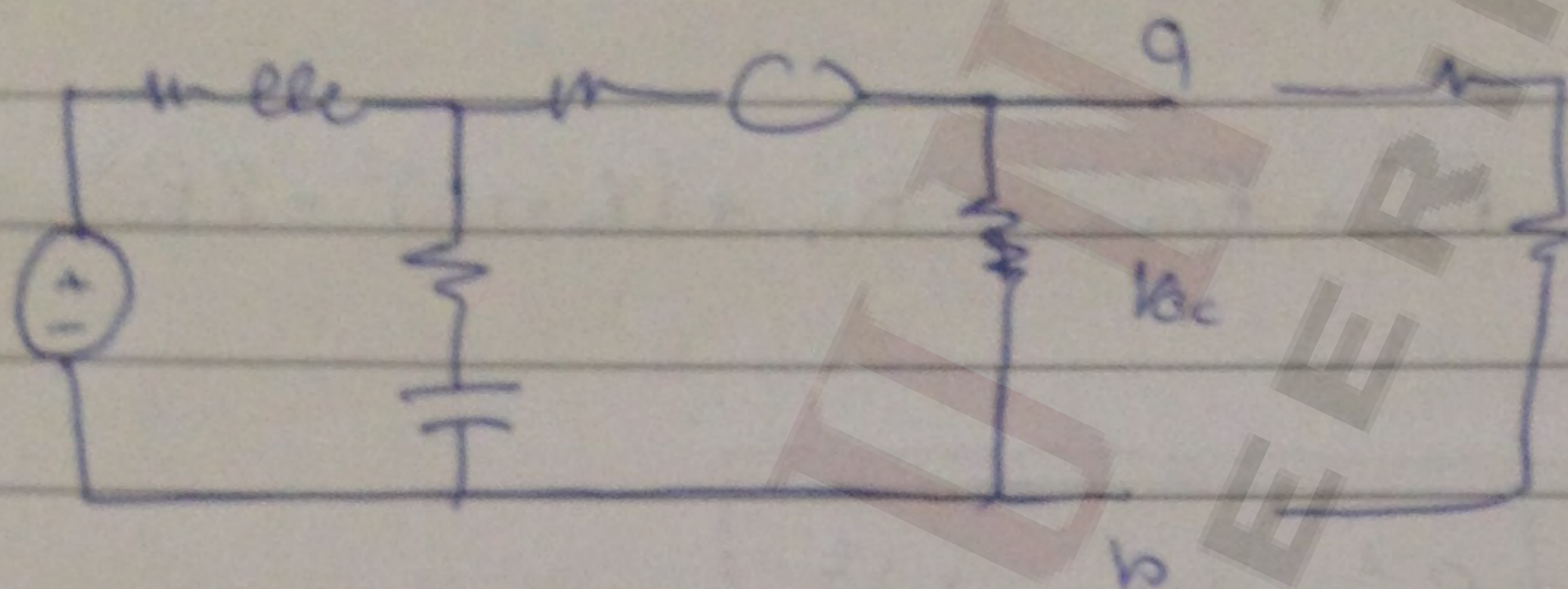
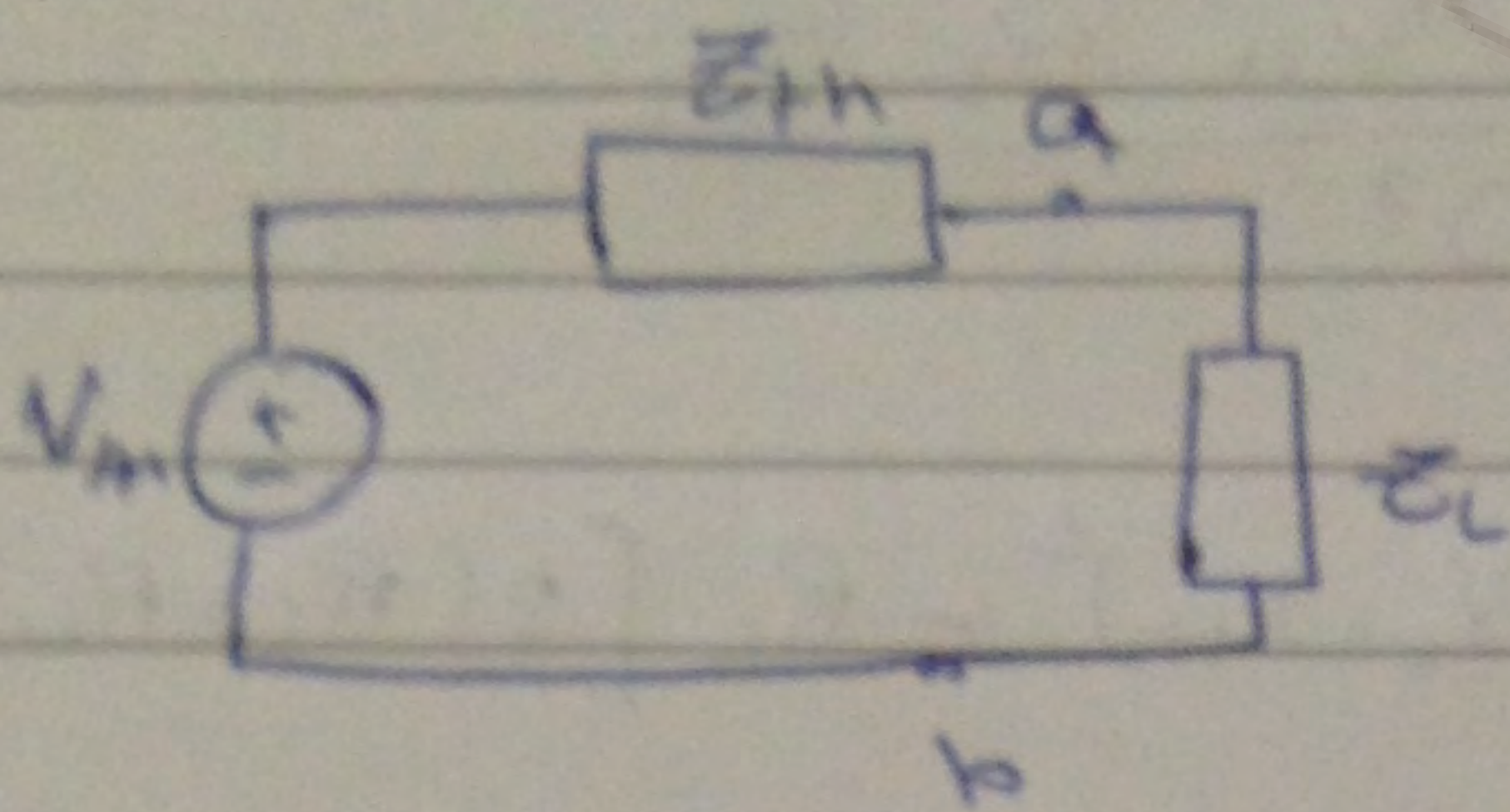
exp



$$i = \begin{cases} 25t & 0 < t < 2 \\ -20 & 2 < t < 3 \end{cases}$$

$$I_{\text{eff}} = \left[\frac{1}{5} \left(\int_0^2 (25t)^2 dt + \int_2^3 (-20)^2 dt \right) \right]^{1/2}$$

- Maximum Power Transfer:



- 1) Remove the branch where we have to find P
- 2) Calc. $V_{oc} = V_{th}$
- 3) Calc. Z_{th} after killing the sources
- 4) reconnect the ckt.

$$\bar{I} = \frac{V_{th}}{Z_L + Z_{th}} = \frac{V_{th}}{(R_L + jX_L) + (R_{th} + jX_{th})} = \frac{V_{th}}{(R_L + R_{th}) + j(X_{th} + X_L)}$$

$$P_L = |I|^2 R_L = \frac{1}{2} |I_{rms}|^2 R$$

$$|I| = I_{eff} = I_{RMS}$$

$$|I| = I_{eff} = \frac{V_{th}}{\sqrt{(R_{th} + R_L)^2 + (X_{th} + X_L)^2}}$$

$$P_L = \frac{V_{th}^2}{(R_{th} + R_L)^2 + (X_{th} + X_L)^2} \cdot R_L$$

$$\frac{\partial P_L}{\partial R_L} = 0 = \frac{V_{th}^2 (D) - V_{th}^2 R_L \frac{\partial P}{\partial R_L}}{D^2}$$

$$= V_{th}^2 \left[(R_L + R_{th})^2 + (X_L + X_{th})^2 - R_L [2(R_L + R_{th})] \right]$$

$$= V_{th}^2 \left[R_L^2 + R_{th}^2 + 2R_L R_{th} + (X_L + X_{th})^2 - 2R_L^2 - 2R_L R_{th} \right] = 0$$

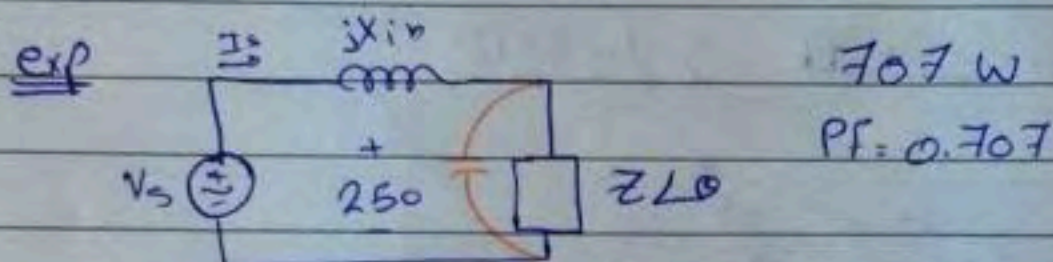
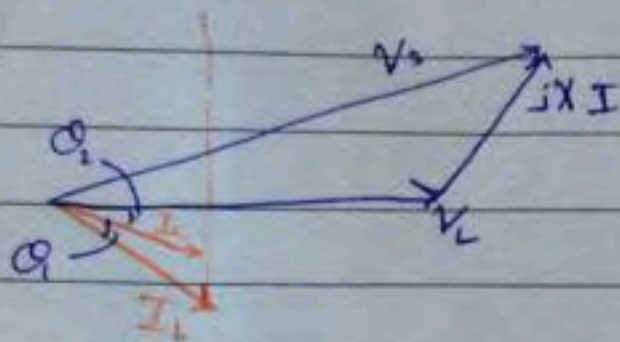
$$R_L = \sqrt{R_{th}^2 + (X_{th} + X_L)^2}$$

$$\frac{\partial P_L}{\partial X_L} = 0 = - \frac{R_L (2(X_L + X_{th}))}{D^2}$$

$$2R_L (X_L + X_{th}) = 0$$

$$X_L = -X_{th}$$

$$\therefore Z_L = R_{th} + jX_{th}$$



- ① increase the PF at the source to be 0.95 lag
- ② " " " " " load

$$|I_L| = \frac{|S|}{|V|}$$

$$I = \frac{S^*}{V^*}$$

$$S = \frac{P}{PF} = 1000 \text{ VA}$$

$$\theta = \cos^{-1} PF = 45^\circ$$

$$S = 1000 \angle 45^\circ$$

$$I = \frac{1000 \angle 45^\circ}{250 \angle 0^\circ} = 4 \angle 45^\circ \text{ A}$$

$$I_L = 4 \angle 45^\circ \text{ A}$$

$$\begin{aligned} S_{\text{line}} &= V_{\text{line}} I_{\text{line}}^* = Z I_L I_L^* \\ &= Z I_L^2 \\ &= j5(4)^2 = j80 \text{ VAR} \end{aligned}$$

Note

$$P = S \cos \theta$$

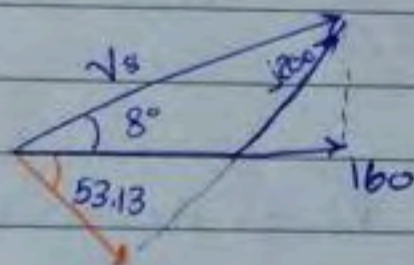
$$S = \frac{P}{\cos \theta} = \frac{P}{PF}$$

exp 500V 0.6 lag
1000 VA

$$|S| = |V| \cdot |I|$$

$$|I| = \frac{1000}{500} = 2A$$

$$\bar{I} = 2 \angle -\cos^{-1} PF = 2 \angle -53.13^\circ$$



$$V_s = (600 + j100) + j100$$

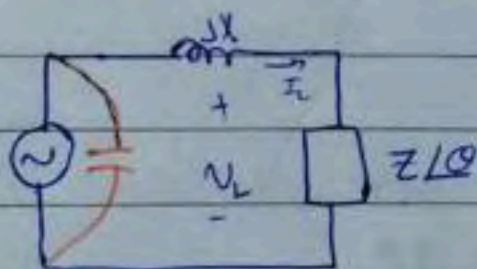
$$= \sqrt{660^2 + 100^2} = 667.5V$$

$$PF = \cos(\theta_{V_s} - \theta_{I_s}) = \cos(8 - (-53.13))$$

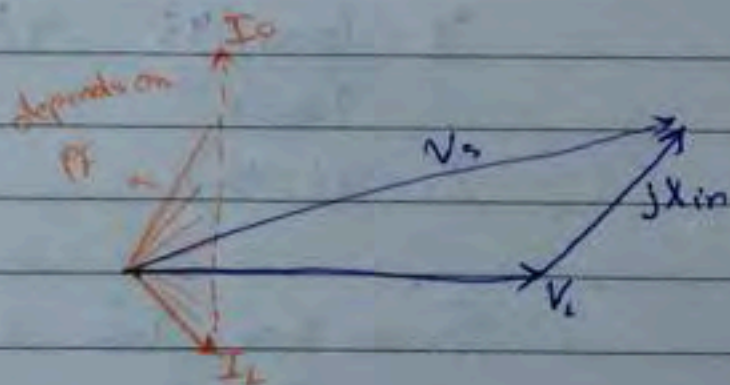
$$= \cos 61.13 = \boxed{0.48} \text{ lag}$$

exp Improve the PF of the source to the value of 0.9 lag
Find the Q_c, C -

$$\cos^{-1} 0.9 = 25.83^\circ$$

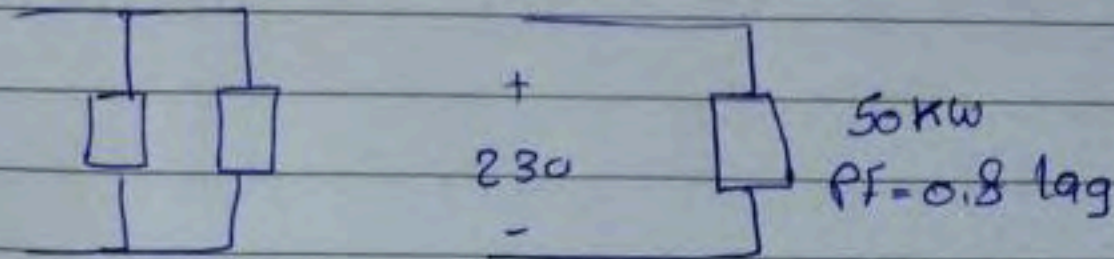


$$\theta_{V_s} = \theta_{I_s} = \theta_s$$

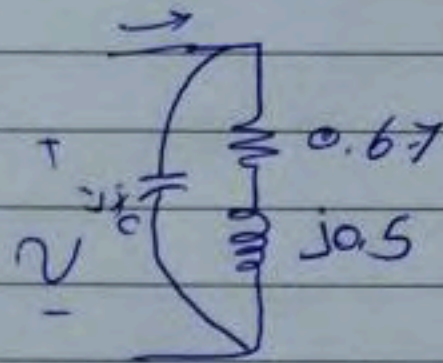
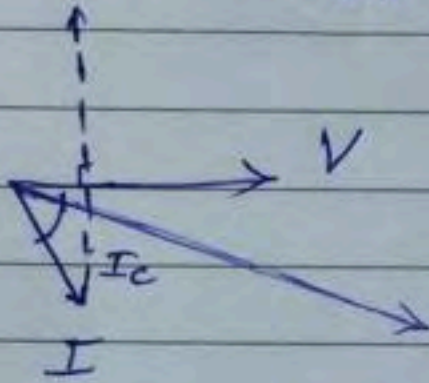


- required to decrease the angle between V_s and I_s to improve PF
- a capacitor is connected in parallel with the load

exp

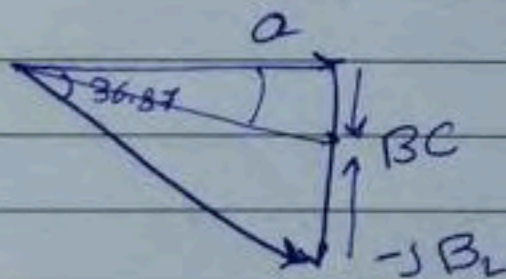


$$Z = \frac{V^2}{S^*} = \frac{230^2}{\frac{50}{0.8} - j\frac{50}{0.6}} = 0.841 \angle 36.87^\circ = 0.67 + j0.6$$



$$\cos(\phi_v - \phi_i) = PF = 0.8$$

$$Y_L = \frac{1}{Z} = \frac{1}{0.841}$$



$$\frac{I}{V} = \omega C = B_C = B_1 - C \tan \phi_2$$

$$X_C = \frac{1}{B_C}$$

$$S_s = \sqrt{707^2 + 78^2}$$

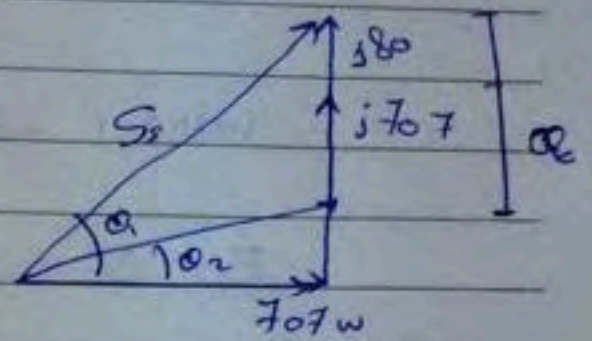
$$= 1057.48$$

$$Q_c = Q_1 - Q_{new}$$

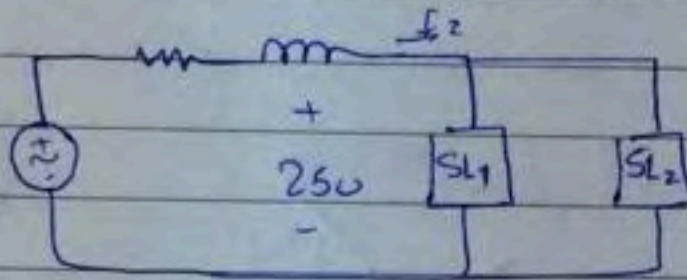
$$= P \tan \theta_1 + P \tan \theta_2$$

$$= P (\tan \theta_1 - \tan \theta_2)$$

$$= W_c V_c^2 = P (\tan \theta_1 - \tan \theta_2)$$



exp

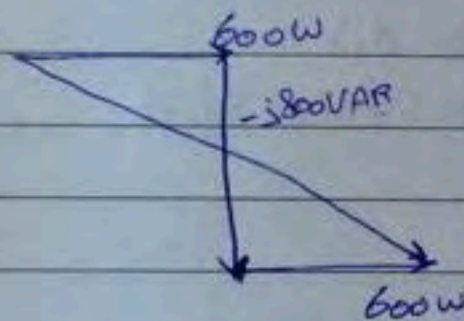


PF load?

$$L_1 = 1000 \text{ W}$$

$$L_2 = 600 \text{ W}$$

$$PF = 0.6 \text{ leading}$$



$$S_L = 1200 - j800 = 1442 \angle -33.7^\circ$$

$$I_L = \frac{1442}{250} = 5.8 \text{ A}$$

$$\tilde{I}_L = 5.8 \angle 33.7^\circ$$

Why the angle of the load cull. is +ve? inductive

$$S = V \cdot I^* = \frac{V^2}{Z^*}$$

$$\therefore Z = \frac{V^2}{S} = \frac{250^2}{1442 \angle 33.7}$$

$$= 36.5 - j24$$

$$\therefore PF_{load} = \cos 33.7 \text{ lead}$$

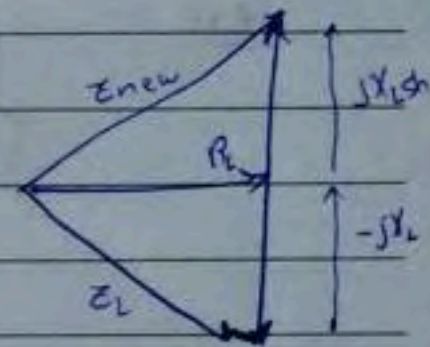
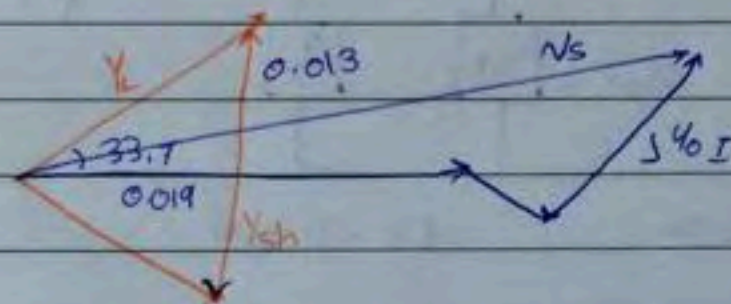
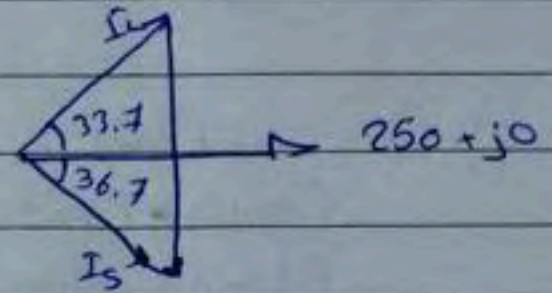
$$= 0.83 \text{ leading}$$

ex Required to find the element that have been connected in parallel with load to obtain a 0.8 PF lagging

inductor $-jX_{Lsh}$

$$Z_L = 36.05 - j24$$

$$Y_L = \frac{1}{Z_L} = 0.019 + j0.013$$



$$Y_{sh} = G \tan 33.7 + G \tan \cos^{-1} 0.8$$

$$= G (\tan \theta_1 + \tan \theta_2)$$

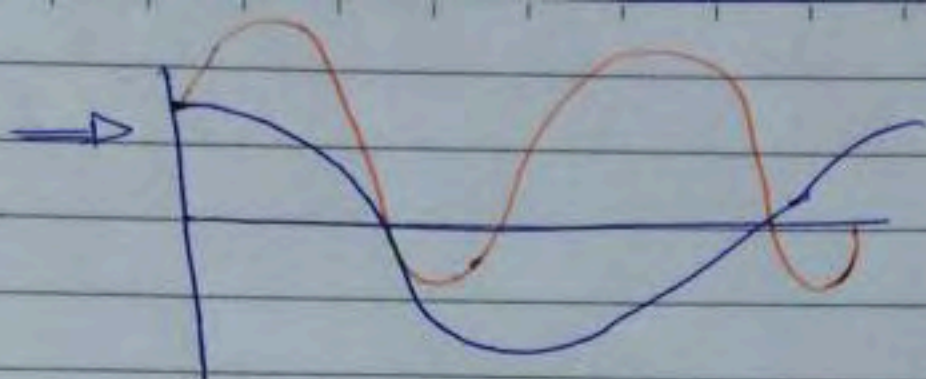
$$\frac{1}{\omega L} = 0.019 (\tan 36.87 + \tan 33.7)$$

Note : $Y = G + jB$

$$= \frac{1}{R} + j(\omega C - \frac{1}{\omega L})$$

$$= \frac{1}{R} + j(B_C - B_L)$$

H.W Prove $\sum P = 0$
 $\sum \theta = 0$



$$p(t) = V_m \cos(\omega t + \theta_v) I_m \cos(\omega t + \theta_i)$$

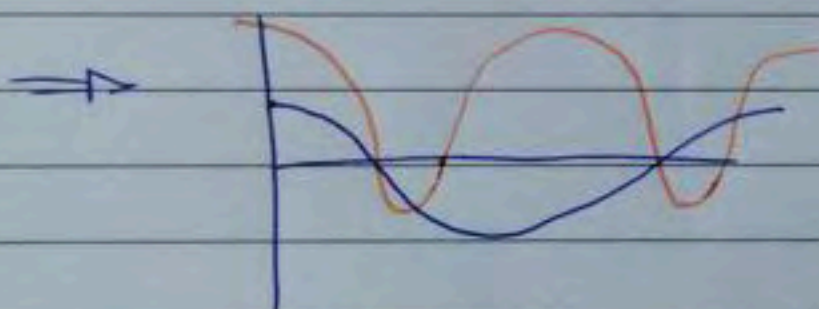
$$p(t) = \frac{1}{2} V_m I_m [\cos(\theta_v - \theta_i) + \cos(2\omega t + \theta_v + \theta_i)]$$

$$P_{avg} = \frac{900 - 900 - 100}{2} = 500$$

$$P = P_{avg} \cos(2\omega t + (\theta_v + \theta_i))$$

$$\theta_v + \theta_i = -\cos^{-1} \frac{400}{500} = -36.87^\circ$$

$$\therefore \theta_i = -36.87^\circ$$

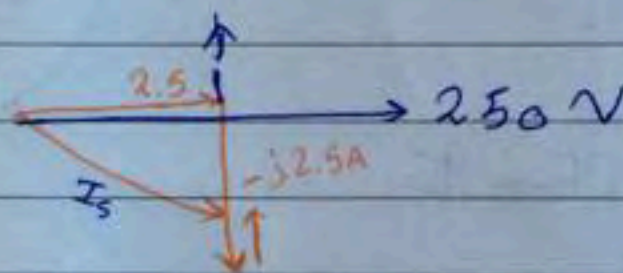
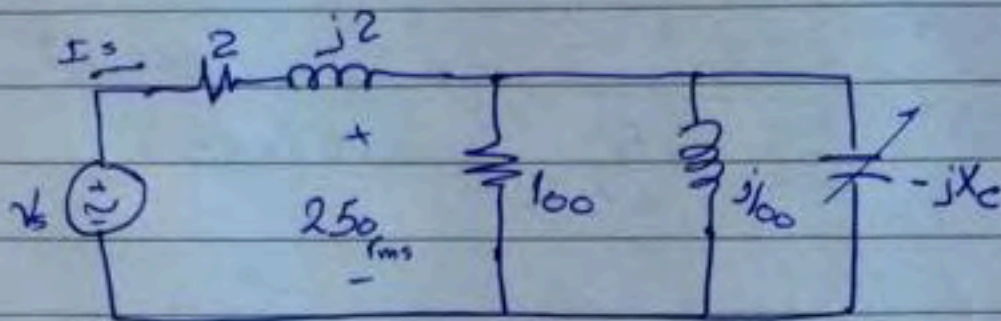


$$P + \frac{P}{PF} \cos(2\omega t + \theta_v + \theta_i)$$

$$\theta_i = 36.8^\circ$$

exp X_c is adjusted in such a way that the current I_s is minimum.

whats the value of X_c and the Complex Value delivered by the source.



$$V_s = 250 + 2.5(2 + j2.5)$$

$$= 255 + j5 = 255.1 \angle 1.1^\circ$$

$$S = -(255.1 \angle 1.1)(2.5) = -637.7 \angle 1.1$$

$$= -637.6 + j12.2$$

$$P = -637.7 \text{ w} \rightarrow \text{delivered}$$

$$Q = -12.2 \text{ VAR} \rightarrow \text{delivered.}$$

prove $\angle P = 0$

$$P_{\text{delv}} = 637$$

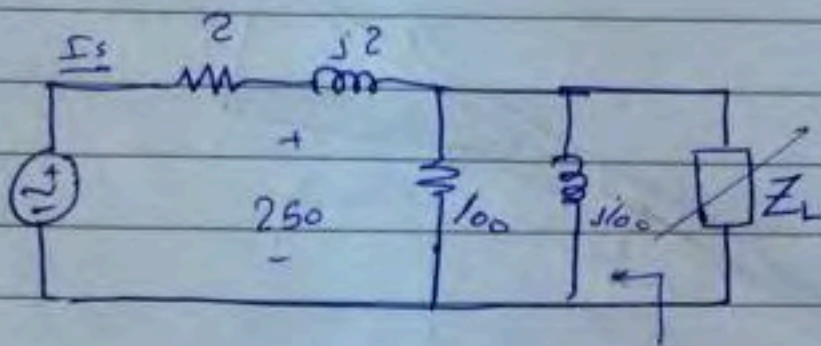
$$P_{\text{absorb}} = (2.5)^2 \times 100 + (2.5)^2 \times 2 = 637.5$$

$$\angle Q = 0$$

$$Q_{\text{abs}} = (2.5)^2 \times j2 + (2.5)^2 \times j100 = j637.5$$

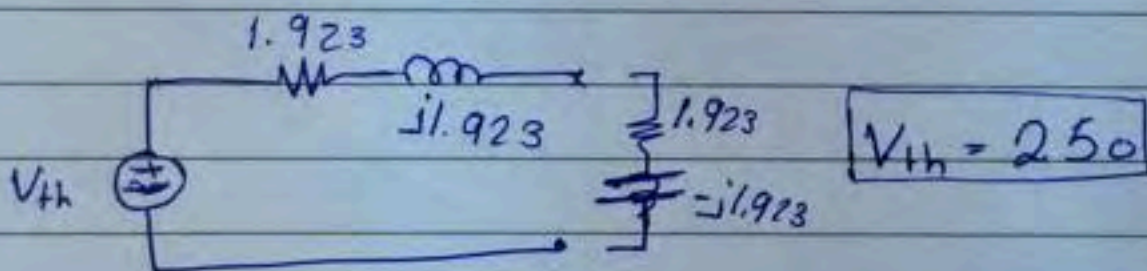
$$Q_{\text{delv.}} = (2.5)^2 - j100 = -j62.5 - j12.2$$

exp What's the value of Z_L that the avg power transfer to it equal to its maximum value.



$$Z_L = \frac{1}{Z_{\text{oc}}} = \frac{1}{\frac{1}{100} + \frac{1}{j100} + \frac{1}{2+j2}} \Rightarrow Z_{\text{oc}} = Z_{\text{th}} = 2.719 \angle 45^\circ$$

$$Z_L = Z_{\text{th}}^* \Rightarrow Z_L = 2.719 \angle -45^\circ$$



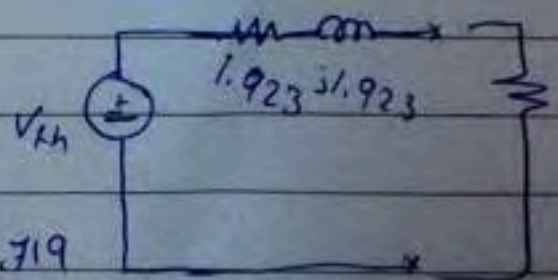
$$P_L = |I|^2 R_L = \left(\frac{V_{\text{th}}}{2 \times 1.923} \right)^2 \cdot 1.923 = 812.5 \text{ W}$$

⇒ IF we replaced Z_L with variable R .

$$R_L = \sqrt{R_{\text{th}} + (X_L + X_{\text{th}})^2}$$

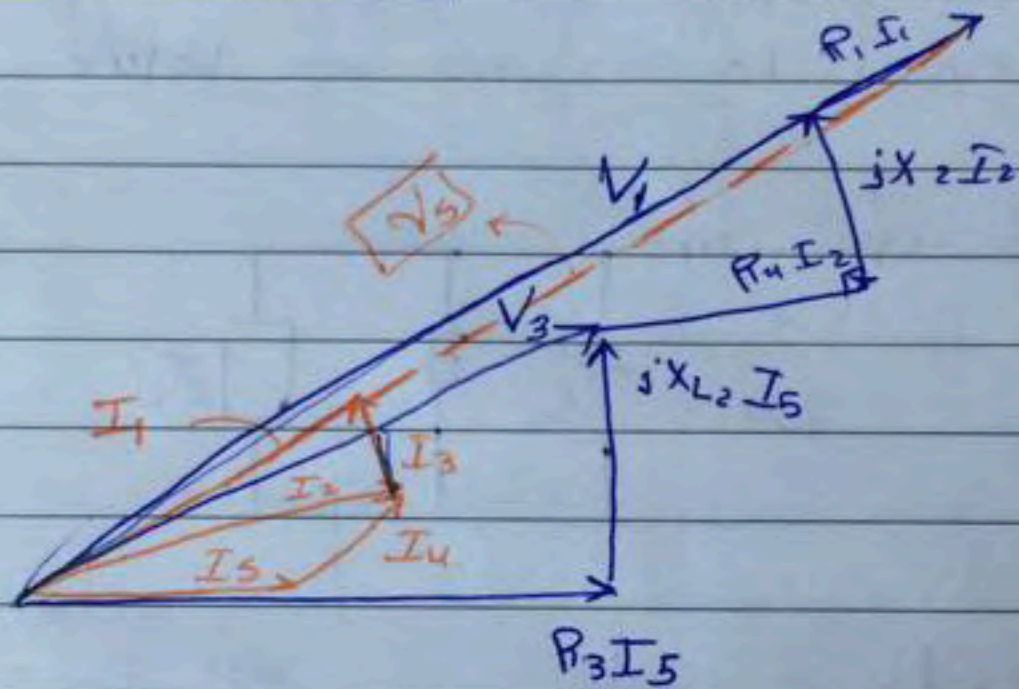
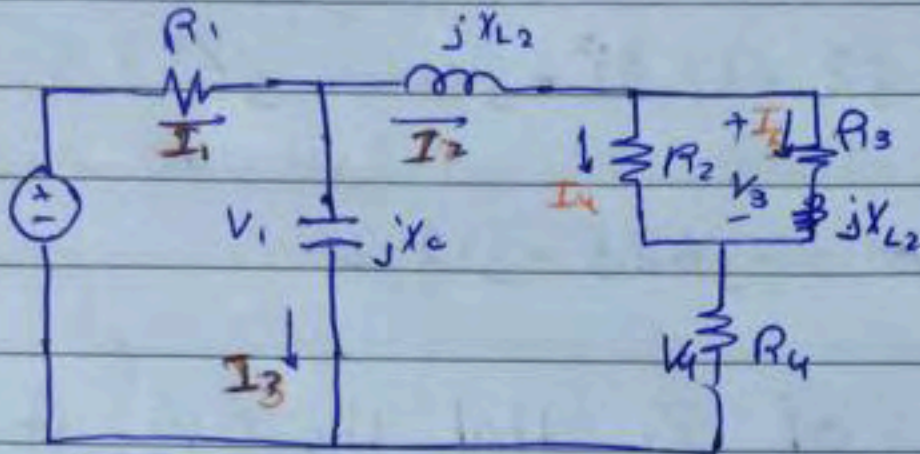
$$|Z_{\text{th}}| = 2.719 \Omega$$

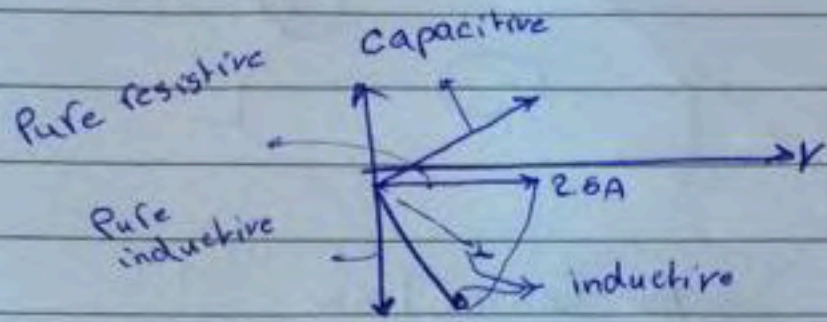
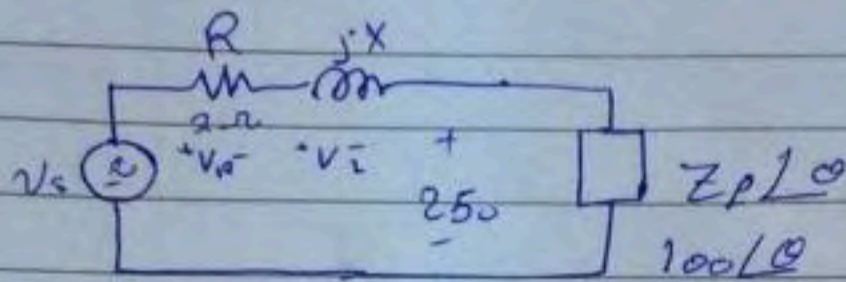
$$P = |I|^2 R = \frac{250^2}{(1.923 + 2.719)^2 + 1.923^2} \times 2.719 = \boxed{6722} \text{ W}$$



* P less than P when the load is Capacitor.

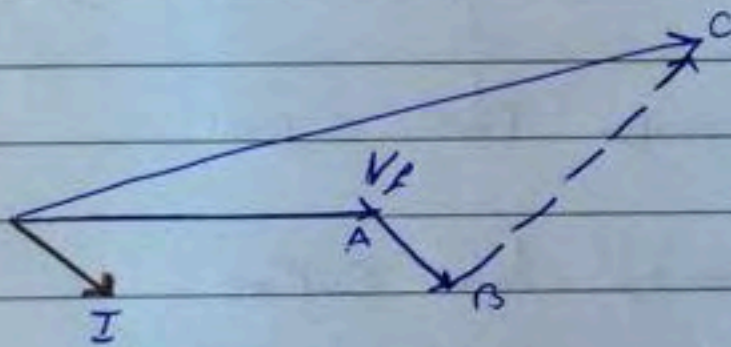
exp Draw phasor diagram.





$$\frac{1}{R + jX} = \frac{R}{R^2 + X^2} - j \frac{X}{R^2 + X^2} = G - jB$$

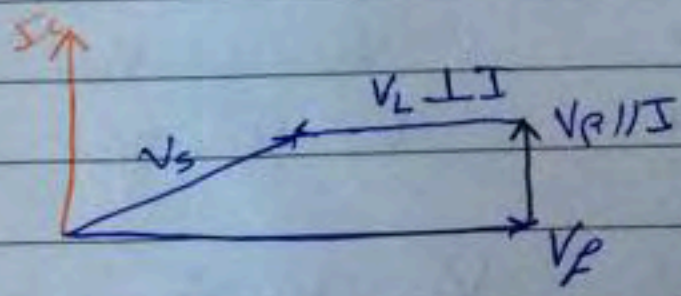
inductive
capacitive
inductive
capacitive



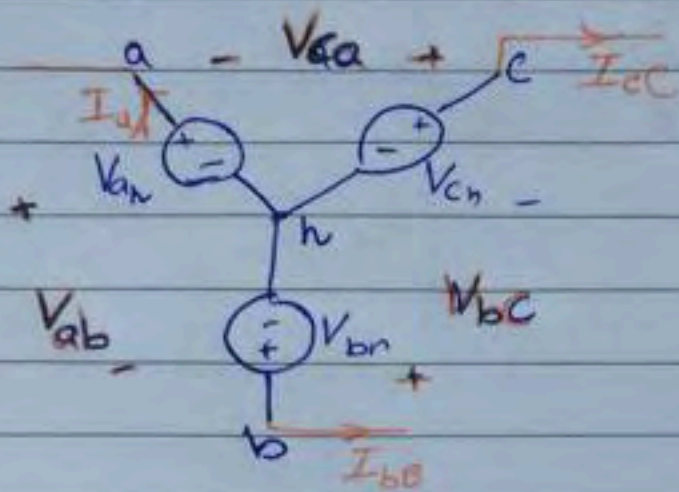
$$V_R = RI, \quad V_R \parallel I$$

$$V_L = jXI, \quad V_L \perp I$$

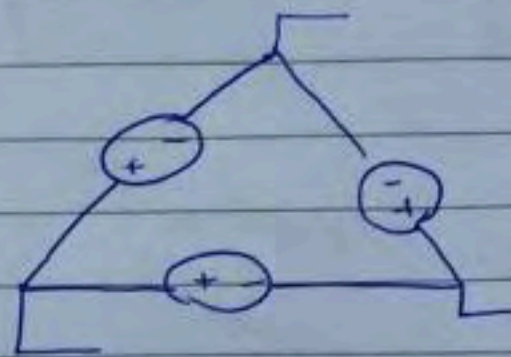
⇒



Three Phase :-

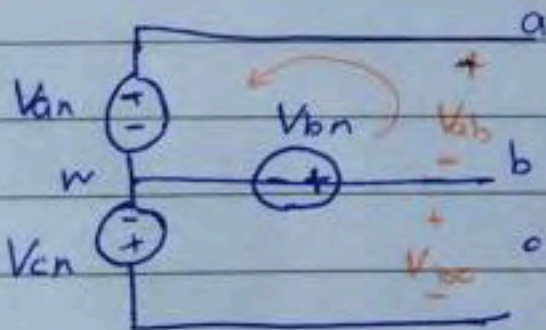


Y-Connection



Δ -Connection

- $V_{ab}, V_{bc}, V_{ca} \Rightarrow$ Line Voltage
- $V_{an}, V_{bn}, V_{cn} \Rightarrow$ Phase Voltage
- $I_{an}, I_{bn}, I_{cn} \Rightarrow$ Phase current
- $I_{aA}, I_{bB}, I_{cC} \Rightarrow$ Line Current

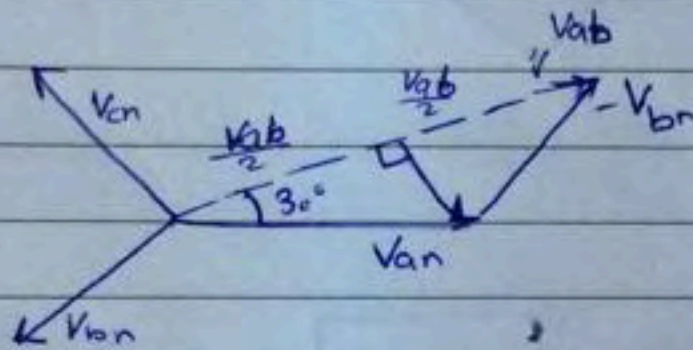
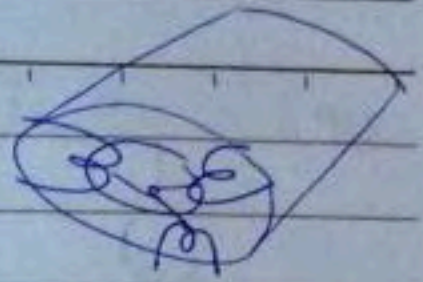
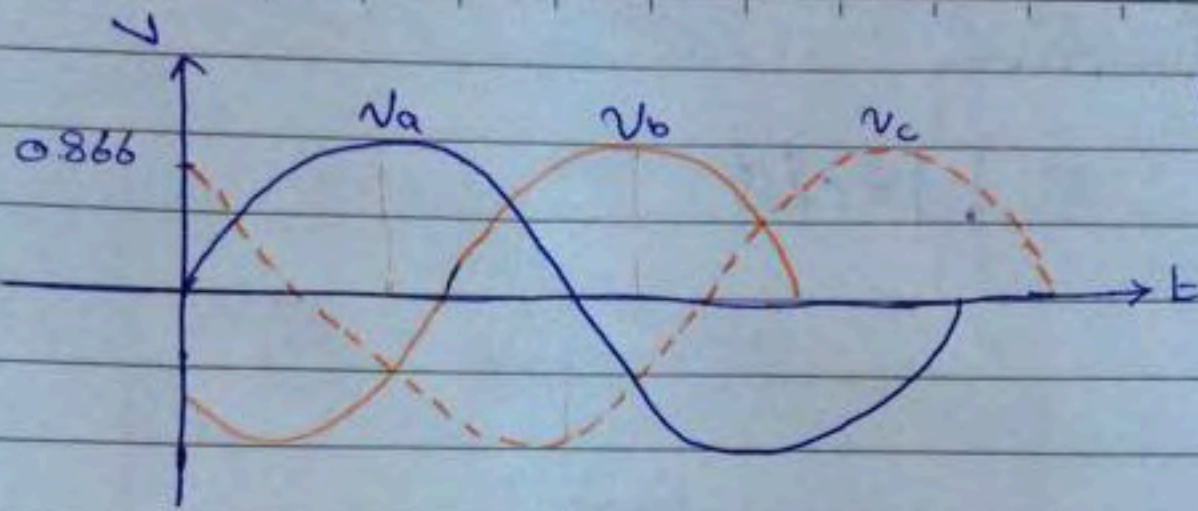


$$+V_{an} - V_{bn} - V_{cb} = 0$$

$$V_{ab} = V_{an} - V_{bn}$$

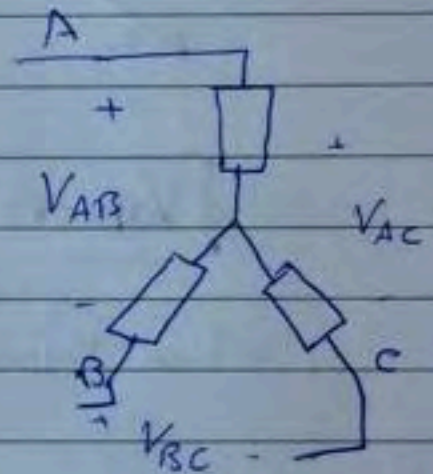
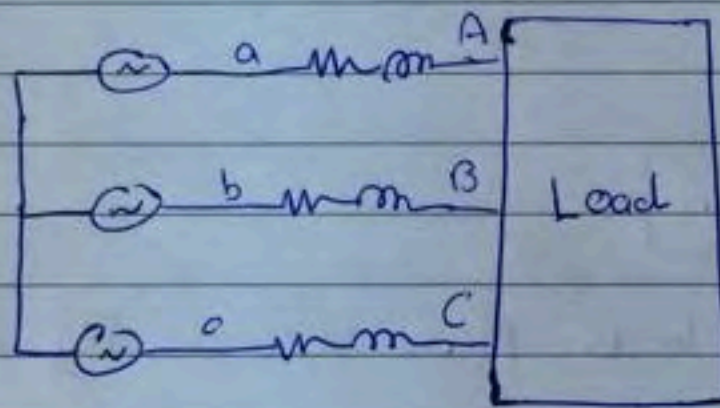
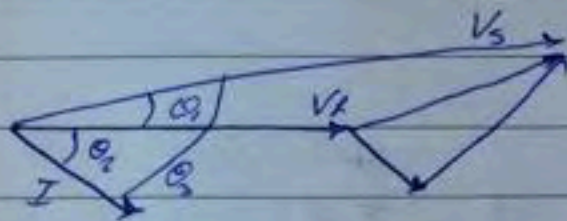
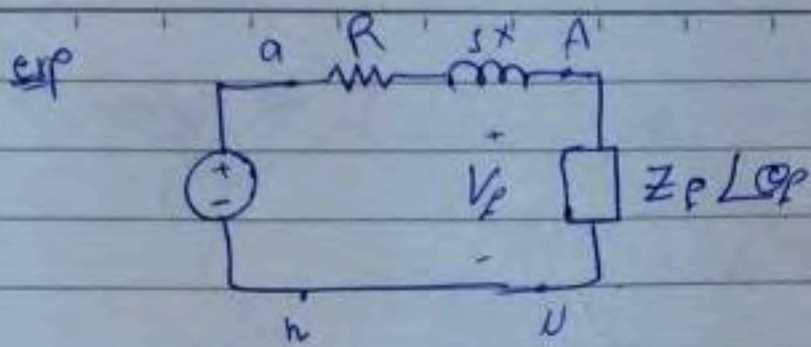
$$V_{bc} = V_{bn} - V_{cn}$$

$$V_{ca} = V_{cn} - V_{an}$$



$$V_{ab} = \sqrt{3} V_{an}$$

* line voltage leads phase voltage by 30°



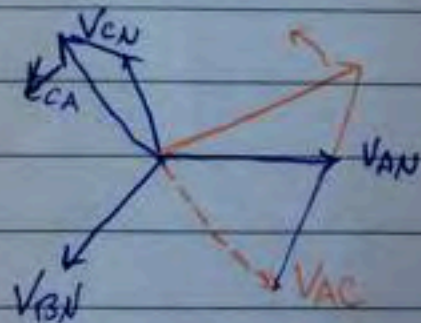
$$V_{AB} = \sqrt{3} V_{AN} \angle 30^\circ$$

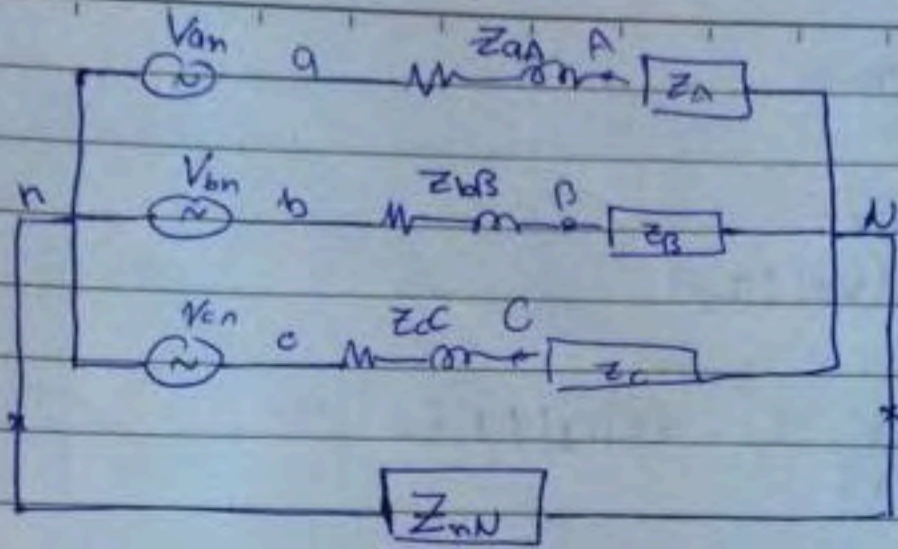
$$V_{CA} = \sqrt{3} V_{CN} \angle 30^\circ = \sqrt{3} V_{BN} \angle -90^\circ$$

$$V_{BC} = \sqrt{3} V_{BN} \angle 30^\circ$$

$$V_{CA} = \sqrt{3} V_{AN} \angle 150^\circ$$

$$V_{AC} = \sqrt{3} V_{AB} \angle -60^\circ$$





$$\frac{V_U - V_{an}}{Z_A + Z_{aA}} + \frac{V_U - V_{bn}}{Z_B + Z_{bB}} + \frac{V_U - V_{cn}}{Z_C + Z_{cC}} + \frac{V_U}{Z_{nN}} = 0$$

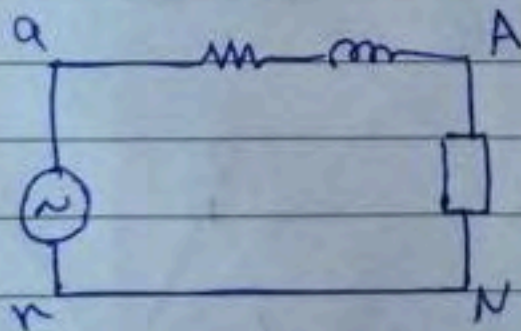
$$\Rightarrow Z_{\phi} = Z_A + Z_{aA} = 0$$

$$Z_A + Z_{aA} = Z_{\phi} = Z_B + Z_{bB} = Z_C + Z_{cC}$$

$$\left(\frac{1}{Z_{\phi}} + \frac{1}{Z_U} + \frac{1}{Z_{nN}} \right) V_U = \frac{V_{an} + V_{bn} + V_{cn}}{Z_{\phi}}$$

Since $V_{an} + V_{bn} + V_{cn} = 0$

$$\therefore V_U = 0$$



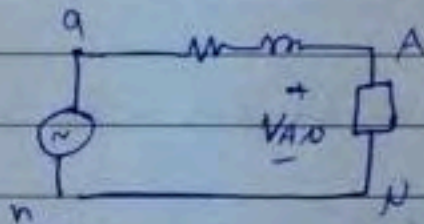
- Power in 3-phase

$$S_{3\phi} = 3 S_{\phi P} \\ = 3 |V_{\phi}| |I_{\phi}|$$

$$= 3 \frac{V_L}{\sqrt{3}} I_L = \sqrt{3} |V_L| |I_L|$$

$$\Rightarrow \boxed{S_{3\phi} = \sqrt{3} |V_L| |I_L|}$$

exp: $V_{AB} = 415 \text{ V}$
 30 kVA
 $PF = 0.8$



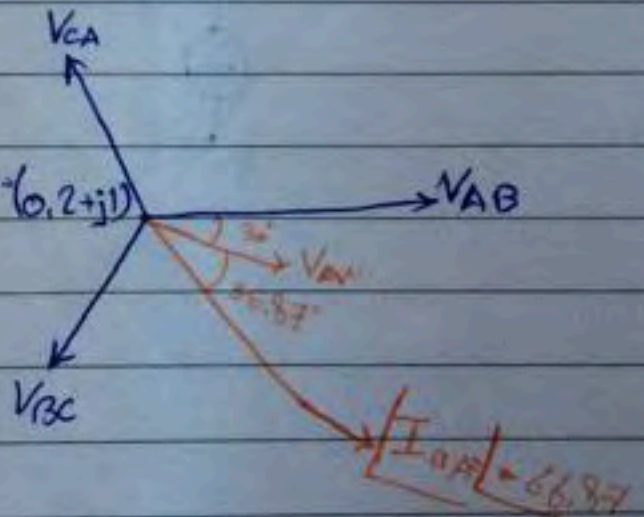
$$V_{AB} = \sqrt{3} V_{AN} \angle 30^\circ \rightarrow V_{AN} = \frac{V_{AB}}{\sqrt{3}} \angle -30^\circ$$

$$V_{AN} = \frac{415}{\sqrt{3}} \angle -30^\circ$$

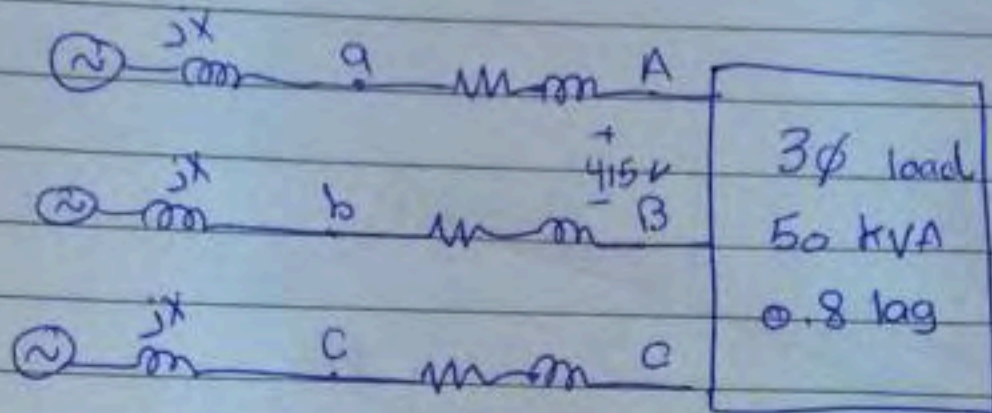
$$|I_L| = \frac{S_{3\phi}}{\sqrt{3} V_L} = \frac{50000}{\sqrt{3} (415)} = \boxed{69.5}$$

$$I_{aA} = 69.5 \angle -66.87^\circ$$

$$V_{an} = \frac{415}{\sqrt{3}} \angle -30^\circ \times 69.5 \angle -66.87^\circ (0.2 + j1)$$



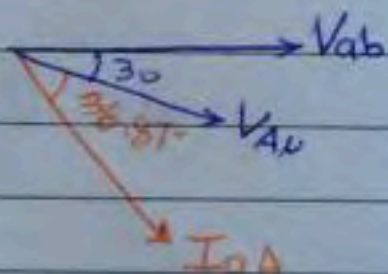
exp



V_{ab}, V_{an}, S_0 ?

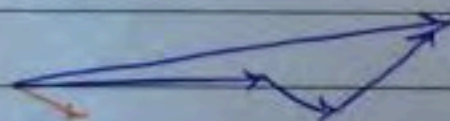
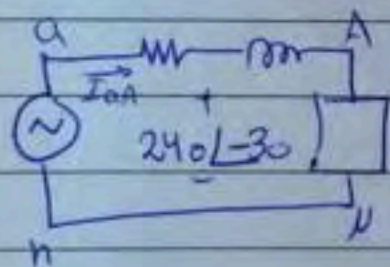
$$|I_L| = \frac{|S_{3\phi}|}{\sqrt{3} |V_L|} = \frac{50,000}{\sqrt{3} (415)} = 69.5 \text{ A}$$

∴ Since PF is lag by 80°
then the phase current has to be
lag by 36.87° phase v



$$I_{aA} = 69.5 \angle -66.87^\circ$$

⇒ Use 1φ ckt to phase a



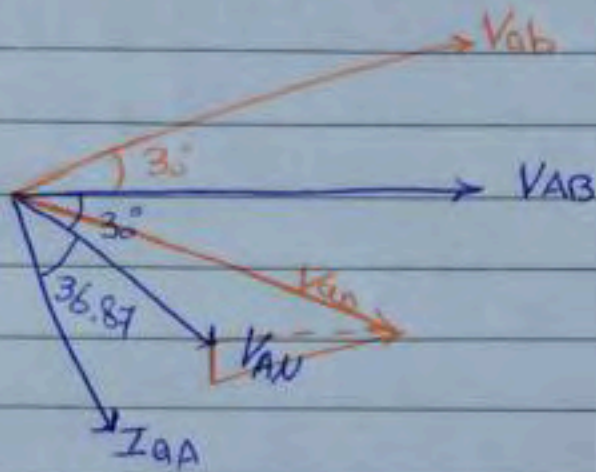
The phase Voltage of Phase - a
is!

$$\frac{V_{ab}}{\sqrt{3}} \angle -30^\circ = \frac{415 \angle 0^\circ}{\sqrt{3}} \angle -30^\circ = 240 \angle -30^\circ$$

$$V_{an} = 240 \angle -30^\circ + 69.5 \angle -66.87^\circ (0.2 + j1) = 296 \angle -20^\circ$$

$$V_{ab} = \sqrt{3} (296 \angle -20^\circ) \angle 30^\circ = 513.7 \angle 9.1^\circ$$

Phase - A :-

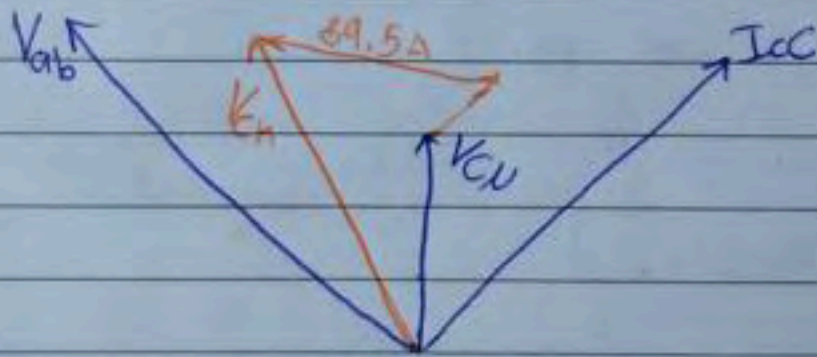


$$V_{bc} = 415 \angle -120$$

$$V_{bu} = 240 \angle -150$$

$$I_{bB} = 69.5 \angle -116.8$$

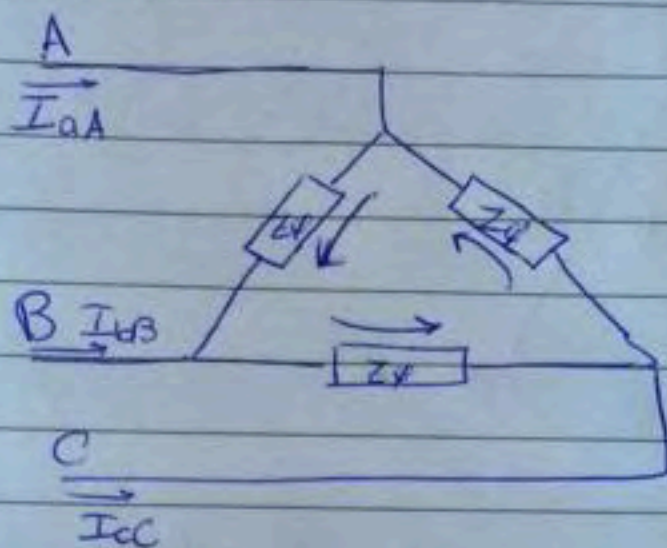
Phase - C :-



$$S = V_{an} I_{aA}^* + V_{bn} I_{bB}^* + V_{cn} I_{cC}^*$$

$$= \sqrt{3} V_L I_L = \sqrt{3} (513.7) (69.5) = 61.8 \text{ KVA}$$

Δ -load



KCL at A

$$I_{aA} = I_{AB} - I_{CA}$$

$$I_{cC} = I_{CA} - I_{AB}$$

$$I_{bB} = I_{BC} - I_{AB}$$

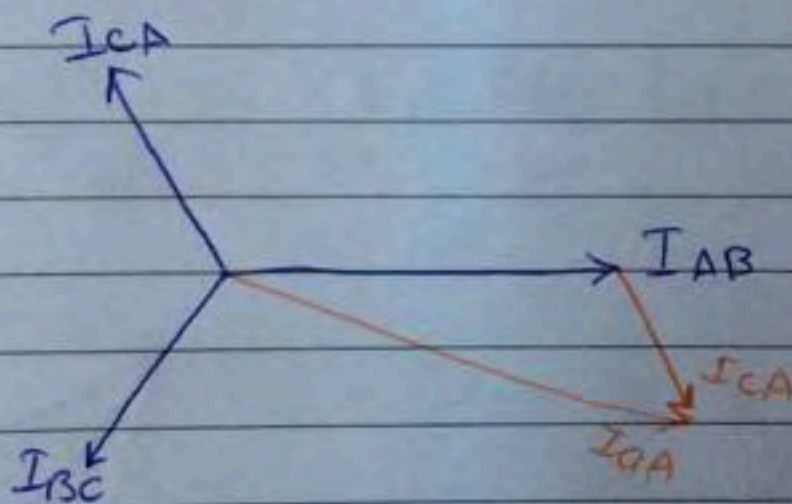
+ve

Sequence

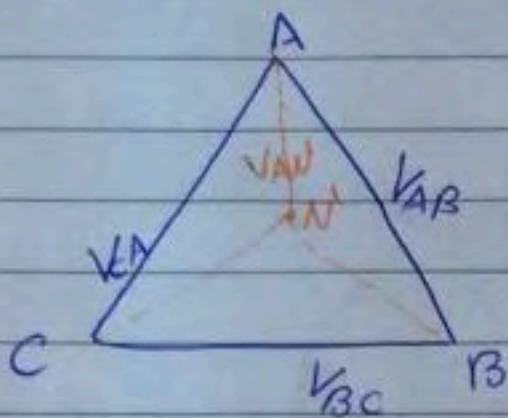
$$S_{\Delta} = 3 S_{\phi}$$

$$= 3 V_{\phi} I_{\phi}$$

$$= 3 V_{\phi} \frac{I_L}{\sqrt{3}} = \underline{\underline{\sqrt{3} V_L I_L}}$$

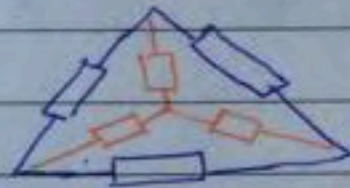


$\Delta \xrightarrow{T_0} 1\phi$ equivalent ckt.



\Rightarrow Keep in mind $\frac{V_L}{\sqrt{3}} \angle -30^\circ = V_{ph}$

$\Rightarrow Z_Y = \frac{Z_\Delta}{3}$



exp: Same as previous question but the load is:-

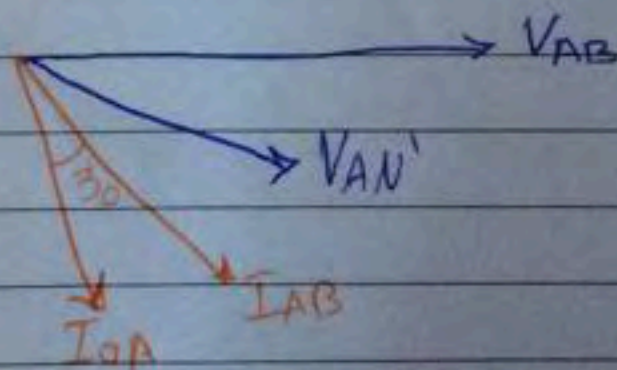
50 kVA

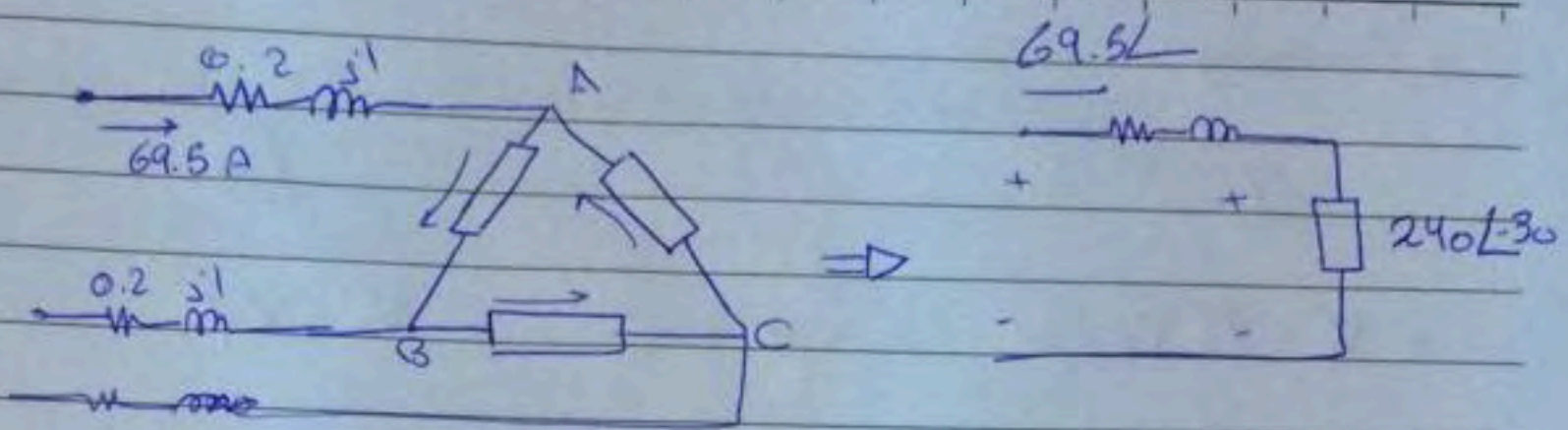
0.8 lag

Δ

$$S_D = 50000 = \sqrt{3} V_L I_L$$

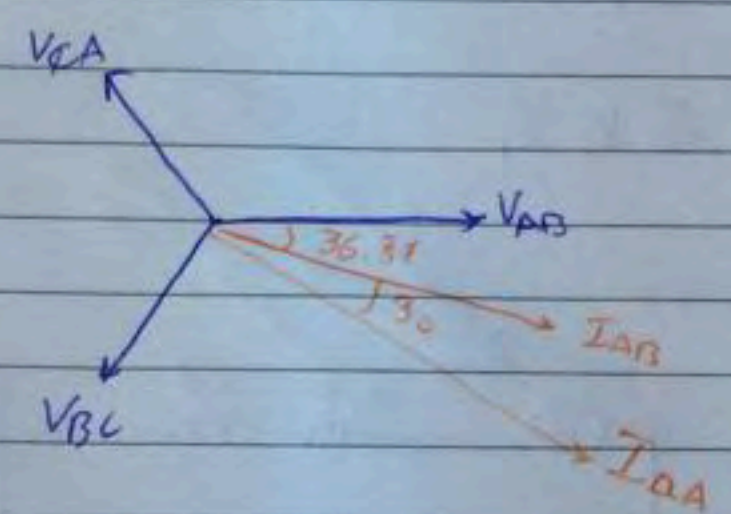
$$I_L = 69.5 \text{ A}$$





$$|I_L| = 69.5 \text{ A}$$

$$\phi = \phi_{V_{AB}} - \phi_{I_{AB}}$$



$$V_{AN} = 296 \angle -20^\circ$$

$$V_{ab} = \sqrt{3} (296 \angle -20^\circ) \angle 30^\circ = 512 \angle 10^\circ$$

$Z_{\Delta} ?$

$$S_{3\phi} = 3 V_{\phi} I_{\phi}^* = 3 V_{\phi} \frac{V_{\phi}^*}{Z_{\phi}^*}$$

$$Z_{\phi} = \frac{3 V_{\phi}^2}{S_{3\phi}^*} = \frac{3 \left(\frac{V_L}{\sqrt{3}} \right)^2}{S_{3\phi}^*} = \frac{V_L^2}{S^*} = \frac{415^2}{50 \text{ k}} = 3.5 \Omega$$

- Summary :-

Y

Δ

• $V_L = \sqrt{3} V_\phi \angle 30^\circ$ +ve seq.

• $V_L = V_\phi$

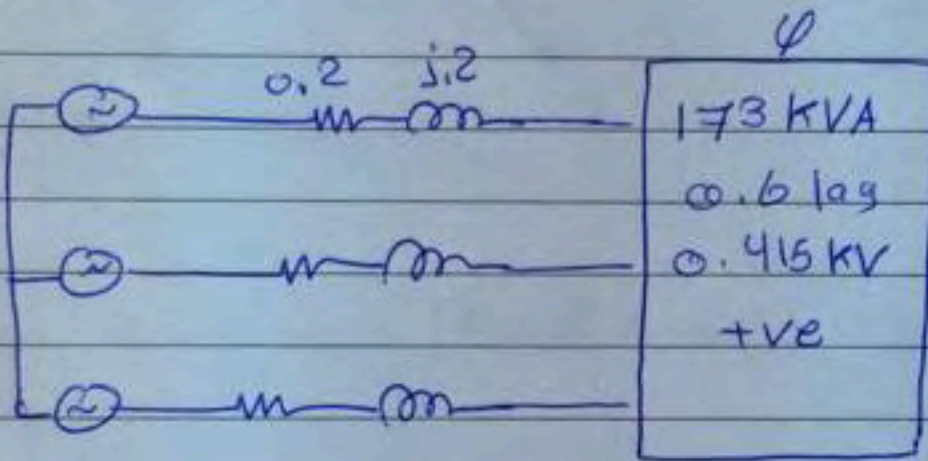
• $I_L = I_\phi$

• $I_L = \sqrt{3} I_\phi \angle -30^\circ$

• $S = \sqrt{3} V_L I_L$

• $S = \sqrt{3} V_L I_L$

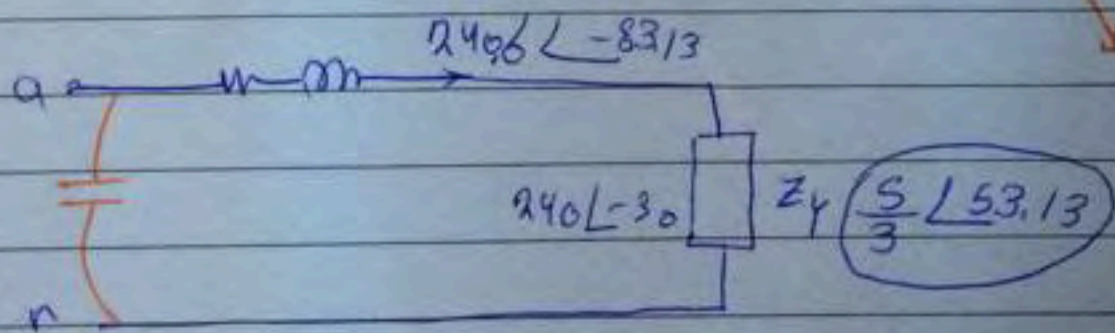
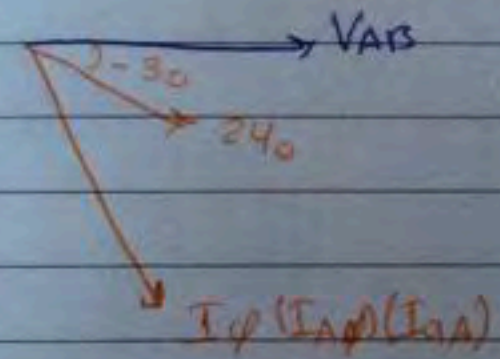
exp



Required to increase the PF to the value of 0.9 lag at source

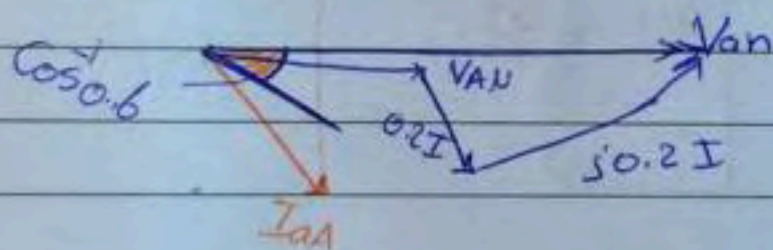
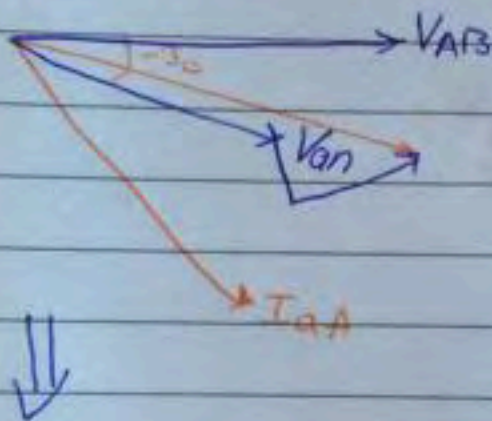
$$I_L = \frac{173}{0.415 \sqrt{3}} = 240.6$$

$$V_\phi I_\phi = \frac{173}{3}$$



$$V_{an} = 240 \angle -30 + (240 \cdot 6 \angle -83.13)(2+j.2)$$

$$= 307 \angle -31$$

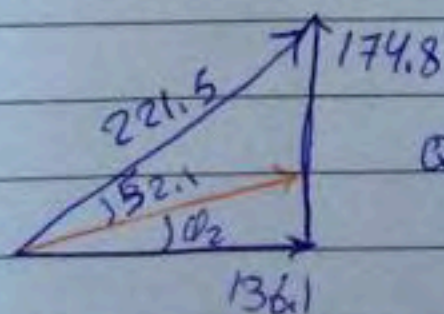


$$S_s = 3 V_{\phi} I_{\phi} = 307 \times 240 \cdot 6 \times 3$$

$$= 221.5 \text{ kVA}$$

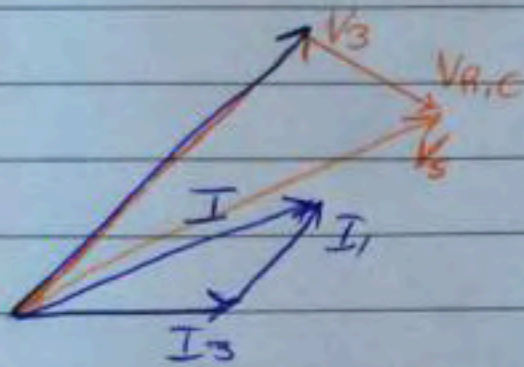
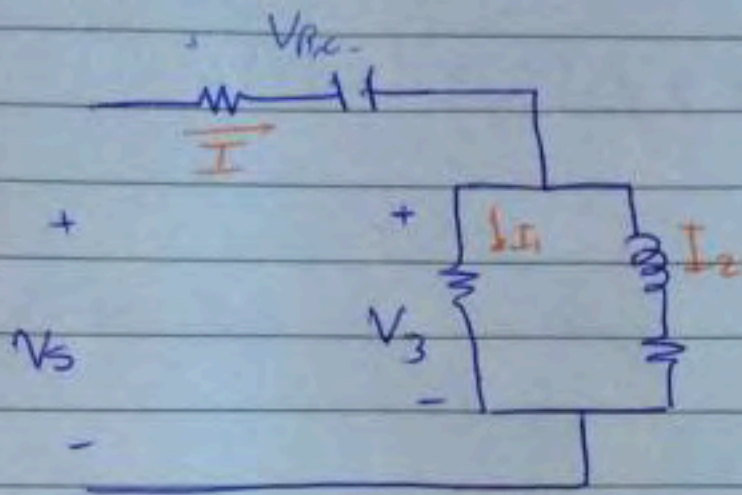
$$S = 221.5 \angle 52.1$$

$$\theta_s = \phi_{Vs} - \phi_{Is} = -31 - (-83.13) = 52.1^\circ$$

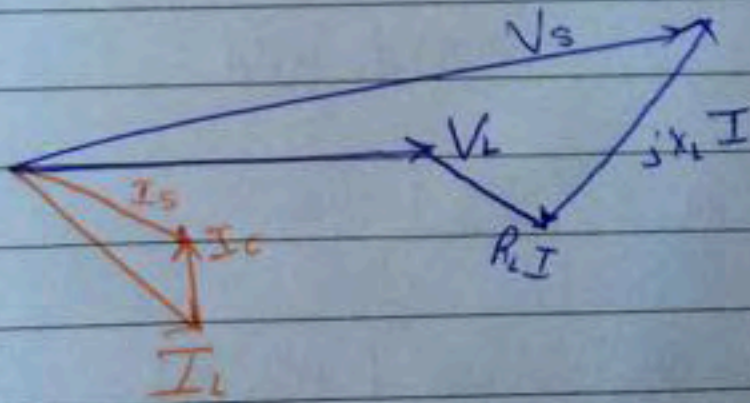
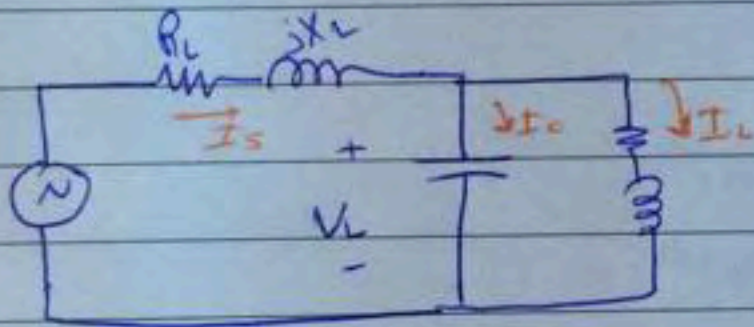


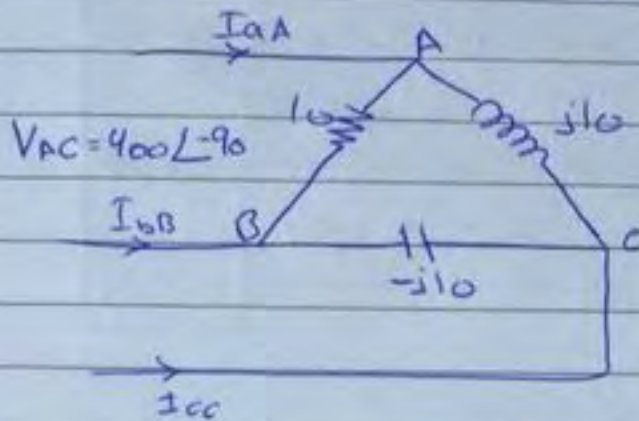
$$Q_c = 174.8 = 136.1 \tan \cos^{-1} 0.9$$

exp



exp



exp

Voltage balanced
load AB "

$$\text{Balanced} = |V| \quad \angle \theta = 120$$

$$I_{aA} = I_{AB} - I_{CA}$$

$$= \frac{V_{AB}}{Z_{AB}} - \frac{V_{CA}}{Z_{CA}} = \frac{400 \angle -30}{10} - \frac{400 \angle +90}{j10}$$

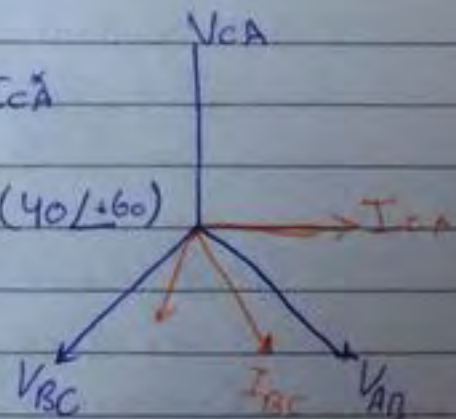
$$= 40 \angle -30 - 40 \angle 0 = 20.7 \angle -105$$

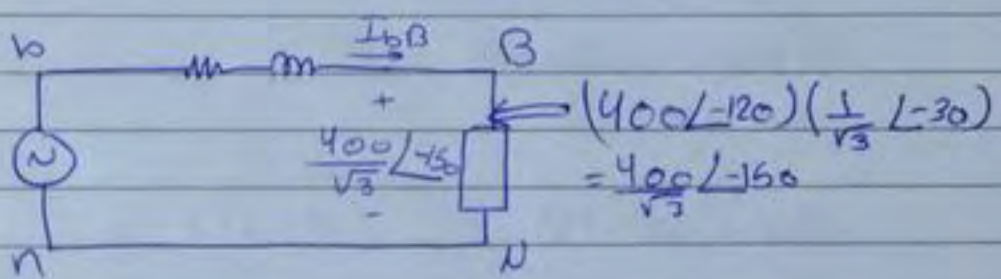
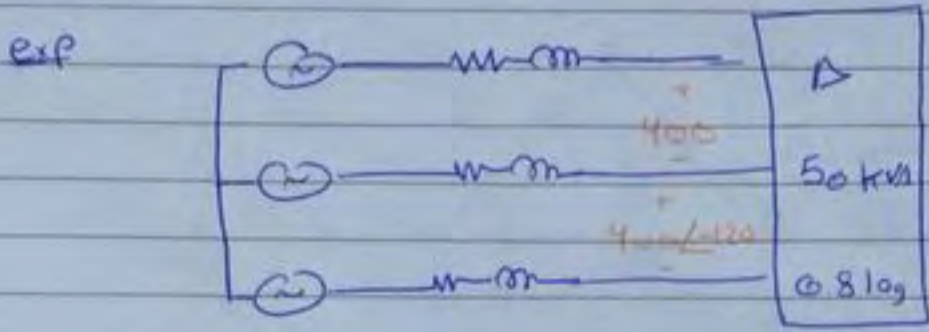
$$S = V_{AB} I_{AB}^* + V_{BC} I_{BC}^* + V_{CA} I_{CA}^*$$

$$= 400 \angle -30 (40 \angle 30) + 400 \angle -150 (40 \angle +60) + 400 \angle 90 (40 \angle 0)$$

$$P_{3\phi} = \frac{400^2}{10}$$

$$Q = \frac{V^2}{Z^*} = \frac{400^2}{-j10} = j1600$$



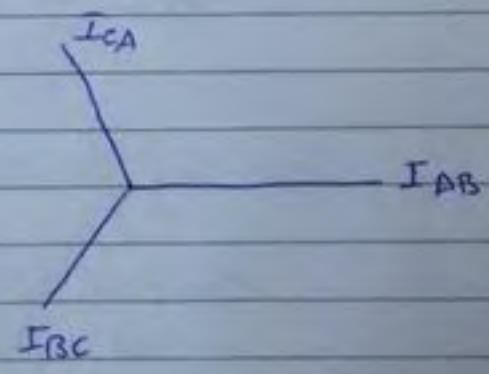


$$|I_{bB}| = |I_L|$$

$$I_L = \frac{|S_{3\phi}|}{\sqrt{3} V_L} = \frac{50 \text{ kW}}{\sqrt{3} \cdot 400} = 69.6 \text{ A}$$

The magnitude of the line current is

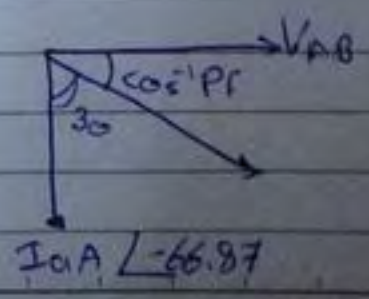
$$I_L = \frac{S_{3\phi}}{\sqrt{3} V_L}$$



To determine the angle of the line current

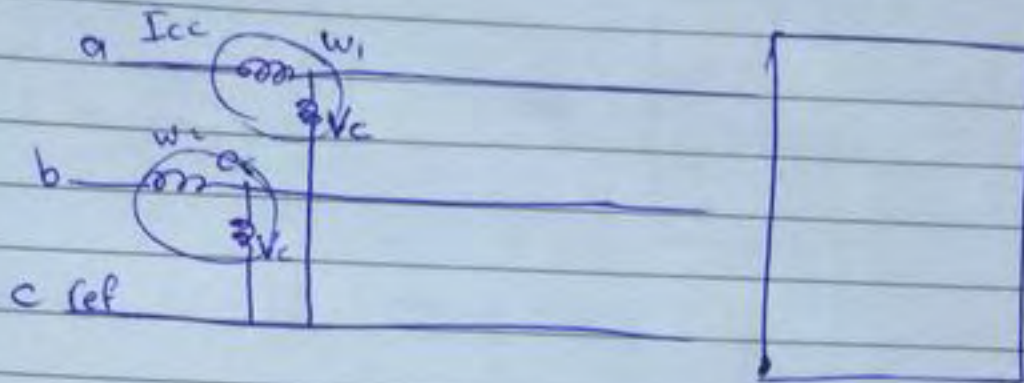
Δ -Connect $V_{\phi} = V_L$

I_{ϕ} (lag or lead)



②

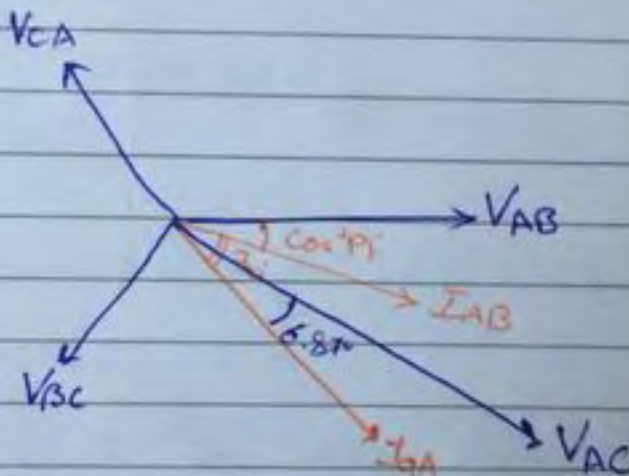
• Wattmeter:-



$$W_1 = |V| |I| \cos \theta \quad , \quad \theta = \phi_{V_{ac}} - \phi_{I_{aA}}$$

$$W_2 = |V| |I| \cos \theta \quad , \quad \theta = \phi_{V_{bc}} - \phi_{I_{bB}}$$

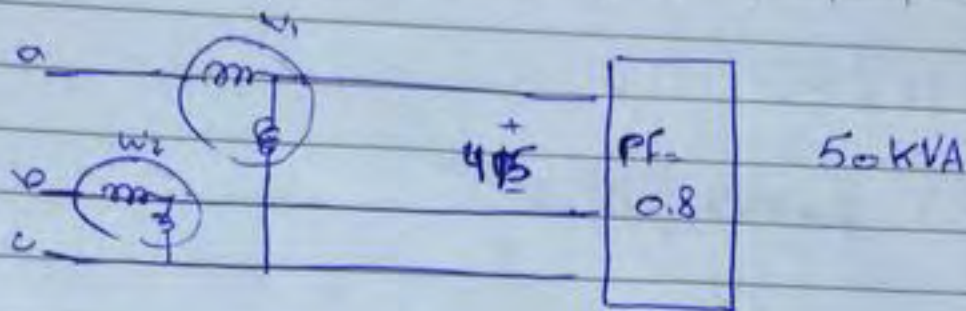
exp



$$W_1 + W_2 = \sqrt{3} V_L I_L \cos \theta = S$$

~~(W1 - W2)~~

$$(W_1 - W_2 = V_L I_L \sin \theta) = \frac{Q}{\sqrt{3}}$$



$$W_1 = |V_{AB}| |I_{aA}| \cos \theta_1$$

$$W_2 = |V_{BC}| |I_{bB}| \cos \theta_2$$

$$S = \sqrt{3} |V_L| |I_L|$$

$$P = \sqrt{3} |V_L| |I_L| \cos \phi$$

$$Q = \sqrt{3} |V_L| |I_L| \sin \phi$$

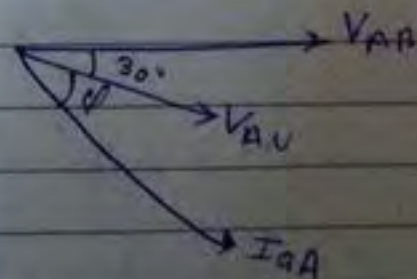
$$I_L = \frac{50000}{\sqrt{3} \times 400} = 69.5 \text{ A}$$

PF determine the angle btw V_p and I_p

$$W_1 = (415)(69.5) \cos(30 - 36.87) = 28.6 \text{ kW}$$

$$W_2 = (415)(69.5) \cos(30 + \phi) = 2.8 \text{ kW}$$

⇒ Sum of $w = w_1 + w_2$
 should = 40 kW
 (50 kW × 0.8 = 40 kVA)



$$W_1 - W_2 = 17.6 \text{ kW}$$

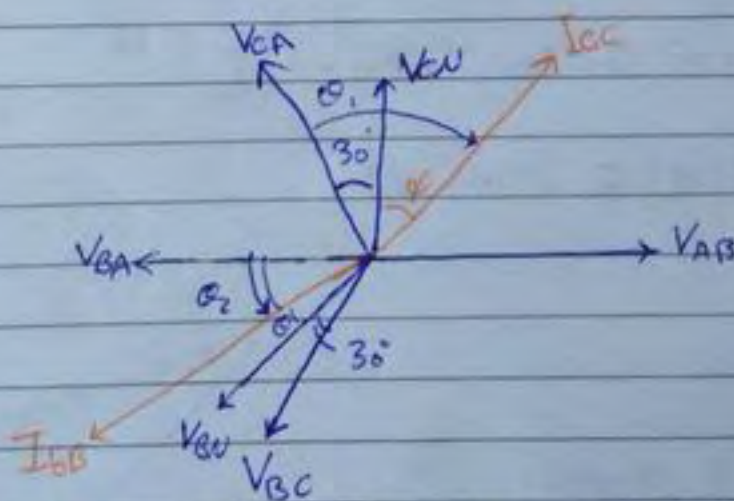
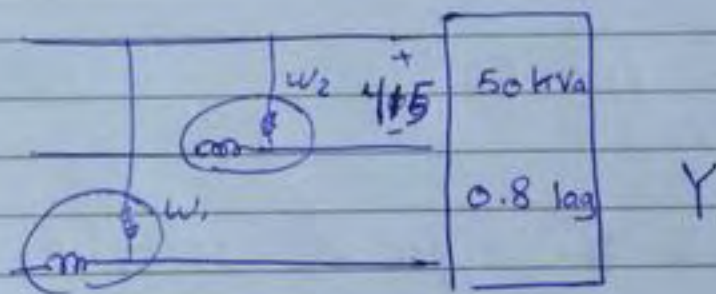
$$\sqrt{3}(W_1 - W_2) = 30 \text{ kVA} = Q_{3\phi}$$

$$\frac{W_1 - W_2}{W_1 + W_2} = \frac{V_L I_L \sin \theta}{\sqrt{3} V_L I_L \cos \theta}$$

$$= \frac{\tan \theta}{\sqrt{3}}$$

$$\phi = \tan^{-1} \frac{(W_1 - W_2) \sqrt{3}}{W_1 + W_2} = 36.87^\circ$$

exp

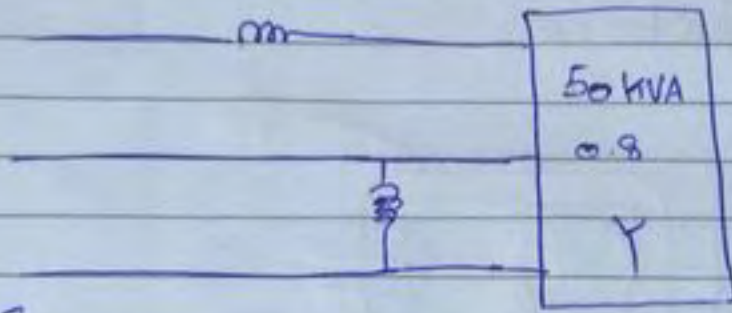


$$W_1 = |V_{CA}| |I_{c}| \cos(30 + \phi)$$

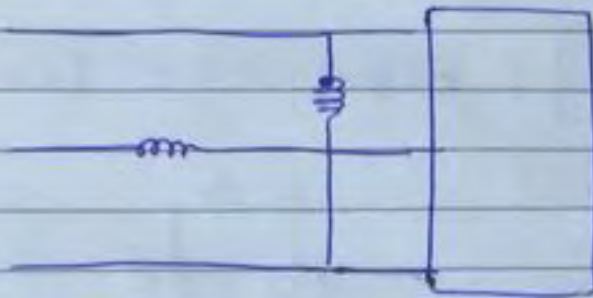
$$W_2 = |V_{BA}| |I_{b}| \cos(30 - \phi)$$

(5)

exp



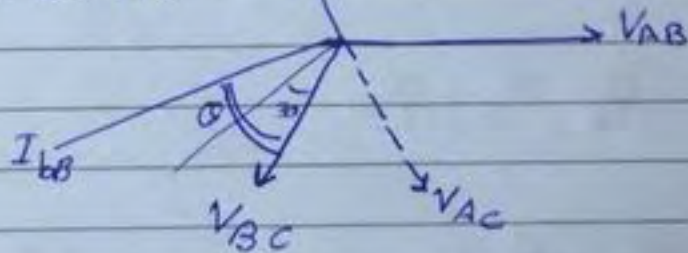
or

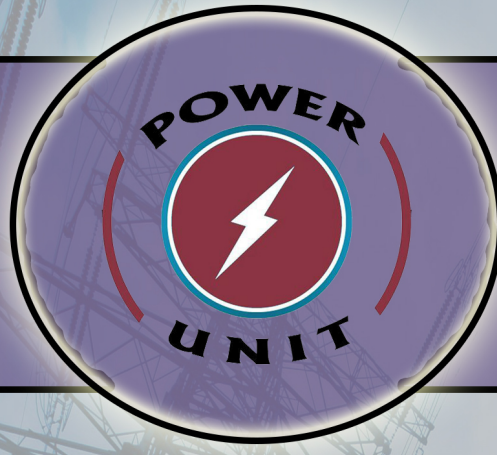


$$W = V_{AC} I_{LB} \cos \phi$$

$$= V_L I_L \cos(90^\circ + \phi) V_{CA}$$

$$= V_L I_L \sin(\phi)$$





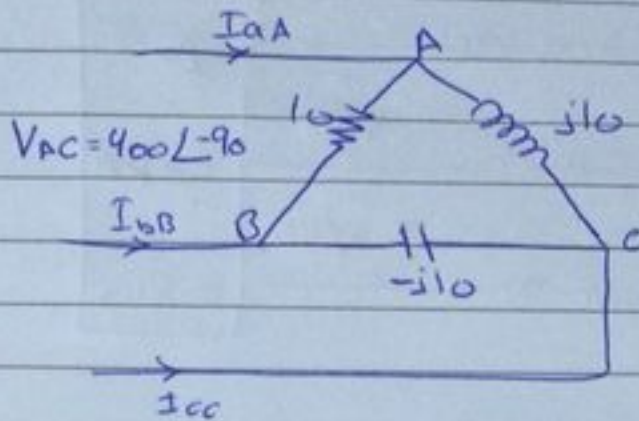
Circuits II Notebook

Dr. Nabeel Tawalbeh

By . Yazan Abawi

بأفكارنا نبدع

4th week

exp

Voltage balanced
load AB "

$$\text{Balanced} = |V| \quad \angle \theta = 120$$

$$I_{aA} = I_{AB} - I_{CA}$$

$$= \frac{V_{AB}}{Z_{AB}} - \frac{V_{CA}}{Z_{CA}} = \frac{400 \angle -30}{10} - \frac{400 \angle +90}{j10}$$

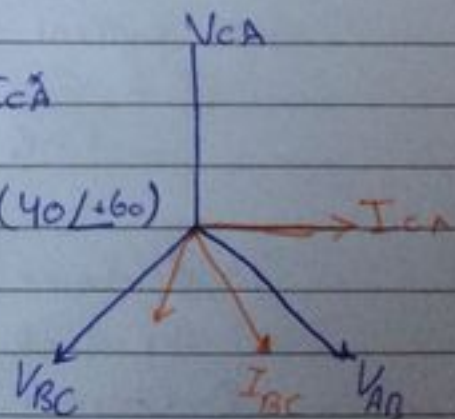
$$= 40 \angle -30 - 40 \angle 0 = 20.7 \angle -105$$

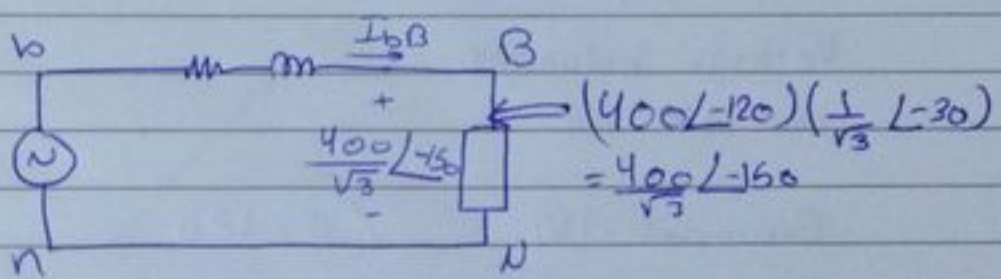
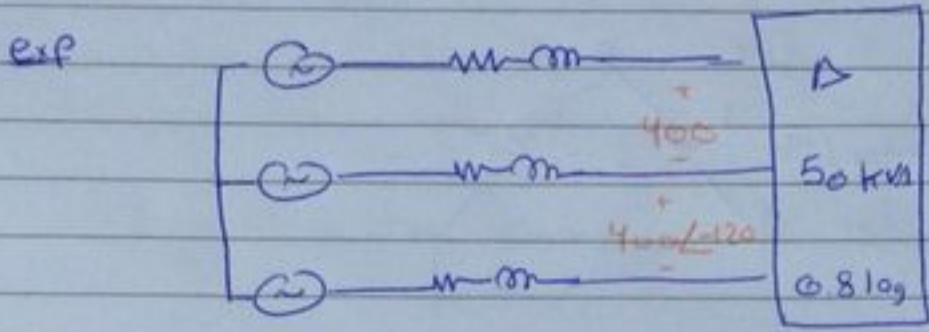
$$S = V_{AB} I_{AB}^* + V_{BC} I_{BC}^* + V_{CA} I_{CA}^*$$

$$= 400 \angle -30 (40 \angle 30) + 400 \angle -150 (40 \angle +60) + 400 \angle 90 (40 \angle 0)$$

$$P_{3\phi} = \frac{400^2}{10}$$

$$Q = \frac{V^2}{Z^*} = \frac{400^2}{-j10} = j1600$$



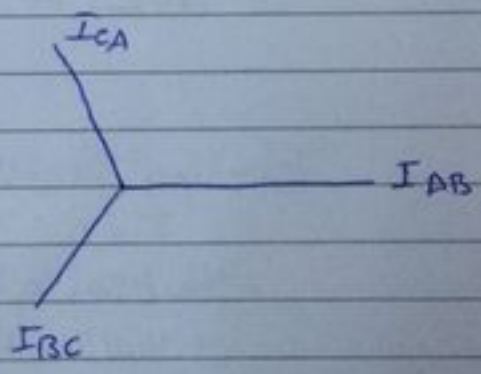


$$|I_{bB}| = |I_L|$$

$$I_L = \frac{|S_{3\phi}|}{\sqrt{3} V_L} = \frac{50 \text{ kW}}{\sqrt{3} \cdot 400} = 69.6 \text{ A}$$

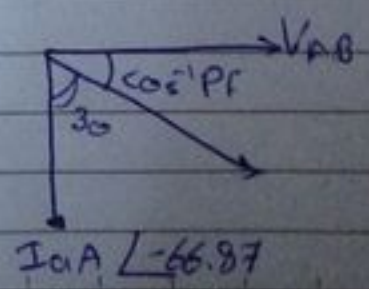
The magnitude of the line current is

$$I_L = \frac{S_{3\phi}}{\sqrt{3} V_L}$$



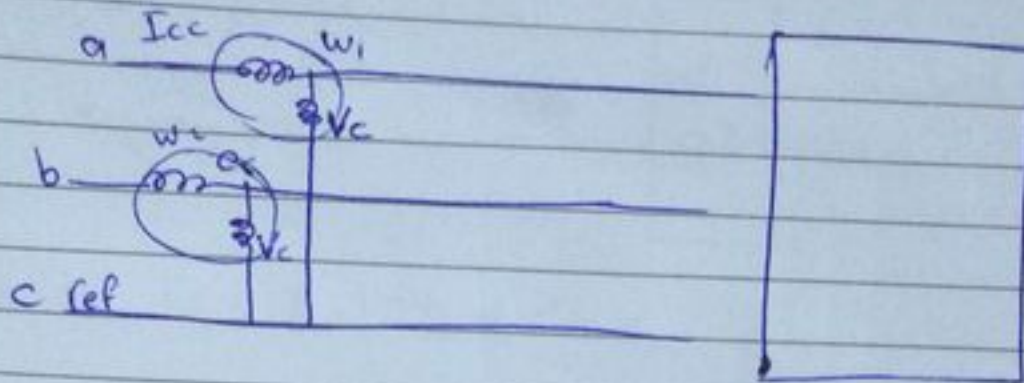
To determine the angle of the line current

Δ -Connect $V_{\phi} = V_L$
 I_{ϕ} (lag or lead)



②

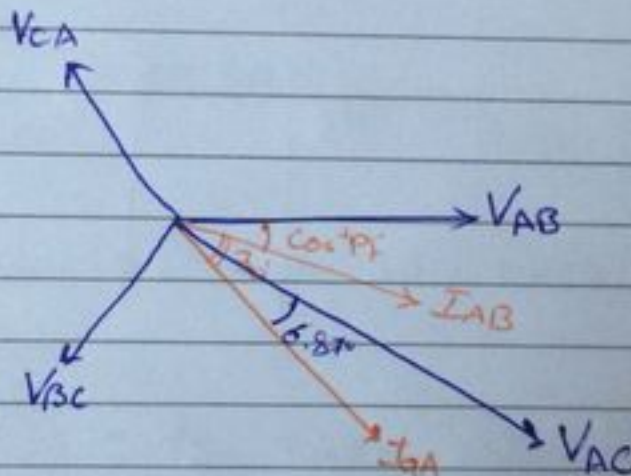
• Wattmeter:-



$$W_1 = |V| |I| \cos \theta, \quad \theta = \phi_{VAC} - \phi_{IaA}$$

$$W_2 = |V| |I| \cos \theta, \quad \theta = \phi_{VBC} - \phi_{IbB}$$

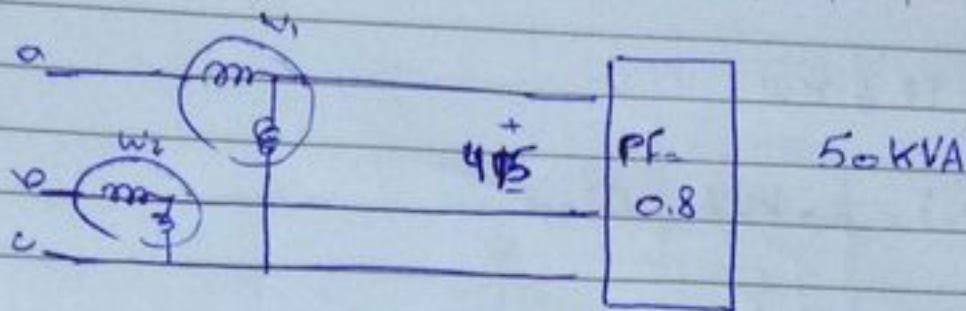
exp



$$W_1 + W_2 = \sqrt{3} V_L I_L \cos \theta = S$$

~~(W1 - W2)~~

$$(W_1 - W_2 = V_L I_L \sin \theta) = \frac{Q}{\sqrt{3}}$$



$$W_1 = |V_{AB}| |I_{aA}| \cos \theta_1$$

$$W_2 = |V_{BC}| |I_{bB}| \cos \theta_2$$

$$S = \sqrt{3} |V_L| |I_L|$$

$$P = \sqrt{3} |V_L| |I_L| \cos \phi$$

$$Q = \sqrt{3} |V_L| |I_L| \sin \phi$$

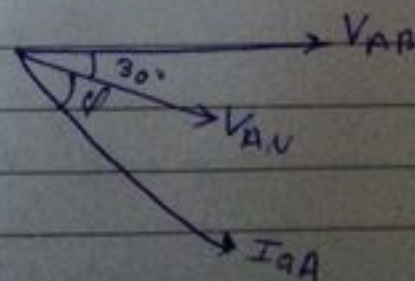
$$I_L = \frac{50000}{\sqrt{3} \times 400} = 69.5 \text{ A}$$

PF determine the angle btw V_p and I_p

$$W_1 = (415)(69.5) \cos(30 - 36.87) = 28.6 \text{ kW}$$

$$W_2 = (415)(69.5) \cos(30 + \phi) = 21.8 \text{ kW}$$

\Rightarrow Sum of $w = w_1 + w_2$
 should = 40 kW
 (50 kW \times 0.8 = 40 kVA)



$$W_1 - W_2 = 17.6 \text{ kW}$$

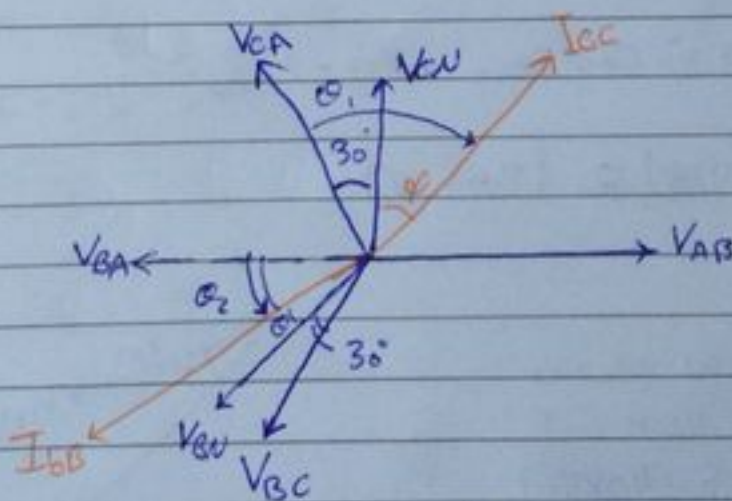
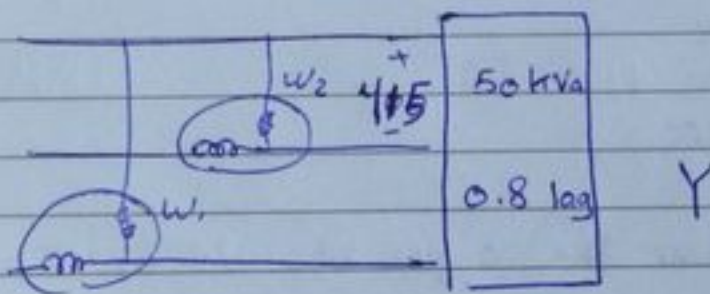
$$\sqrt{3}(W_1 - W_2) = 30 \text{ kVA} = Q_{3\phi}$$

$$\frac{W_1 - W_2}{W_1 + W_2} = \frac{V_L I_L \sin \theta}{\sqrt{3} V_L I_L \cos \theta}$$

$$= \frac{\tan \theta}{\sqrt{3}}$$

$$\phi = \tan^{-1} \frac{(W_1 - W_2) \sqrt{3}}{W_1 + W_2} = 36.87^\circ$$

exp

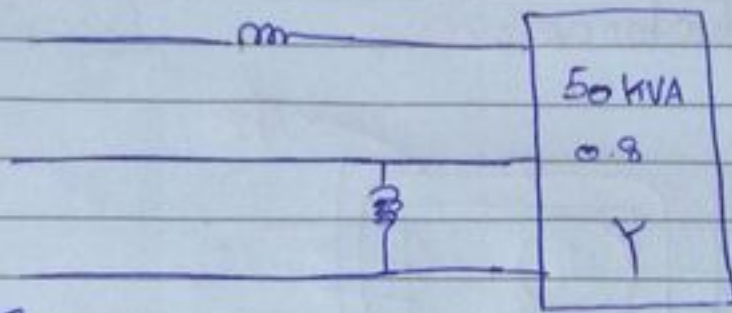


$$W_1 = |V_{AN}| |I_{AC}| \cos(30 + \phi)$$

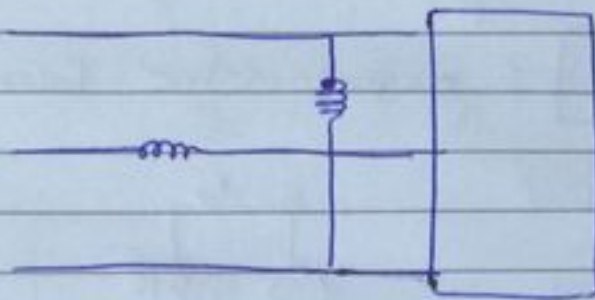
$$W_2 = |V_{BA}| |I_{BC}| \cos(30 - \phi)$$

(5)

exp



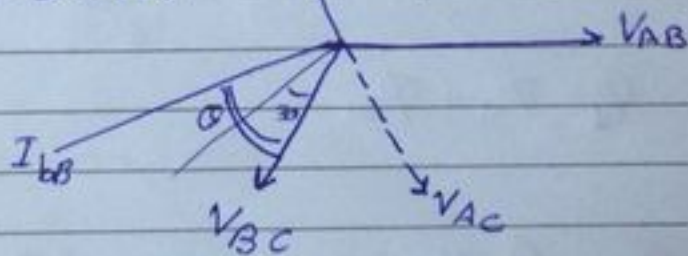
or



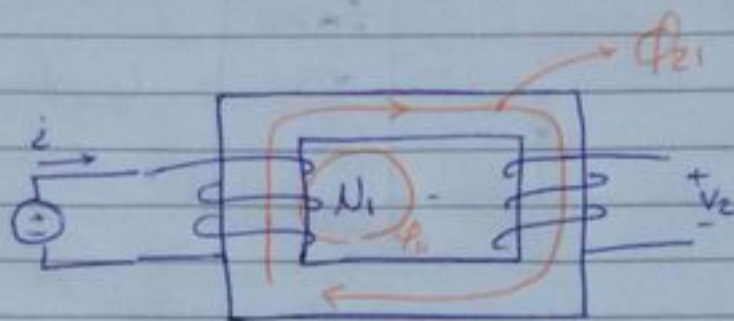
$$W = V_{AC} I_{LB} \cos \phi$$

$$= V_L I_L \cos(90^\circ + \phi) V_{CA}$$

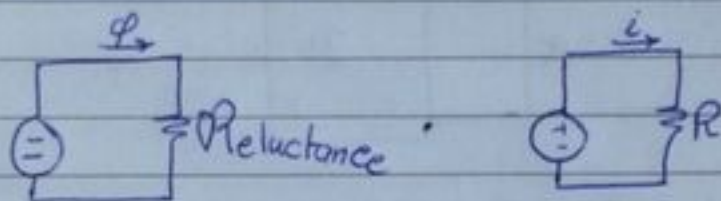
$$= V_L I_L \sin(\phi)$$



Mutual Conductance :-



$$i N = \text{mmf} \quad ; \quad \text{magnetomotive Force}$$



MMF

EMF

Φ_{11} : leakage Flux

Φ_{21} : Mutual Flux

$$\Phi_1 = \Phi_{11} + \Phi_{21}$$

$$\Phi_{11} = P_{11} N_1 i$$

P_{11} : Reluctance of the material occupied by P_1

$$\Phi_{21} = P_{21} N_1 i$$

$$\Phi_1 = P_1 N_1 i$$

$$\therefore \boxed{P_1 = P_{11} + P_{21}}$$

$$V = \frac{\partial \psi}{\partial t}$$

ψ : Flux linkage

$$\psi = N \phi$$

$$V = \frac{\partial N \phi}{\partial t} = N \frac{\partial \phi}{\partial t}$$

$$V_1 = N_1 \frac{d\phi}{dt} = N_1 \frac{\partial}{\partial t} (\mu_r N_1 i_1) = N_1^2 \mu_r \frac{d i_1}{dt} = L \frac{d i_1}{dt}$$

$$V_1 = L_1 \frac{d i_1}{dt}$$

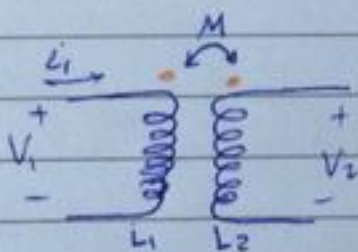
L_1 : Self Voltage

$$V_2 = N_2 \frac{d\phi_{21}}{dt} = N_2 \frac{d}{dt} (\mu_{r1} N_1 i_1) = N_2 N_1 \mu_{r1} \frac{d i_1}{dt}$$

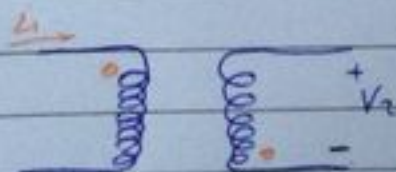
$$V_2 = M_{21} \frac{d i_1}{dt}$$

M_{21} : mutual Voltage

• Dot Convention:-



$$V_2 = M \frac{d i_1}{dt}$$



$$V_2 = -M \frac{d i_1}{dt}$$

$$L_1 = P_1 N_1^2$$

$$L_2 = P_2 N_2^2$$

$$L_1 L_2 = P_1 N_1^2 N_2^2 P_2 = (P_{11} + P_{21}) M^2 = (P_{22} + P_{12}) N_2^2$$

$$= N_1^2 P_{21} \left(1 + \frac{P_{11}}{P_{12}}\right) N_2^2 P_{12} \left(1 + \frac{P_{22}}{P_{12}}\right)$$

$$= N_1^2 N_2^2 P_{12}^2 \left(1 + \frac{P_{11}}{P_{12}}\right) \left(1 + \frac{P_{22}}{P_{12}}\right)$$

$$= N_1^2 N_2^2 P_{12}^2 \frac{1}{k^2} = M^2$$

$$L_1 L_2 = M^2 \frac{1}{k^2}$$

$$\boxed{M = k \sqrt{L_1 L_2}}$$

k : Coupling Factor

$$\rightarrow \text{at } P_{11} = 0 \rightarrow k = 1$$

$$\rightarrow \text{at } P_{12} = 0 \rightarrow k = 0$$

(a)

exp



$$E_1 = L_1 \frac{d(i_1 - i_2)}{dt} + L_3 \frac{d(i_1 - i_3)}{dt}$$

Self Voltage

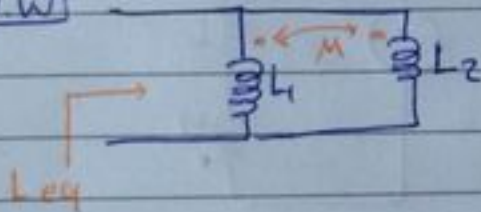
$$+ M_1 \frac{d i_2}{dt} + M_2 \frac{d(i_1 - i_3)}{dt}$$

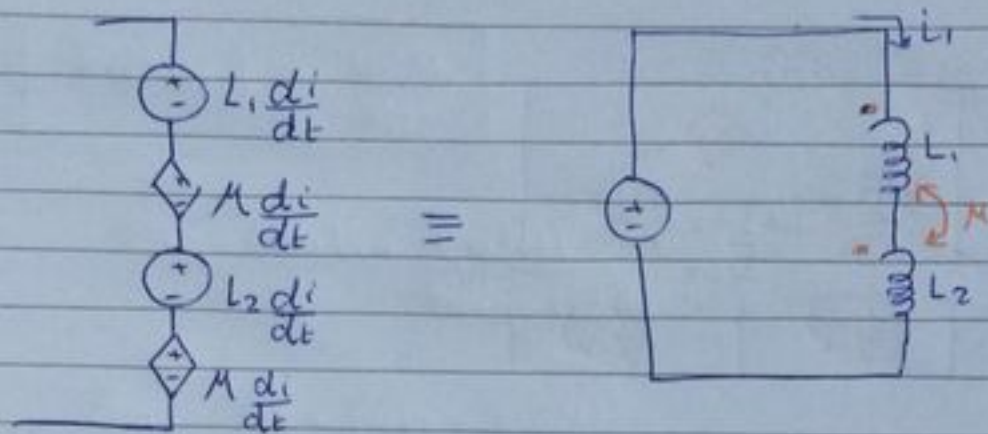
Mutual Voltage

$$+ M_3 \frac{d(i_1 - i_3)}{dt} + M_3 \frac{d(i_2 - i_3)}{dt}$$

⋮

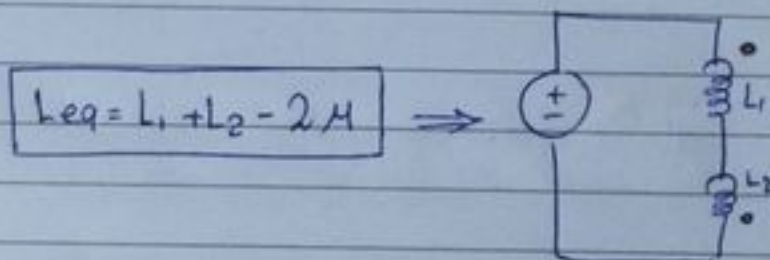
H.W



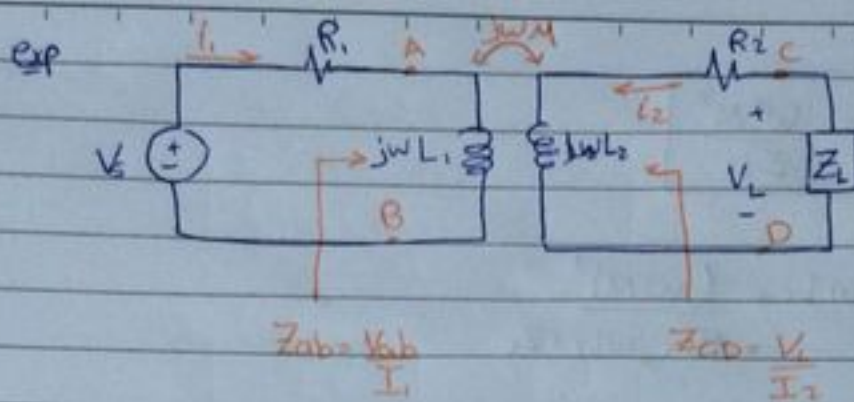


$$E_1 = \left(L_1 \frac{di}{dt} + L_2 \frac{di}{dt} + M \frac{di}{dt} + M \frac{di}{dt} \right)$$

$\boxed{L_{eq} = L_1 + L_2 + 2M}$ في حالة التسلسل



①



$$V_s = \underbrace{(R_1 + j\omega L_1)}_{Z_{11}} I_1 + j\omega M I_2$$

$$0 = \underbrace{(R_2 + j\omega L_2 + Z_L)}_{Z_{22}} I_2 + j\omega M I_1$$

$$V_s = Z_{11} I_1 + j\omega M I_2$$

$$0 = j\omega M I_1 + Z_{22} I_2$$

$\therefore Z_{11}$ loop impedance in ① - forward

Z_{22} loop impedance in ② - backward

$$I_1 = \frac{\begin{vmatrix} V_s & j\omega M \\ 0 & Z_{22} \end{vmatrix}}{\begin{vmatrix} Z_{11} & j\omega M \\ j\omega M & Z_{22} \end{vmatrix}} = \frac{Z_{22}}{Z_{11} Z_{22} + \omega^2 M^2} V_s$$

$$Z_{in} = \frac{V_s}{I_1} = \frac{Z_{11} Z_{22} + \omega^2 M^2}{Z_{22}} = \boxed{\frac{Z_{11} + \omega^2 \frac{M^2}{Z_{22}}}{1}}$$

$$\omega M = j\omega M = |Z_M|$$

$$Z_M = j\omega M = jX_M$$

$$Z_{in} = Z_{p.f.w} + Z_{reflected}$$

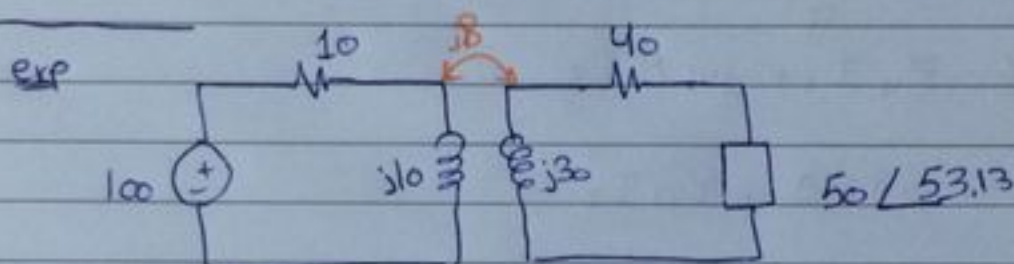
(12)

$$Z_{\text{reflected}} = \frac{\omega^2 M^2}{Z_{22}}$$

$$Z_{AB} = j\omega L_1 + \frac{(\omega M)^2}{R_2 + j\omega L_2 + Z_L}$$

$$Z_{CO} = R_2 + j\omega L_2 + \frac{(\omega M)^2}{R_1 + j\omega L_1}$$

* All power independent sources are zeroing



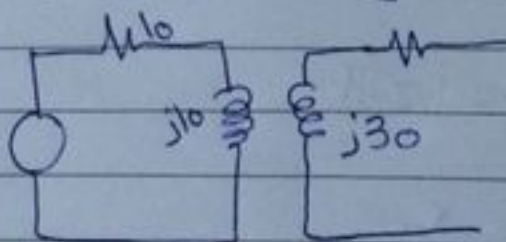
Find I_1, I_2 ?

$$I_1 = \frac{100}{Z_s} = \frac{100}{10 + j10 + \frac{64}{70 + j30}} = \boxed{7.1 / -42}$$

$$0 = (70 + j30)I_2 = 8(I_1)$$

$$I_2 = \boxed{5.7 / 2.6}$$

Find I_2 without using Mesh (use Thevenin)



13

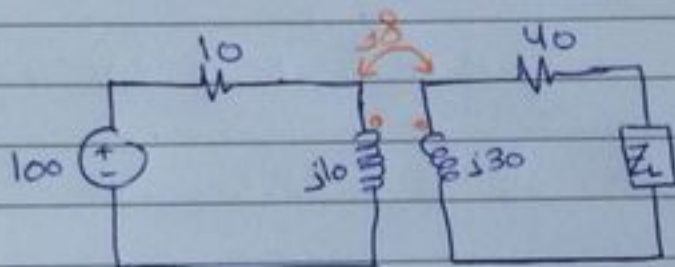
Kill the source

$$Z_{th} = 40 + j30 + \frac{64}{10 + j10}$$

$$V_{th} = V_{oc} = j8 I_1 = j8 \left(\frac{100}{10 + j10} \right) = \boxed{56.5 \angle 45^\circ}$$

$$I_2 = \frac{V_{th}}{Z_L + Z_{th}} = \frac{56.5 \angle 45^\circ}{50 \angle 52.13^\circ + 50.8 \angle 31^\circ} = \boxed{0.57 \angle 2.6^\circ}$$

* Take home exam



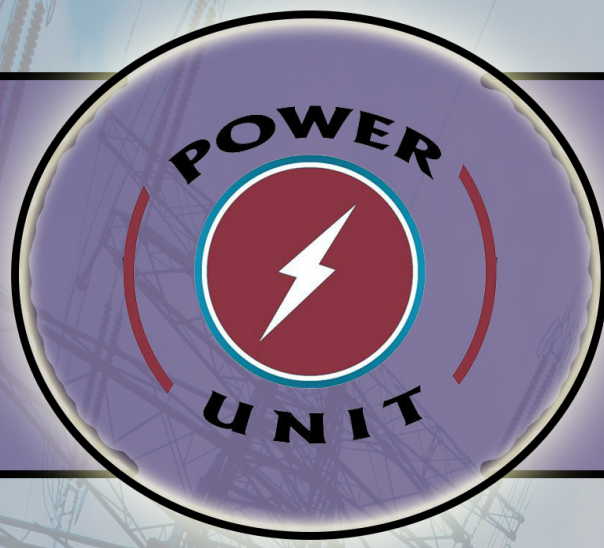
Find the Z looking by the source if the Z_L is adjusted that make maximum power

$$Z = 10 + j10 + \frac{64}{40 + j30 + Z_{th}} \rightarrow \text{seen by } Z_L \text{ (terminal)}$$

$$Z_{th} = 40 + j30 + \frac{64}{10 + j10} = 50 \angle 32^\circ$$

$$Z = 10 + j10 + \frac{64}{40 + j30 + 50 \angle -32^\circ}$$

14



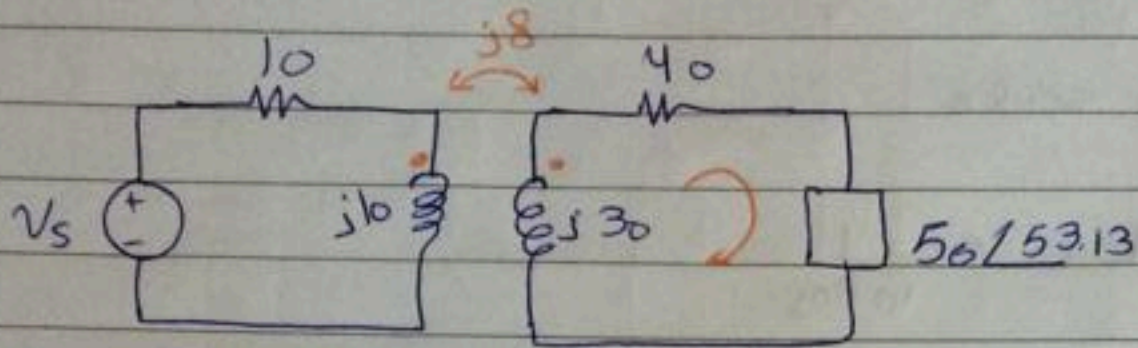
Circuits II Notebook

Dr. Nabeel Tawalbeh

By . Yazan Abawi

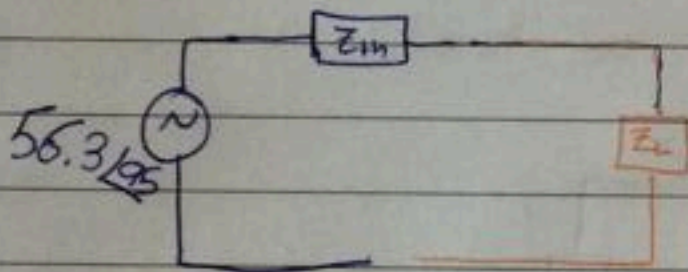
بِأَفْكَارِنَا نَبْدَعُ

5th week



- Find Z seen by the source terminal if Z_L is adjusted

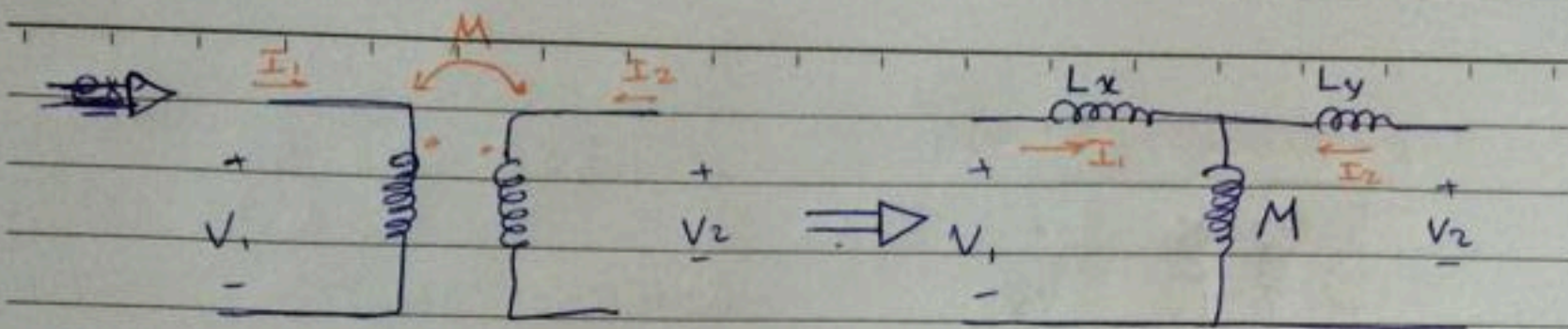
$$Z_{th} = 50.8 \angle -31^\circ$$



$$Z_s = 10 + j10 + \frac{64}{40 + j30 + 50.8 \angle -31^\circ}$$

$$R_o = \left| \frac{40 + 30j + \frac{64}{10 + j10}}{10 + j10} \right| = |50.8 \angle -31^\circ| = 50.8$$

$$Z_s = 10 + j10 + \frac{64}{40 + j30 + 50.8}$$



$$* V_1 = j\omega L_2 I_1 + j\omega M I_2$$

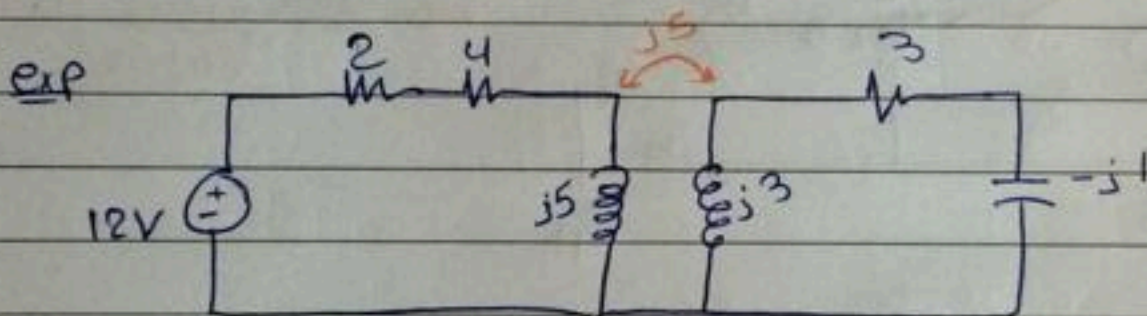
$$* V_2 = j\omega L_2 I_2 + j\omega M I_1$$

$$* V_1 = j\omega(L_x + LM)I_1 + j\omega M I_2$$

$$* V_2 = j\omega L_m I_1 + j\omega M(LM + L_y)I_2$$

$$\Rightarrow LM = M$$

$$L_x + M = L_1 \rightarrow L_x = L_1 - M$$



$$Z_s, Z_{ab}, Z_r$$

$$Z_s = 2 + 4 + j6 + \frac{25}{j3 - j}$$

mag of ref.

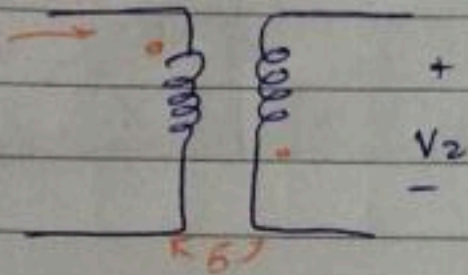
$$Z_{th} = 3 + j3 + \frac{25}{6 + j6} \quad , \quad V_{th} = \left(\frac{12}{6 + j6} \right) \cdot j5$$

$$Z_{ab} = Z_s - 2$$

$$Z_r = \frac{25}{3 + j2}$$

(2)

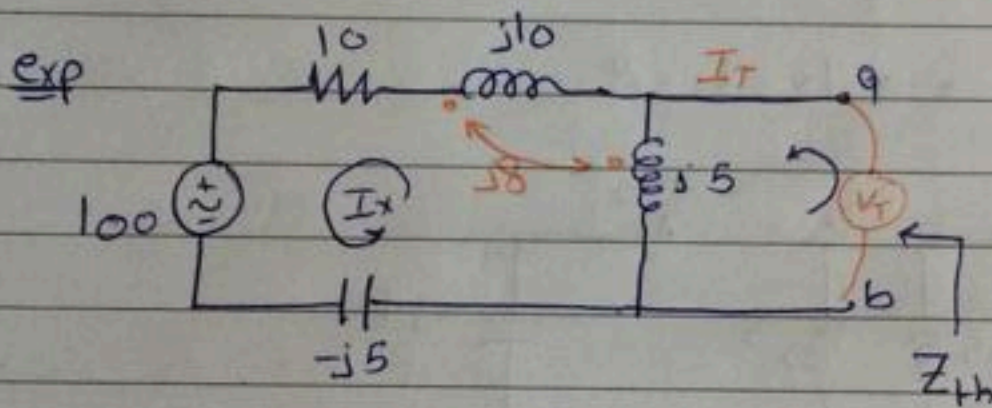
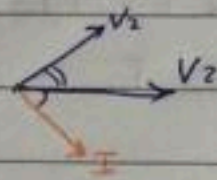
$$i = 10 \cos(100\pi t + 30^\circ)$$



$$V_2 = -M \frac{d}{dt} 10 \cos(100\pi t - 30^\circ)$$

$$= +5 \times 10 \times 100\pi \cos(100\pi t + -120^\circ)$$

$$V = 500\pi / 60^\circ$$



$$\Rightarrow V_{ab} = \frac{(100)}{(26j + 10)} (j13)$$

$$V_{ab} = I j5 + I j8$$

$$100 = (10 + j10 + j5 - 5) I + j8 I + j8 I$$

$$I = \frac{100}{10 + 26j}$$

$$V_T = j5 I_T - j5 I_x - j8 I_x$$

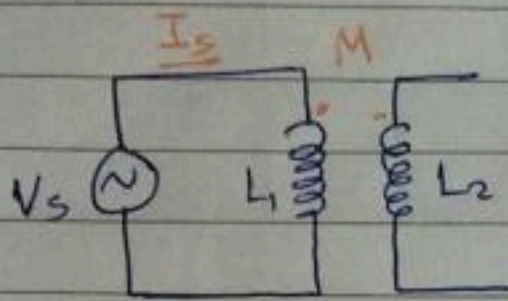
$$0 = (10 + j10) I_x - j5 I_T - j8(I_T - I_x) - j8 I_x$$

(8)

$$= 5 I_T - j13 I_V = V_T$$

$$-j13 I_T + (10 + j18) I_V = 0$$

$$I_T = \frac{\begin{vmatrix} V_T & -j13 \\ 0 & 10 + j18 \end{vmatrix}}{\begin{vmatrix} j5 & -j13 \\ -j13 & 10 + j18 \end{vmatrix}} = \frac{(10 + j18) V_T}{T}$$



$$W_1 = \int_0^{I_1} P dt = \int_0^{I_1} L_1 i_1 di_1 = \boxed{\frac{1}{2} L_1 i_1^2}$$

$$\boxed{P_1 = V_1 i_1} = L_1 \frac{di_1}{dt}$$

$$P_2 = V_2 i_2$$

$$W_2 = \int P dt = \int L_2 \frac{di_2}{dt} i_2 = \int_0^{I_2} L_2 di_2$$

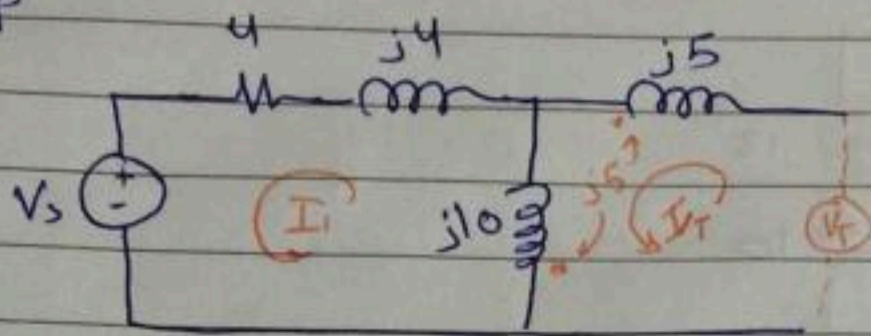
$$V_2 = i_2 \frac{di_2}{dt}, \quad V = M \frac{di_1}{dt}$$

$$I_1 V_2 = i_1 M \frac{di_2}{dt}$$

$$W = \int_0^{I_2} i_1 M \frac{di_2}{dt} dt = M I_1 I_2$$

$$\boxed{W = \frac{1}{2} L_1 i_1^2 + \frac{1}{2} L_2 i_2^2 + M I_1 I_2} \leftarrow \text{Energy}$$

exp



$$V_T = (j5 + j10) I_T - j10 I_1 + j5(I_1 - I_T) - j5 I_T$$

$$\boxed{V_T = j5 I_T - j5 I_1} \quad \text{--- (1)}$$

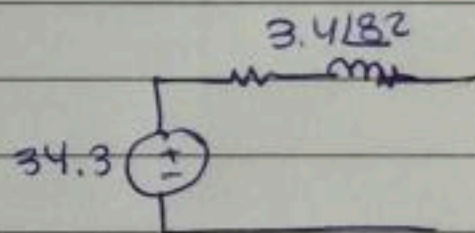
$$-100 = (4 - j4) I_1 - j10 I_T + j5 I_T$$

$$\boxed{-100 = -j I_T + (4 + j4) I_1} \quad \text{--- (2)}$$

$$I_T = \frac{\begin{pmatrix} V_T & -j5 \\ 0 & (4 + j4) \end{pmatrix}}{\begin{pmatrix} j5 & -j5 \\ -j5 & 4 + j14 \end{pmatrix}} = \frac{(4 + j14) V_T}{(-70 + j20) + 25}$$

$$Z_{Th} = \frac{V_T}{I_T} = \frac{-45 + j20}{4 + j14} = 3.4 \angle 82^\circ$$

$$I_1 = \frac{100}{4 + j14} = 6.9 \angle -74^\circ$$



$$i(t) = 6.9 \sqrt{2} \cos(100\pi t - 74^\circ)$$

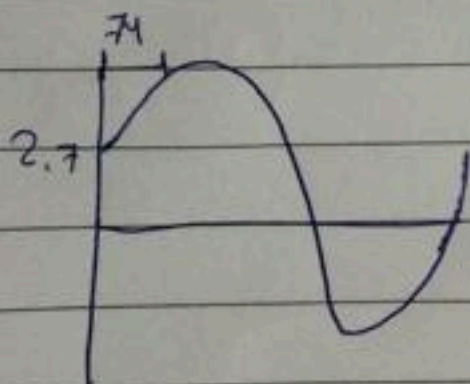
$$i\left(\frac{1}{4}\right) = i(5 \text{ ms}) = 4.38 \text{ A}$$

$$W = \frac{1}{2} L_1 i_1^2 + \frac{1}{2} L_2 i_2^2 - M i_1 i_2$$

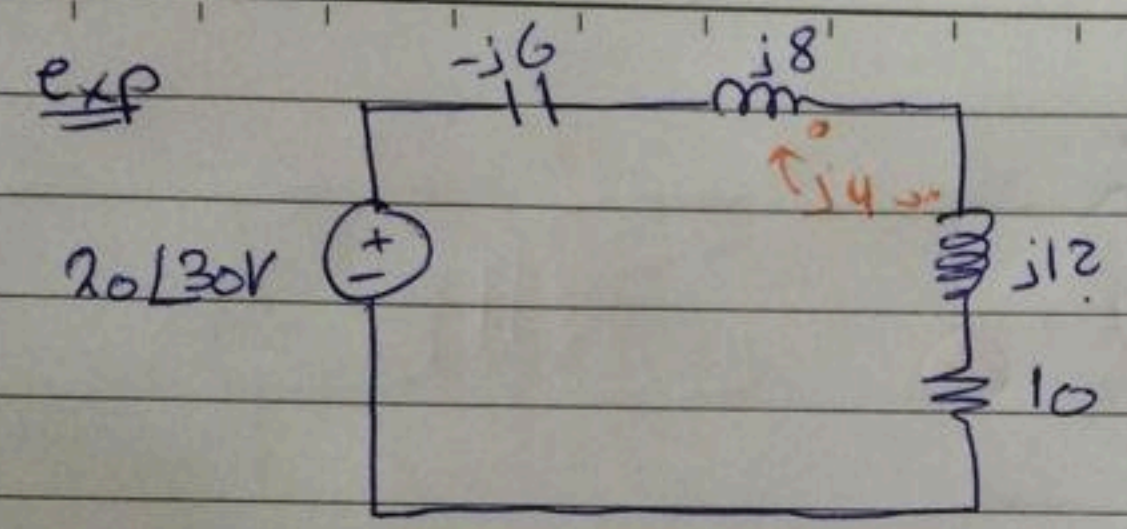
$$L_1 = \frac{X}{\omega} = \frac{10}{100\pi}$$

$$L_2 = \frac{X}{\omega} = \frac{5}{100\pi}$$

$$M = \frac{S}{\omega}$$

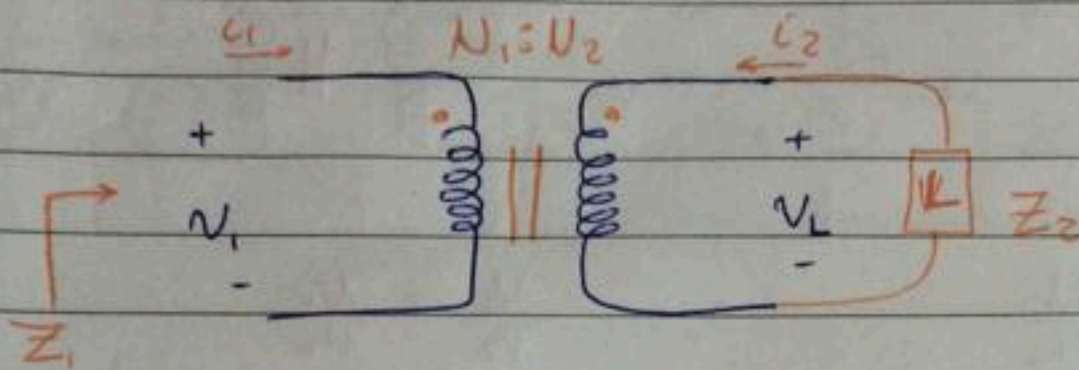


(6)



(7)

Ideal Transformer :-

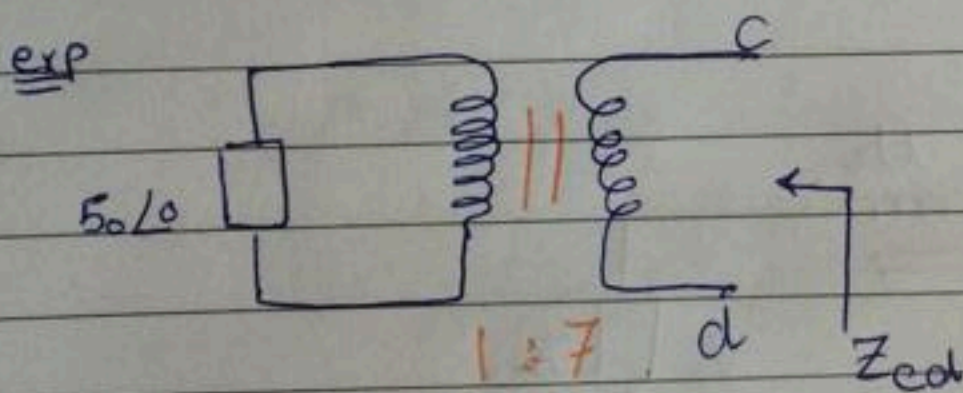


$$\frac{N_1}{N_2} = \frac{V_1}{V_2} = \frac{I_2}{I_1}$$

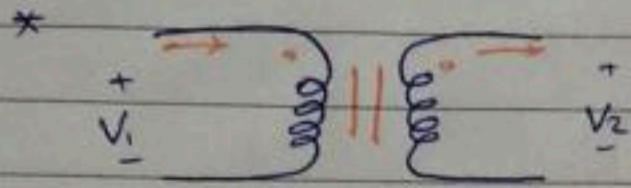
$$\Rightarrow N_1 I_1 = N_2 I_2$$

$$Z_1 = \frac{V_1}{I_1} = Z_2' = \frac{N_1 V_2 N_1}{N_2 N_2 I_2} = \left(\frac{N_1}{N_2}\right)^2 \cdot \frac{V_2}{I_2} = \left(\frac{N_1}{N_2}\right)^2 Z_2$$

Z_2' \Rightarrow الممانعة في الطرف الثاني
متصورة للطرف الاول

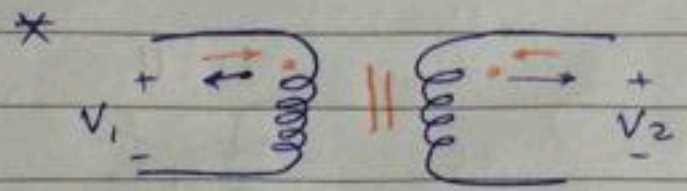


$$50 \Omega \left(\frac{7}{1}\right)^2 = Z_{cd}$$



$$\frac{V_1}{V_2} = \frac{N_1}{N_2}$$

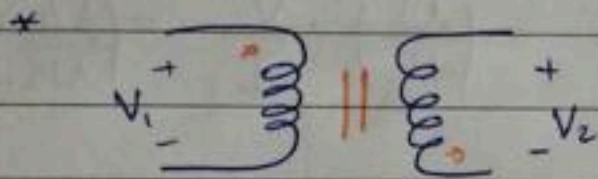
$$N_1 I_1 = N_2 I_2$$



$$\frac{V_1}{V_2} = \frac{N_1}{N_2}$$

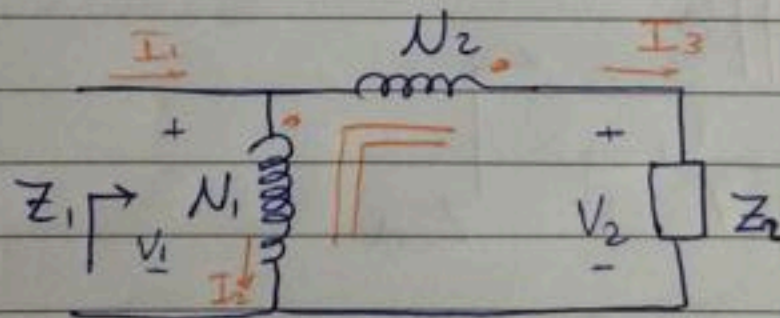
$$N_1 I_1 = -N_2 I_2$$

(كل السهين داخلين)
او خارجين -ve



$$\frac{V_1}{V_2} = -\frac{N_1}{N_2}$$

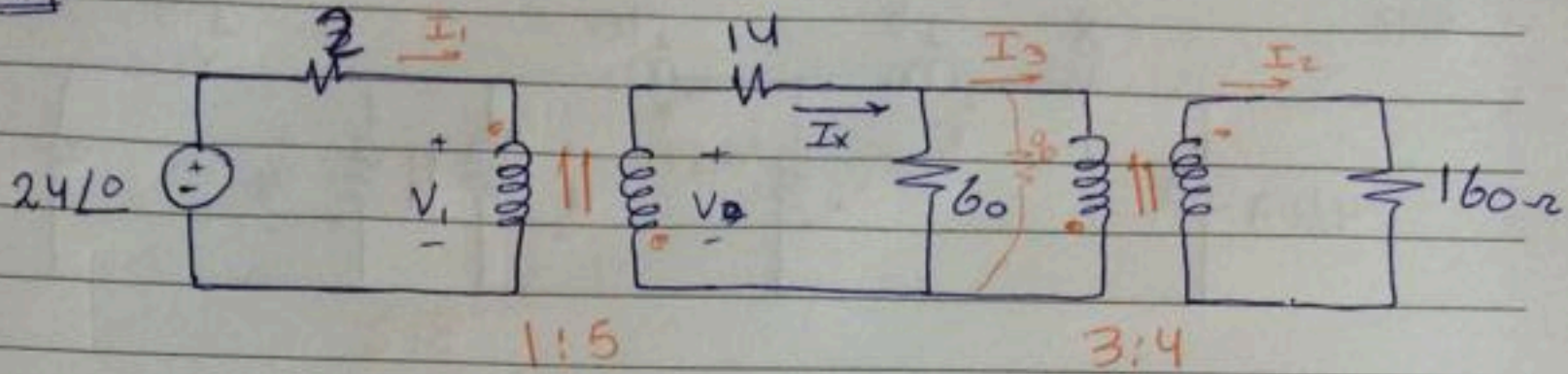
H.W



Prove that $Z_1 = Z_2 \left(\frac{N_2}{N_2 + N_1} \right)^2$

(9)

exp



I_1 , I_2 , I_3 and V_0

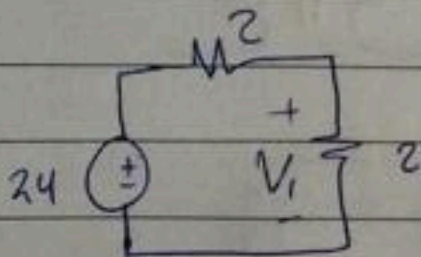
Sol: $Z_1 = \frac{V_1}{I_1} \rightarrow I_1 = \frac{V_1}{Z_1} = \frac{24}{4} = \boxed{6} \text{ A}$

$$\left\{ \left[\left(160 \left(\frac{3}{4} \right)^2 \right) \parallel 60 \right] + 14 \right\} \times \left(\frac{1}{5} \right)^2 + 3 = \boxed{4}$$

$$I_x = -6 \left(\frac{1}{5} \right)$$

$$I_3 = \left(\frac{-6}{5} \right) \frac{60}{150} = \left(\frac{-6}{5} \right) \times 0.4 = -0.48 \text{ A}$$

$$I_2 = - \left(\frac{-6}{5} \right) (0.4) + \frac{3}{4} = 0.36 \text{ A}$$

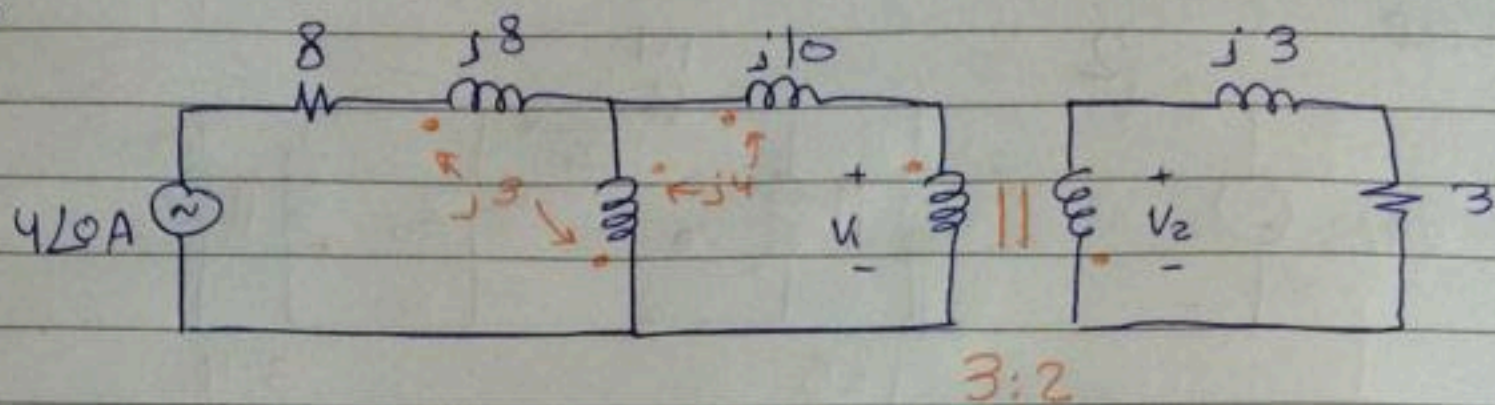


$$V_1 = 24 \times \frac{2}{4} = \boxed{12} \text{ V}$$

$$V_0 = \boxed{-60} \text{ V}$$

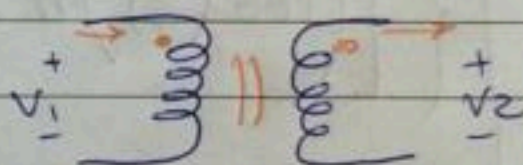
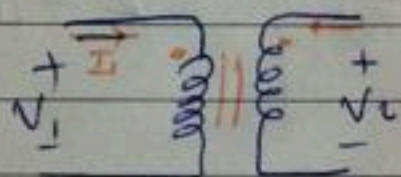
10

exp



Find V_2 and I_2

$$Z' = (3 + j3) \left(\frac{3}{2}\right)^2$$

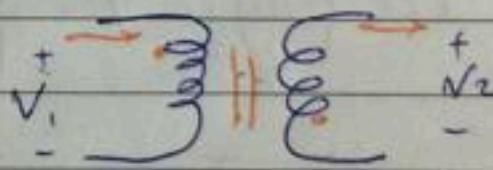
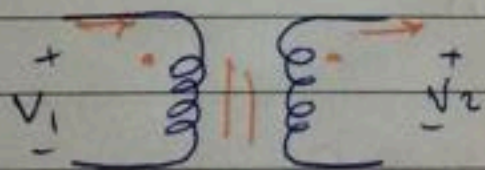


$$\frac{V_1}{N_1} = \frac{V_2}{N_2}$$

$$\frac{V_1}{N_1} = \frac{V_2}{N_2}$$

$$N_1 I_1 = -N_2 I_2$$

$$N_1 I_1 = N_2 I_2$$



$$\frac{V_1}{N_1} = \frac{V_2}{N_2}$$

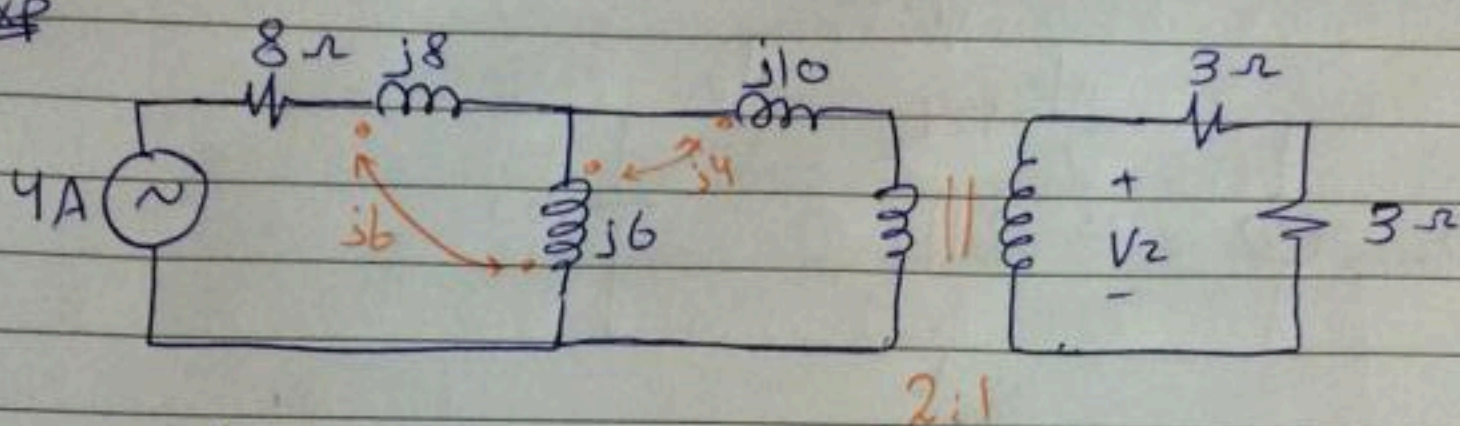
$$\frac{V_1}{N_1} = -\frac{V_2}{N_2}$$

$$N_1 I_1 = N_2 I_2$$

$$I_1 N_1 = -I_2 N_2$$

(11)

Exp



Find V_2 ?

$$G\left(\frac{2}{1}\right)^2 = \boxed{24}$$

$$V_{th} = [(-j4) + j6 - j3] = -j4$$

$$= 4 \angle -90$$

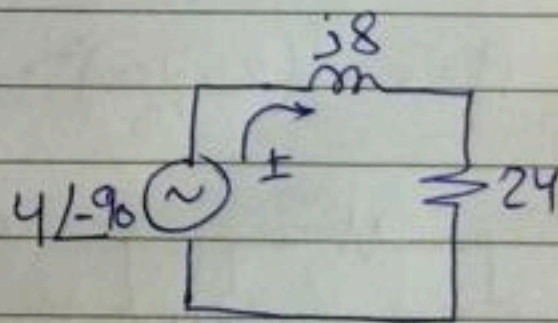
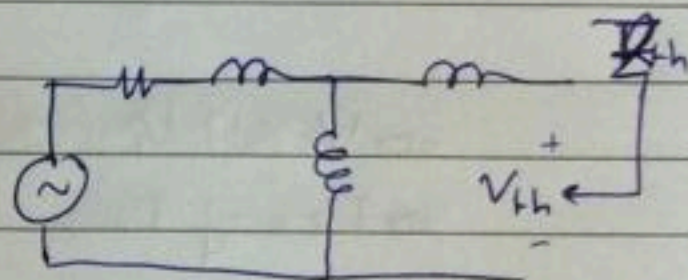
$$Z_{th} = j8$$

$$V_{j10} = -j4(4)$$

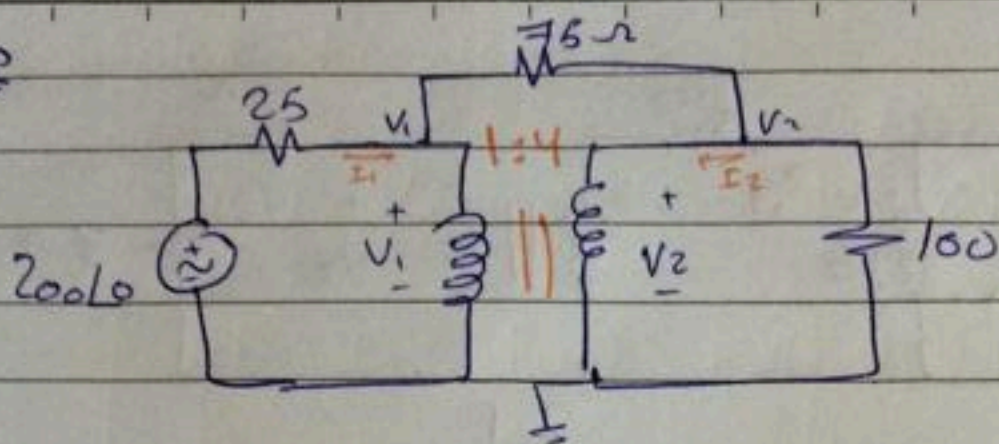
$$V_{j6} = (j6(4)) - j3(4)$$

$$\left(\frac{(4 \angle -90) * 2}{24 + j8} \right) * 3$$

$$I = \frac{4 \angle -90}{24 + j8}$$



exp



$$\frac{V_1 - 200}{25} + \frac{V_1 - V_2}{75} + I_1 = 0$$

$$\Rightarrow V_2 = 4 V_1$$

$$\Rightarrow I_2 = -\frac{1}{4} I_1$$

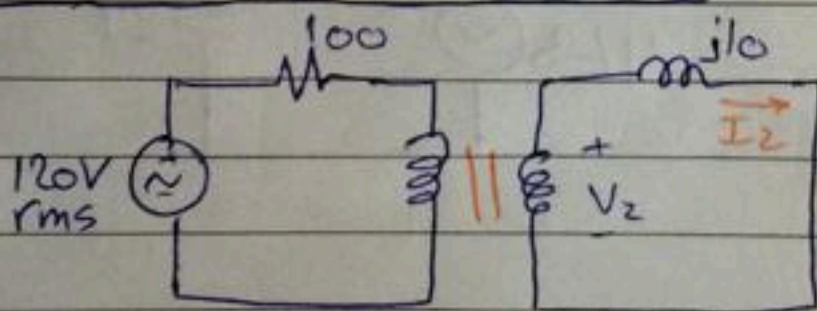
$$\frac{V_2 - V_1}{75} + \frac{V_2}{100} + I_2 = 0$$

$$V_1 = 25$$

$$V_2 = 100$$

$$P_{25\Omega} = \left| \frac{(25 - 100)^2}{75} \right| \times 75$$

exp



if $|I_2| = 1 \text{ A}$
Find n

$$120 = 100 \left(\frac{1}{n} \right) + (j10 I_2) n$$

$$V_2 = j10 I_2$$

or

$$Z_{in} = 100 + j10 \left(\frac{n}{1} \right)^2$$

$$\frac{120}{1/n} = 100 + j10 (n)^2$$

(13)

exp Linear Transformer $\omega = 5000 \text{ rad/sec}$

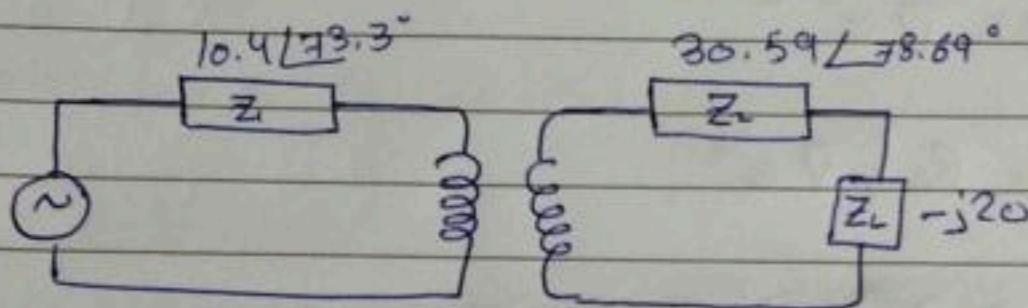
$$Z_1 = 10.4 \angle 73.3^\circ \text{ (Primary imp.)}$$

$$Z_2 = 30.59 \angle 78.69^\circ \text{ (secondary imp.)}$$

(reflected impedance magnitude = 13.08Ω)

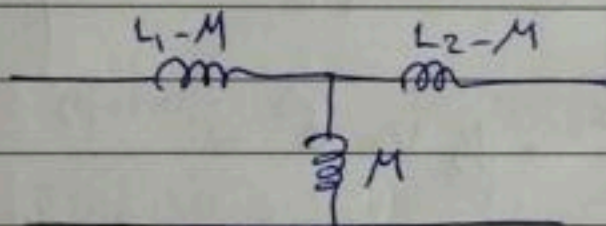
$$Z_L = -j20 \text{ (load imp.)}$$

Find T_{eq} ?



$$\Rightarrow \left| \frac{\omega^2 M^2}{Z_2} \right| = 13.08 \Omega$$

$$Z_{in} = Z_{11} + Z_r$$



$$Z_{11} = R + j\omega L_1$$

Complex Frequency:-

$$v(t) = V_m e^{\sigma t} \cos(\omega t - \varphi)$$

$$v(t) = V_m \cos(\omega t) = \operatorname{Re}\{V(t)\} = \operatorname{Re}\{\bar{V} e^{j\omega t}\}$$

$$V(t) = \bar{V} e^{j\omega t}$$

$$v(t) = \operatorname{Re}\{\bar{V} e^{j\omega t}\}$$

~~$$e^{jx} = \cos x + j \sin x$$~~

$$\Rightarrow \sin x = \frac{e^{jx} - e^{-jx}}{j2}$$

$$v(t) = V_m e^{\sigma t} \cos(\omega t - \varphi)$$

$$= \frac{1}{2} V_m e^{\sigma t} \left[e^{j(\omega t - \varphi)} + e^{-j(\omega t - \varphi)} \right]$$

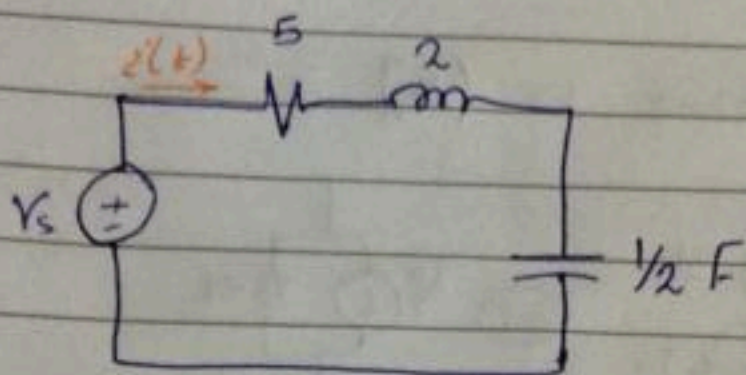
$$= \frac{1}{2} V_m e^{\sigma t} e^{j0} + \frac{1}{2} V_m e^{\sigma t} e^{-j(\omega t - \varphi)}$$

$$= \frac{1}{2} V_m e^{j0} e^{(\sigma + j\omega)t} + \frac{1}{2} V_m e^{-j\varphi} e^{(\sigma - j\omega)t}$$

$$= \boxed{\bar{V} e^{s_1 t} + \bar{V}^* e^{s_1^* t}}$$

(15)

exp



$$v(t) = 100 e^{-3t} \cos(4t - 50)$$

$$\bar{V} = 100 \angle -50$$

$$s = -3 + j4$$

$$v_s(t) = 5 i(t) + 2 \frac{d i(t)}{dt} + 2 \int_0^t i(t) dt$$

$$= 5 \bar{I} e^{st} + 5s \bar{I} e^{st} + \frac{2}{s} \bar{I} e^{st} = V e^{st}$$

$$\bar{I} = \frac{100 \angle -50}{5 + 2s + \frac{2}{s}} \Big|_{s = -3 + j4} = \boxed{12.9 \angle +149}$$

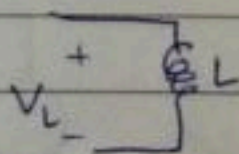
$$i(t) = \bar{I} \cdot e^{st}$$

$$= I e^{-st} \cos(4t + \phi_i)$$

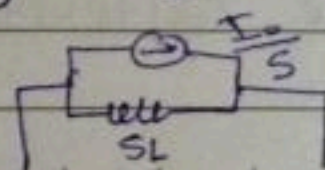
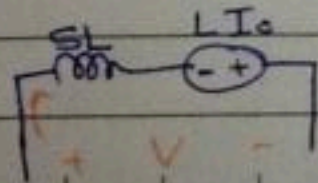
$$i(t) = 12.9 e^{-st} \cos(4t - 149)$$

In Laplace

$$\textcircled{1} v_L = L \frac{di}{dt} \Big| \mathcal{L}$$

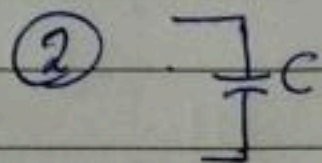
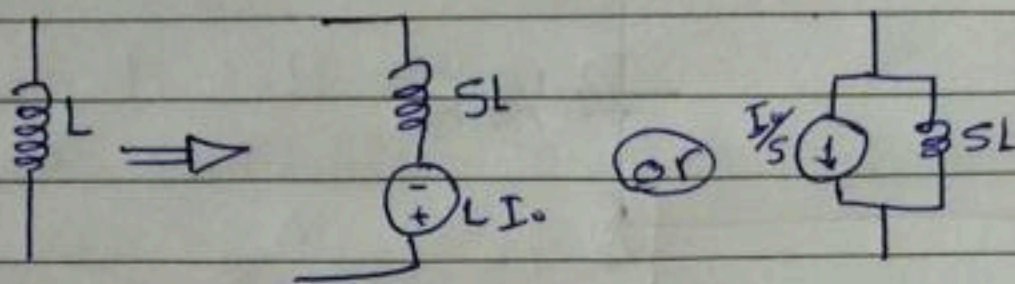


$$V(s) = \mathcal{L}\{v(t)\} = \mathcal{L}\left\{L \frac{di}{dt}\right\} = L [s I(s) - I_0] = sL I(s) - L I_0$$



16

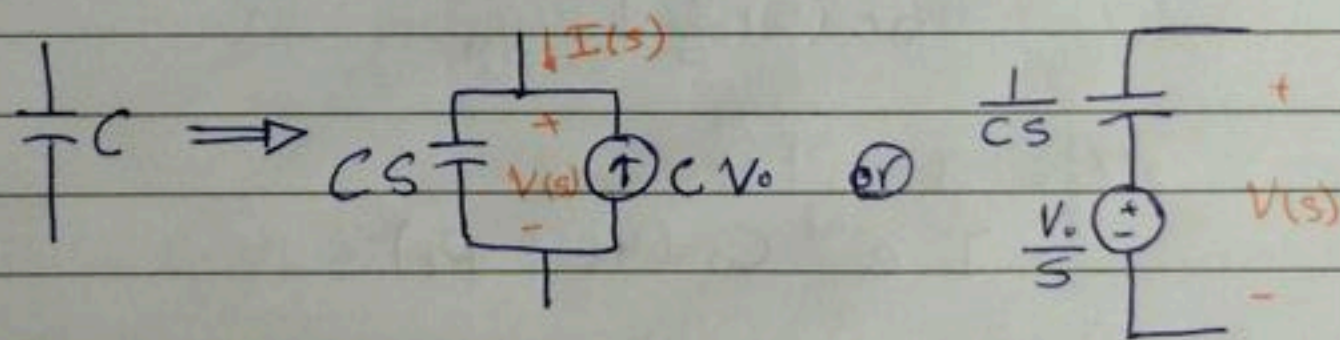
⇒ Convert (L) from time to (S) domain



$$z = C \frac{dv}{dt} \quad | \quad \mathcal{L}$$

$$\mathcal{L}\{i(t)\} = I(s) = \mathcal{L}\left\{C \frac{dv}{dt}\right\} = C [sV(s) - V_0]$$

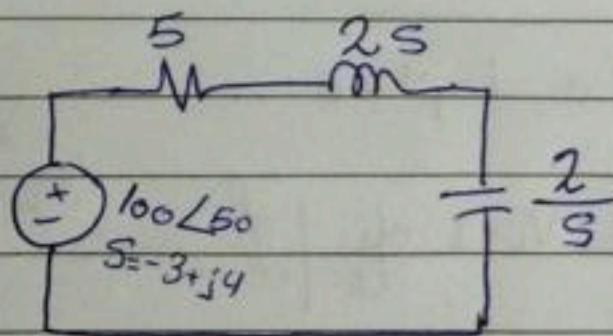
⇒ Convert to S domain



* Back to the example :-

Convert to S domain

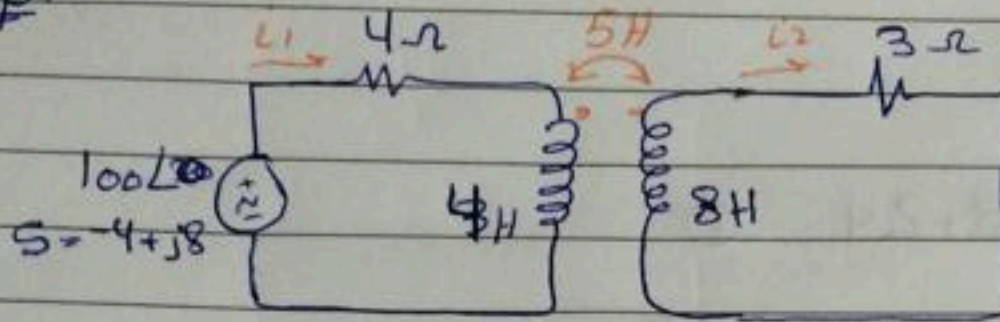
$$\bar{I} = \frac{100 \angle -50}{5 + 2s + \frac{2}{s}} = 12.9 \angle -149$$



$$i'(t) = 12.9 e^{-3t} \cos(4t - 149)$$

17

exp:-



$$i_1(0) = i_2(0) = 4 \text{ A}$$

$$100 \angle 0 = (4 + 3s) I_1(s) - 5s I_2 \quad \text{--- ①}$$

$$0 = 5s I_1 + (3 + 8s) I_2 \quad \text{--- ②}$$

$$I_2 = \frac{\begin{pmatrix} 4+3s & 100 \\ -5s & 0 \end{pmatrix}}{\begin{pmatrix} 4+3s & -5s \\ -5s & 3+8s \end{pmatrix}} = \frac{500s}{7s^2 + 44s + 12}$$
$$= \frac{500s}{7(s + \frac{2}{7})(s + 6)}$$

$$\bar{I}_2(s) = \frac{500s}{7(s + \frac{2}{7})(s + 6)} \Big|_{s=4+j8} = 8.8 \angle -74^\circ$$

• Poles: القيم التي تجعل المقام صفر
• Zeros: البسط صفر " " "

$$i_2(t) = A e^{-\frac{2}{7}t} + B e^{-6t} \quad (\text{natural})$$

$$i_2(t) = i_{2n}(t) + i_p(t)$$

$$i_2(t) = A e^{-\frac{2}{7}t} + B e^{-6t} + 8.8 e^{-4t} \cos(8t - 74^\circ)$$

To Find A, B

$$i_2(0) = 4 = A + B + 2 \cdot 4 \dots \textcircled{1}$$

~~$$\frac{d i_2(t)}{dt} = \frac{2A}{7} - 6B - 18 \cdot 18$$~~

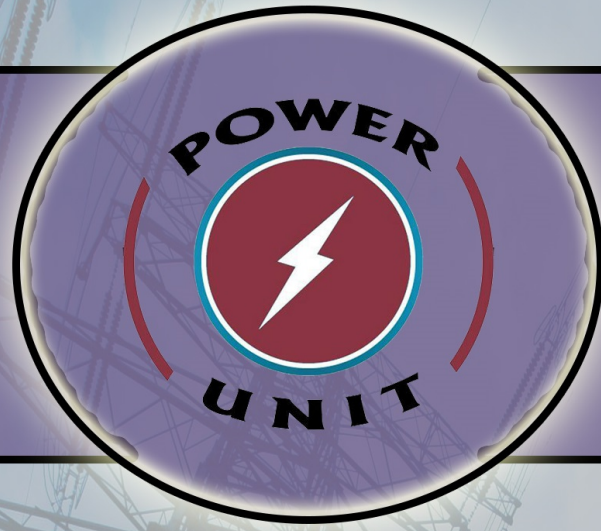
$$\frac{d i_2(t)}{dt} = -\frac{2A}{7} - 6B - 18 \cdot 18$$

$$\frac{d i_2(t)}{dt} = ??$$

$$\begin{aligned} \Rightarrow 100 \angle 30 &= (4 + 4s) I_1(s) - 5s I_2 \\ &= 4 i_1(0) + 4 \frac{d i_1(0)}{dt} - 5 \frac{d i_2(0)}{dt} \end{aligned}$$

$$= 4 i_1(0) + 4 \frac{d i_1(0)}{dt} - 5 \frac{d i_2(0)}{dt}$$

$$\textcircled{2} = -5 \frac{d i_1(0)}{dt} + 3 i_2(0) + 8 \frac{d i_2(0)}{dt}$$



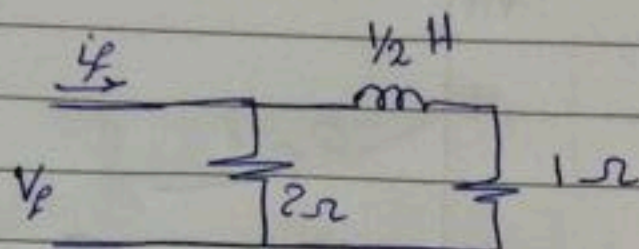
Circuits II Notebook

Dr. Nabeel Tawalbeh

By . Yazan Abawi

بِأفكارنا نبدع

6th week

exp

① Find the forced current response at input if the voltage $V_p(t) = 2 e^{-t}$ for all t ?

$$H(s) = \frac{\text{Output}}{\text{input}} = \frac{I(s)}{V(s)} = Y(s)$$

$$I_p(s) = H(s) \cdot V(s) = Y(s) \cdot V(s)$$

$$i_p(t) = I_p \cdot e^{-t}$$

$$I_p(s) = V \cdot Y(s) = 2 \cdot \frac{2 + (1 + \frac{s}{2})}{(1 + \frac{1}{2}s) \cdot 2} \Big|_{s=-1} = 2 \frac{2 + 1 - \frac{1}{2}}{(1 - \frac{1}{2}) \cdot 2} = 5$$

$$i_p(t) = 5 e^{-t}$$

② Find the forced voltage response at input if the current $i_p(t) = 1.5 \cos(2t + 42^\circ)$ for all t ?

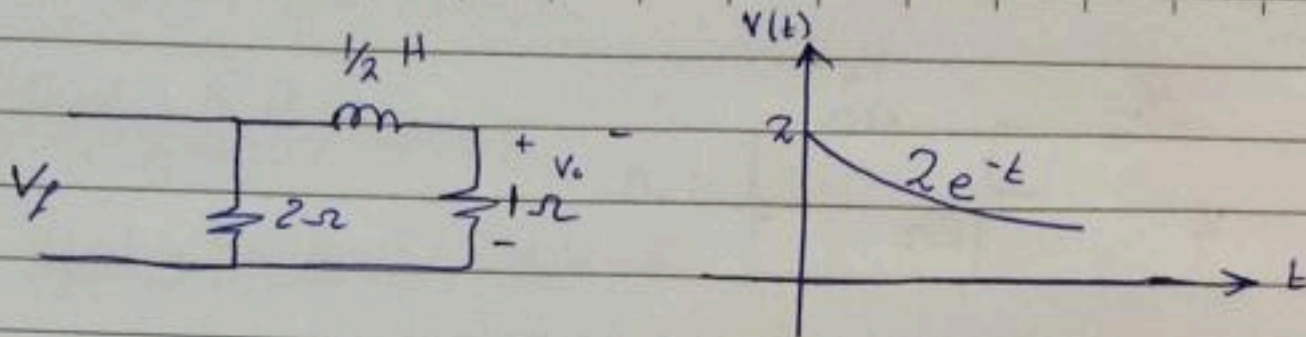
$$H(s) = \frac{V(s)}{I(s)} = H(s) = Z(s)$$

$$V_p(s) = H(s) I(s) \Big|_{s=j2} = 1.5 \frac{(2 + 5) \cdot 2}{(6 + s)} \Big|_{s=j2} = 1.34 \angle 68.6^\circ$$

$$V_p(t) = 1.34 \cos(2t + 68.6^\circ)$$

①

exp



Find $i(t)$?

$$i(t) = i_n(t) + i_p(t) \\ = A e^{p_1 t} + B e^{p_2 t} + \dots + I_f e^{st}$$

$$Z(s) = \frac{2(s+2)}{s+6}$$

$$I(s) = \frac{V_s}{Z(s)} = \frac{2(s+6)}{2(s+2)} \Big|_{s=-1} = 5$$

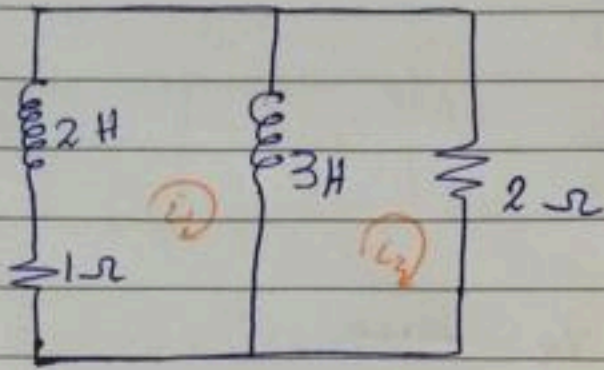
$$I(s) = \frac{2(s+6)}{2(s+2)} \Rightarrow i_n(s) = A e^{-2t}$$

$$i(t) = 3e^{-t} + A e^{-2t}$$

$$2(0) = 5 + A = 1 \Rightarrow \boxed{A = -4}$$

$$\boxed{i(t) = 3e^{-t} - 4e^{-2t}}$$

exp



Find $i_1(t)$, $i_2(t)$ if $i_1(0) = i_2(0) = 11 \text{ A}$

$$V_s = (1 + 5s)I_1 - 3sI_2 \quad \text{--- ①}$$

$$0 = -3sI_1 + (2 + 3s)I_2 \quad \text{--- ②}$$

$$I_1 = \frac{\begin{pmatrix} V_s & -3s \\ 0 & 2+3s \end{pmatrix}}{\begin{pmatrix} 1+5s & -3s \\ -3s & 2+3s \end{pmatrix}} = \frac{(2+3s)V_s}{6\left(s^2 + \frac{13}{6}s + \frac{2}{6}\right)}$$

$$I(s) = \frac{3\left(s + \frac{2}{3}\right)}{6\left(s+2\right)\left(s + \frac{1}{6}\right)}$$

$$\left[\begin{array}{l} \text{Zeros: } -2/3 \\ \text{Poles: } -2, -1/6 \end{array} \right.$$

$$i_1(t) = A e^{-1/6t} + B e^{-2t}$$

$$i_1(0) = 11 = A + B \quad \text{--- ①}$$

$$0 = 1 i_1(0) + 5 \frac{d i_1(0)}{dt} - 3 \frac{d i_2(0)}{dt} \quad \text{--- ②}$$

$$0 = -3 \frac{d i_1(0)}{dt} + 2 i_2(0) + 3 \frac{d i_2(0)}{dt} \quad \text{--- ③}$$

$$5x_1 - 3x_2 = -11$$

$$-3x_1 + 3x_2 = -22$$

$$x_1 = \frac{-16.5}{-16.5}, \quad x_2 = \frac{-23.8}{-23.8}$$

$$\frac{di_1(0)}{dt} = -16.5 = -\frac{1}{6}A - 2B \quad \text{--- [2]}$$

$$A = 3$$

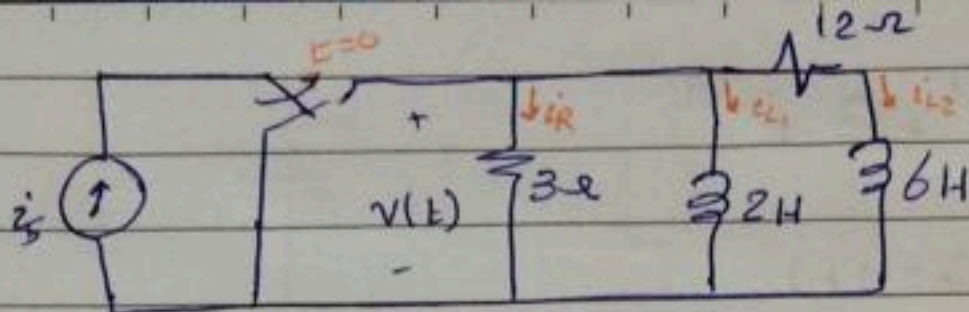
$$B = 8$$

$$i_1(t) = 3e^{-1/6t} + 8e^{-2t}$$

$$\frac{di_2(0)}{dt} = -23.8 = -\frac{1}{6}A - 2B \quad \text{--- [3]}$$

$$i_2(t) = -0.98e^{-1/6t} + 11.98e^{-2t}$$

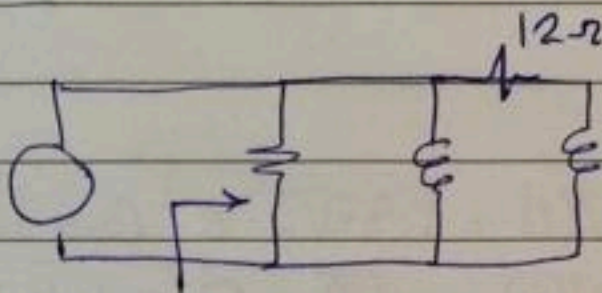
exp



$$i_s(t) = e^{-t} \cos 2t$$

Find $v(t)$

$$V = Z \cdot I$$



$$V(s) = \frac{1}{Y} \cdot I = \frac{I(s)}{\frac{1}{3} + \frac{1}{2s} + \frac{1}{6s+12}}$$

$$\frac{V(s)}{I(s)} = H(s)$$

$$V(s) = H(s) I(s)$$

~~$$H(s) = \frac{3/6 s^2}{12(s^2 + \frac{18}{12}s + \frac{3}{12})}$$~~

$$V(s) = \frac{6s(12+6s)}{12s^2+24s+36+18s+6s} = \frac{36(s^2 + \frac{72}{36}s)}{12(s^2+4s+3)}$$

$$= \frac{3s(s+2)}{(s+1)(s+3)}$$

Zeros: 0, -2

Poles: -1, -3

(5)

$$v_n(t) = A e^{-t} + B e^{-3t}$$

$$v_f(t) = \sqrt{V_f} e^{-t} \cos(2t)$$

$$v_f(s) = I(s) + H(s) \Big|_{s=-1+j2} = \frac{3s(s+2)}{(s+1)(s+3)} \Big|_{s=-1+j2} = 1.875\sqrt{2} / 45^\circ$$

$$\Rightarrow v(t) = 1.875\sqrt{2} e^{-t} \cos(2t + 45^\circ) + A e^{-t} + B e^{-3t}$$

$$v(0) = \boxed{A + B + 1.875 = 3} \quad \dots \text{[1]}$$

$$V_R = R \cdot i_s$$

$$i(0) = 1 \Rightarrow V_R = v(t) = 3V$$

$$v(t) = 3i_R = 3i_s - 3i_{L_1} - 3i_{L_2}$$

$$\frac{dv(t)}{dt} = 3 \frac{di_s}{dt} - 3 \frac{di_{L_1}}{dt} - 3 \frac{di_{L_2}}{dt}$$

$$\Rightarrow \frac{dv(t)}{dt} \Big|_{t=0} = 1.875\sqrt{2} (-e^{-t} \cos(2t+45^\circ) - 2e^{-t} \sin(2t+45^\circ)) \Big|_{t=0} = -A - 3B$$

$$-A - 3B - 3 \times 1.875 = \frac{dv(0)}{dt}$$

$$\Rightarrow \frac{dv(t)}{dt} = -3 - 3 \times \frac{3}{2} - 3 \times \frac{1}{2} = \boxed{-9}$$

$$\times 3 = L_1 \frac{di_{L_1}}{dt}$$

$$\boxed{A + 3B = 3.375} \quad \dots \text{[2]}$$

$$\frac{di_{L_1}}{dt} = \frac{3}{2}$$

$$A = 0$$

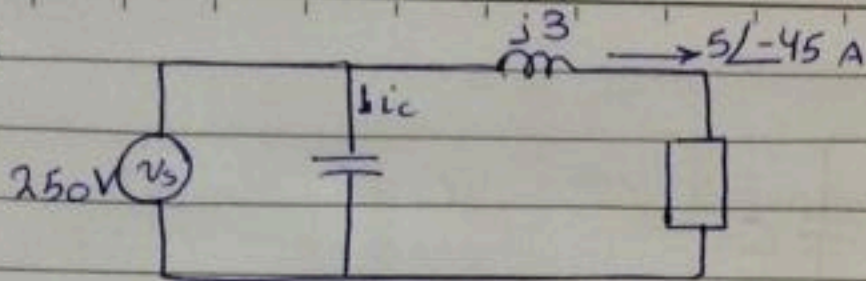
$$B = 1.125$$

$$\times 3 = L_2 \frac{di_{L_2}}{dt}$$

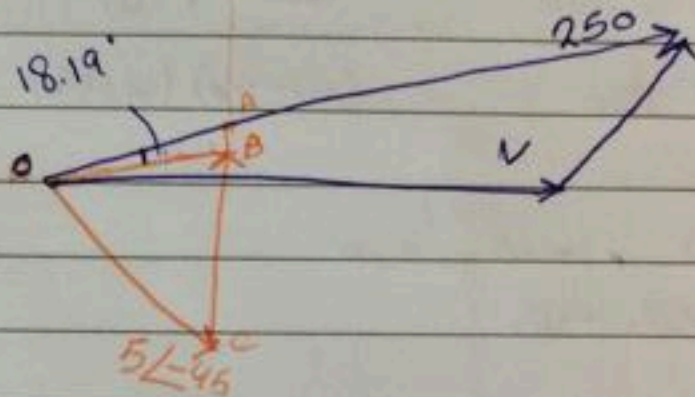
(b)

Quiz

[1]



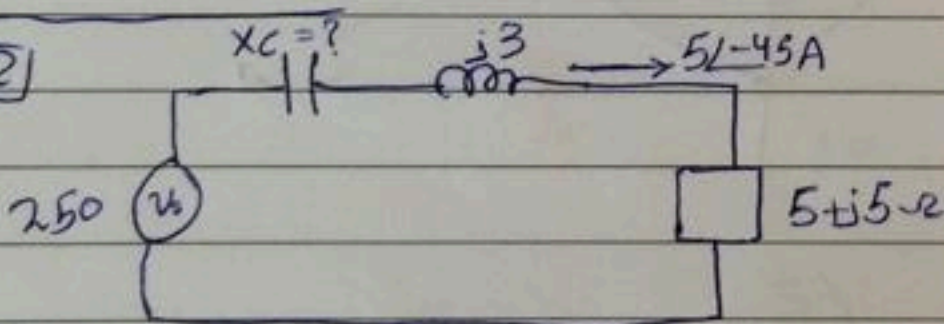
Find the value of I_c that improve the PF to 0.95 lagging using the phasor diagram.



$$BC = I_c = 5 \text{ A} (\tan \phi_1 - \tan \phi_2)$$

$$\text{Req} \{ I \} (\tan \phi_1 - \tan \phi_2)$$

[2]

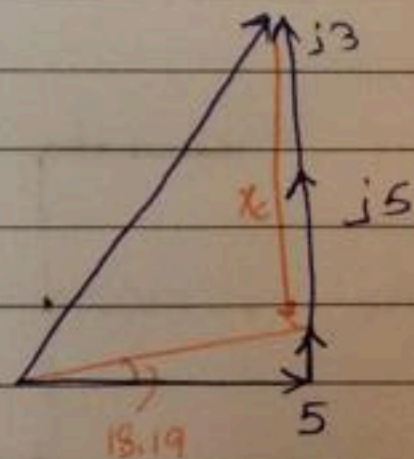


$$\text{PF} = 0.95$$

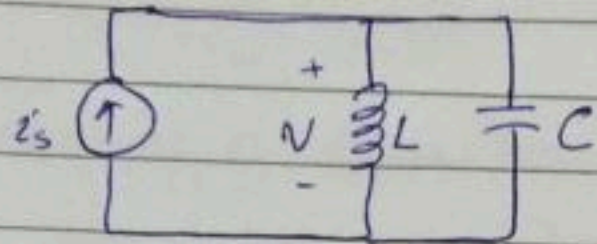
$$X_c = ?$$

$$X_c = j(8 - 1.6) \\ = -j6.4$$

$$Z_{in} = 5 + j1.6 \Rightarrow \theta = \tan^{-1} \left(\frac{1.6}{5} \right)$$



exp

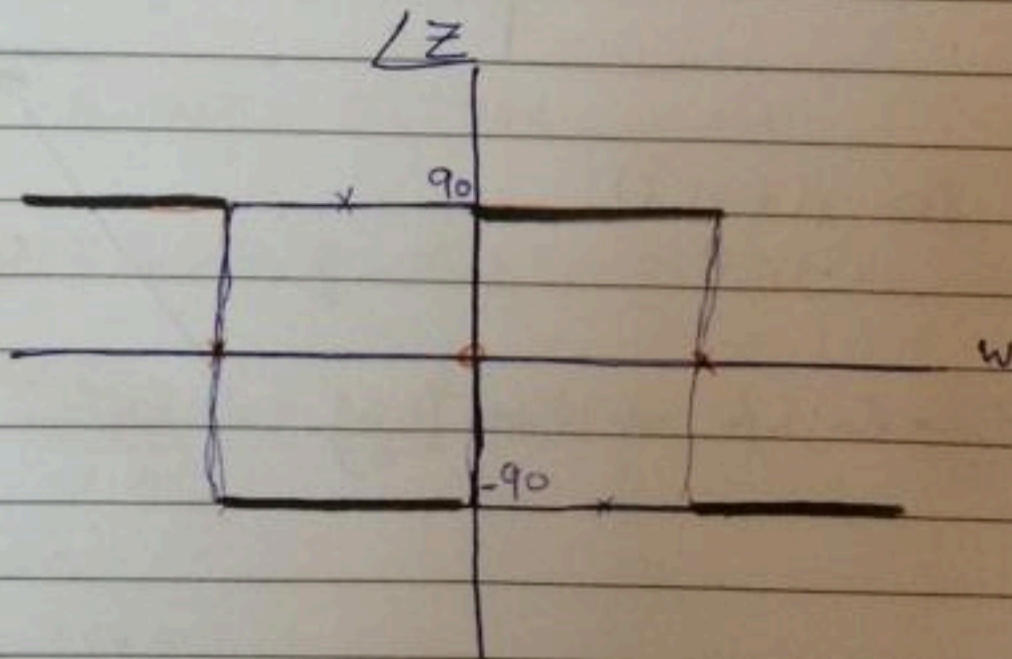
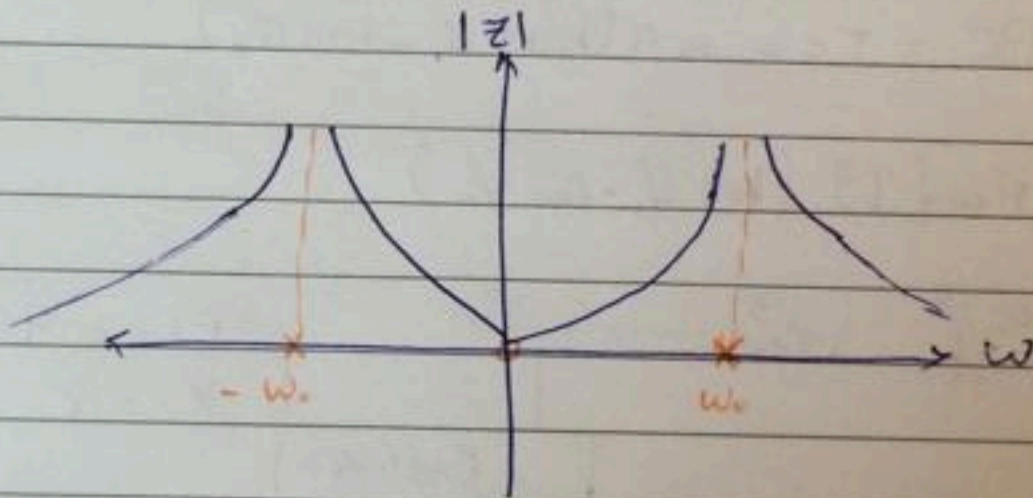


$|Z|$ v.s. ω

$$Z = \frac{\frac{1}{j\omega C} \times j\omega L}{j\omega L + \frac{1}{j\omega C}} = \frac{\frac{1}{C} \times \omega}{(\omega - \omega_0)(\omega + \omega_0)}$$

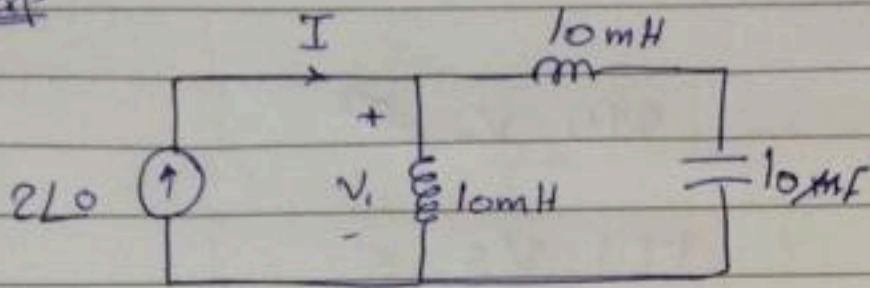
Zeros: $0, -\infty, +\infty$
Poles: $-\omega_0, +\omega_0$

$$V_0 = \frac{1}{\sqrt{LC}}$$



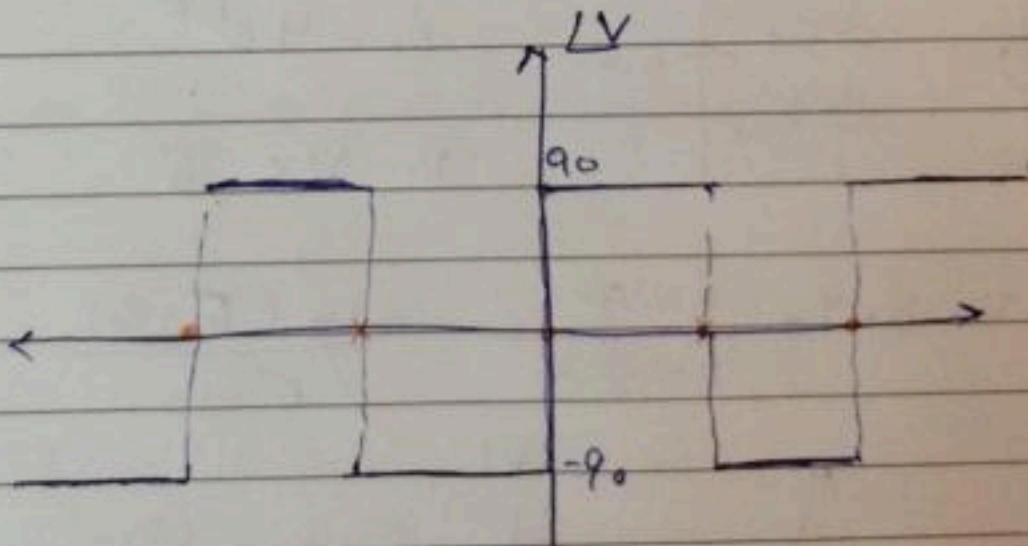
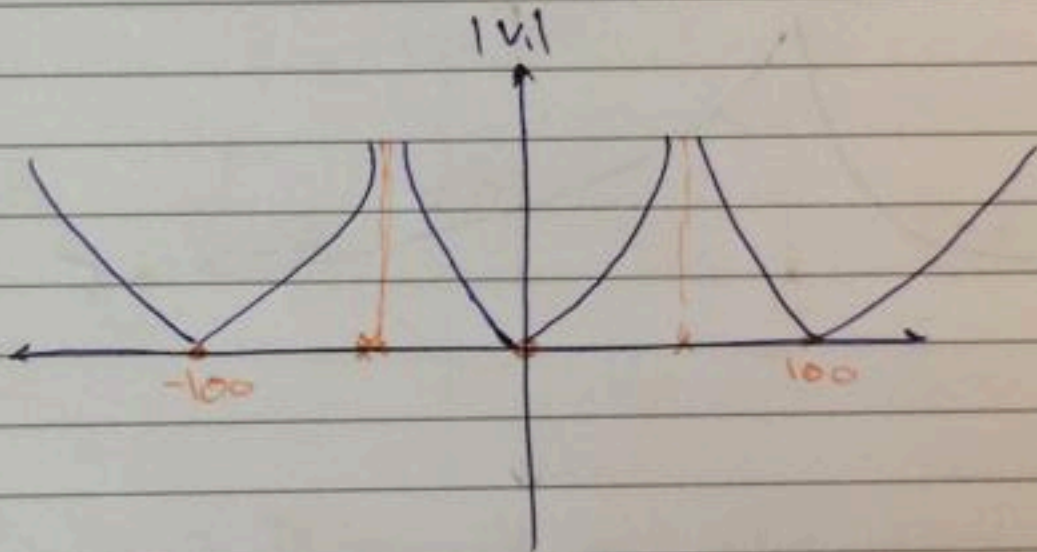
8

exp



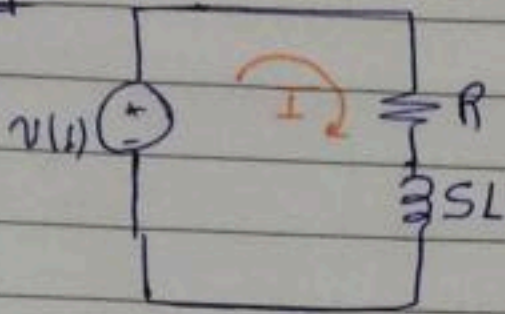
$$V_1 = I \cdot Z = I \left(j\omega 10\text{mH} + \frac{1}{j\omega 10\mu\text{F}} \parallel j\omega 10\text{mH} \right)$$

$$= \frac{j2\omega(\omega^2 - 1000)}{200(\omega^2 - 5000)}$$



(a)

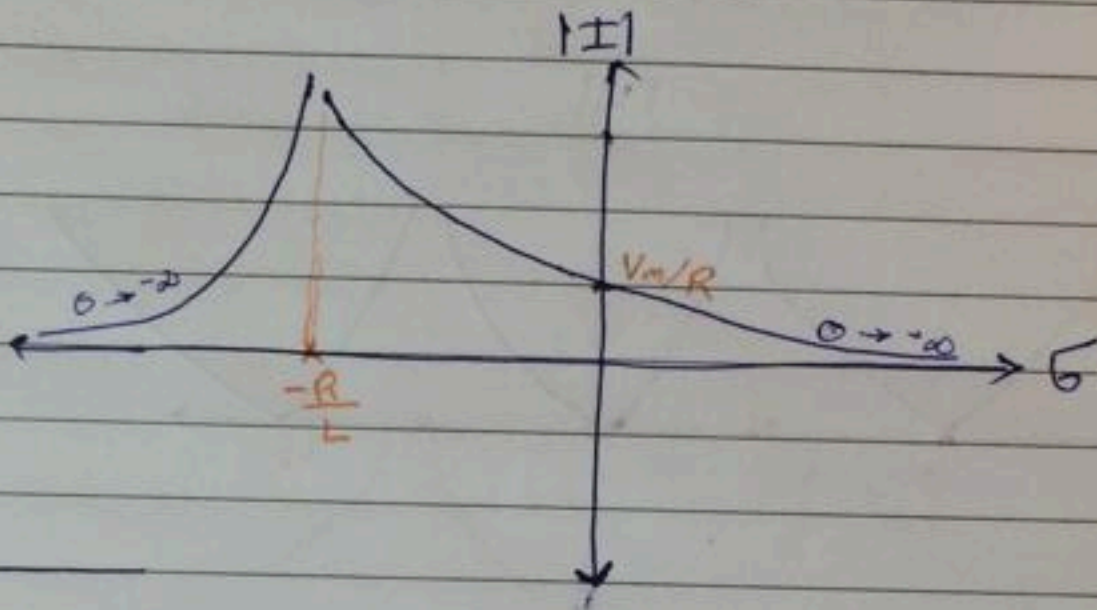
exp:



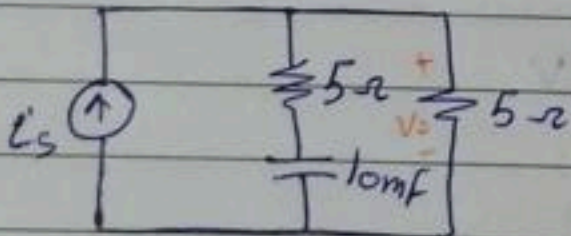
$$v(t) = V_m e^{\sigma t}$$

$|I|$ v.s. σ

$$I = \frac{V_m}{R + \sigma L} = \frac{V_m/L}{\left(\frac{R}{L} + \sigma\right)}$$



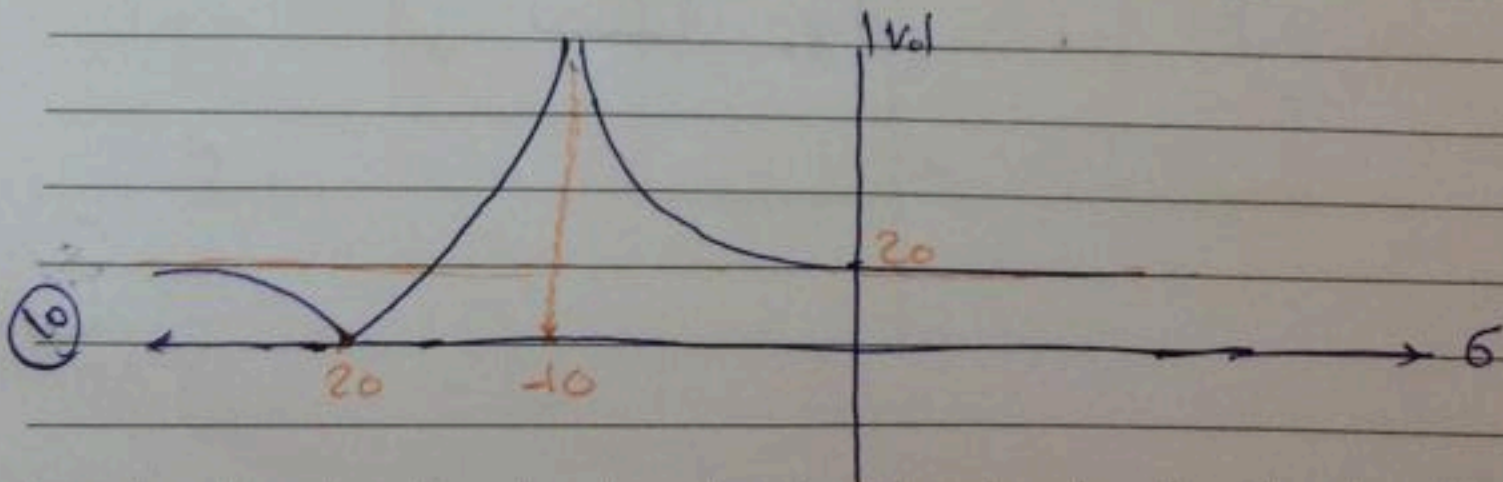
exp:



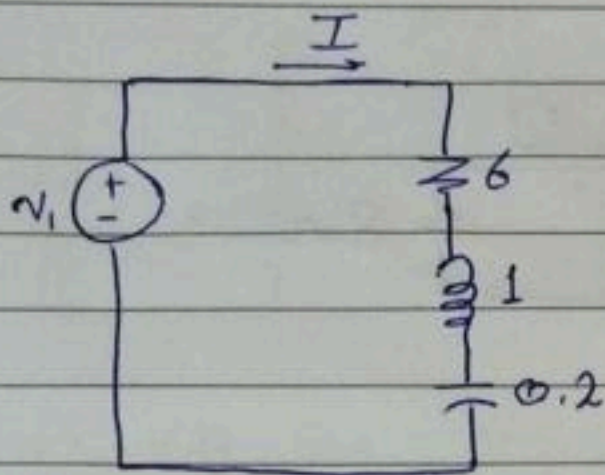
$$I_s = 4 e^{\sigma t}$$

v_0 v.s. σ

$$v_0 = 5 \times 4 = \frac{5 + \frac{100}{\sigma}}{5 + \frac{100}{\sigma} + 5} = \frac{10(\sigma + 20)}{\sigma + 10}$$



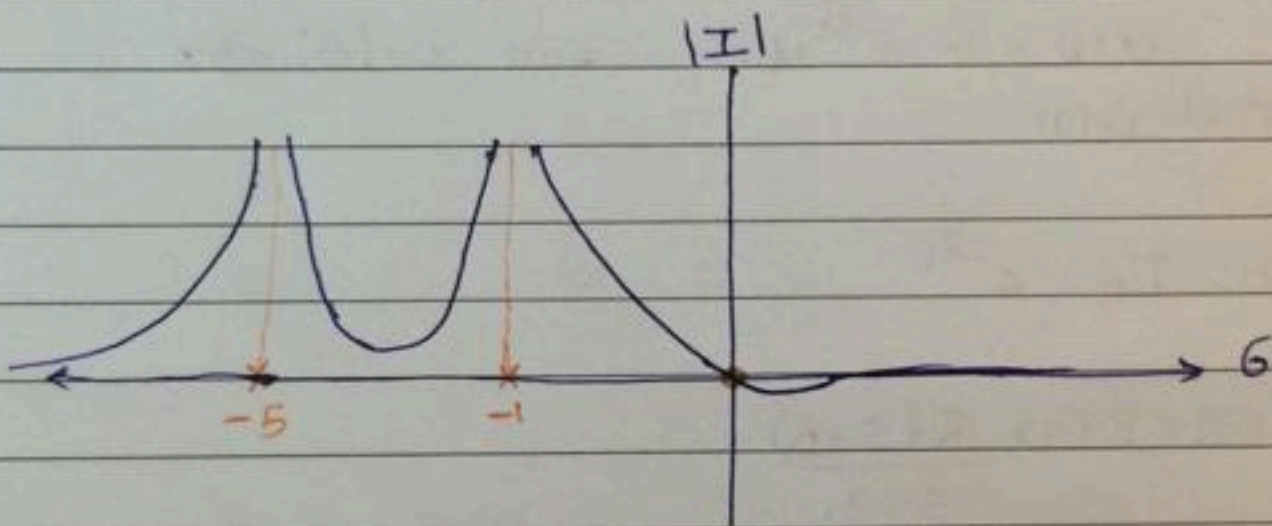
H.W



$$v_1 = 100 e^{5t}$$

$$|I| \text{ v.s. } \delta$$

$$I = \frac{100}{6 + s + \frac{s}{5}} = \frac{100 \delta}{(6 + 1)(6 + 5)}$$



Critical Points (minimum and maximum)

$$\Rightarrow \frac{\partial I}{\partial \delta} = \frac{(6+1)(6+5) - 6(2\delta+6)}{(\quad)^2}$$

$$-6\delta + 5 = 0$$

$$\boxed{\delta = \pm \sqrt{5}}$$

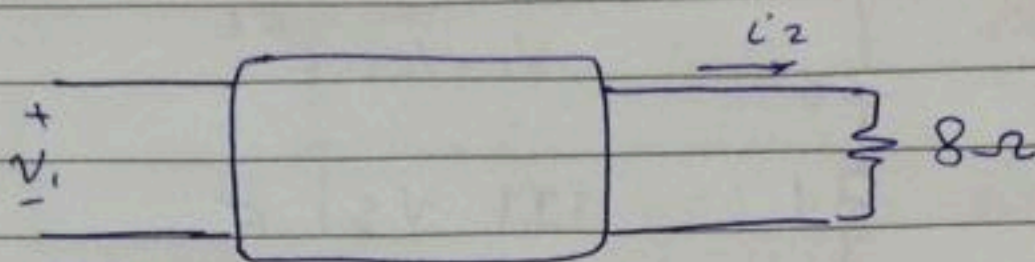
$$I(\sqrt{5}) = 9.55 \quad \text{maximum}$$

$$I(-\sqrt{5}) = 5.45 \quad \text{minimum}$$

$$I(\sqrt{3}) = 9.42$$

11

Quiz



$$\frac{I_2(s)}{V_1} = \frac{8(s+5)}{s+20}$$

a) if $v_1(t) = 10 e^{-8t}$ Find $i_{2p}(t)$

b) " $v_1(t) = 20$ " " "

c) " $v_1(t) = 10 e^{-8t} u(t)$ and $i_2(0) = 50$
Find $i_2(t)$

$$a) i_{2p} = I_{2p} e^{-8t}$$

$$I_2(s) = V_1(s) \frac{8(s+5)}{s+20}$$

$$I_{2p} = 10 \times \frac{8(s+5)}{s+20} \Big|_{s=-8} = \cancel{10} - 20$$

$$\boxed{i_{2p}(t) = -20 e^{-8t}}$$

$$b) \boxed{I_{2p} = 40}$$

$$c) i_2(t) = i_{2n}(t) + i_{2p}(t)$$

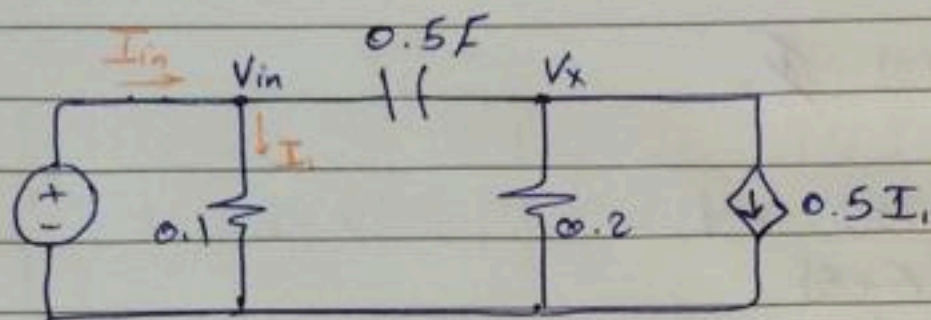
$$i_2(t) = A e^{-20t} - 20 e^{-8t}$$

$$i_2(0) = 50 = A - 20$$

$$\boxed{A = 70}$$

$$i_2(t) = (70 e^{-20t} - 20 e^{-8t}) u(t)$$

exp



a) write an expression of $Z(s) = \frac{V_{in}}{I_{in}}$

b) If V_{in} is replaced by a current source
 $I_{in} = 4e^{5t}$, Plot $|V_{in}(s)|$ v.s. s

(a) $Z(s) = \frac{V_{in}}{I_{in}} = H(s)$

$$\frac{V_{in} - V_x}{2/s} + \frac{V_{in}}{0.1} = I_{in} \quad \text{--- node ①}$$

$$\frac{V_x - V_{in}}{2/s} + \frac{V_x}{0.2} + 0.5 I_1 = 0 \quad \text{--- node ②}$$

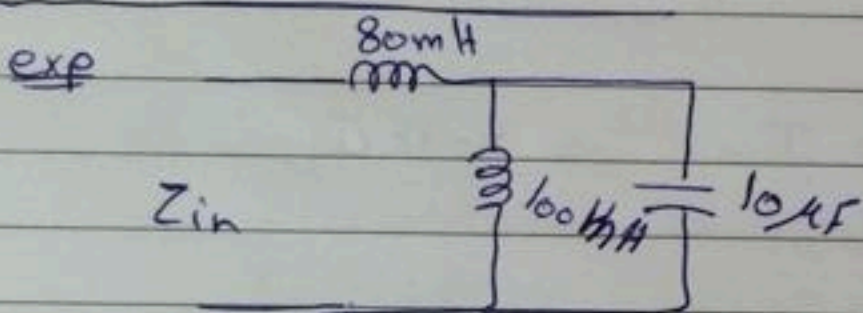
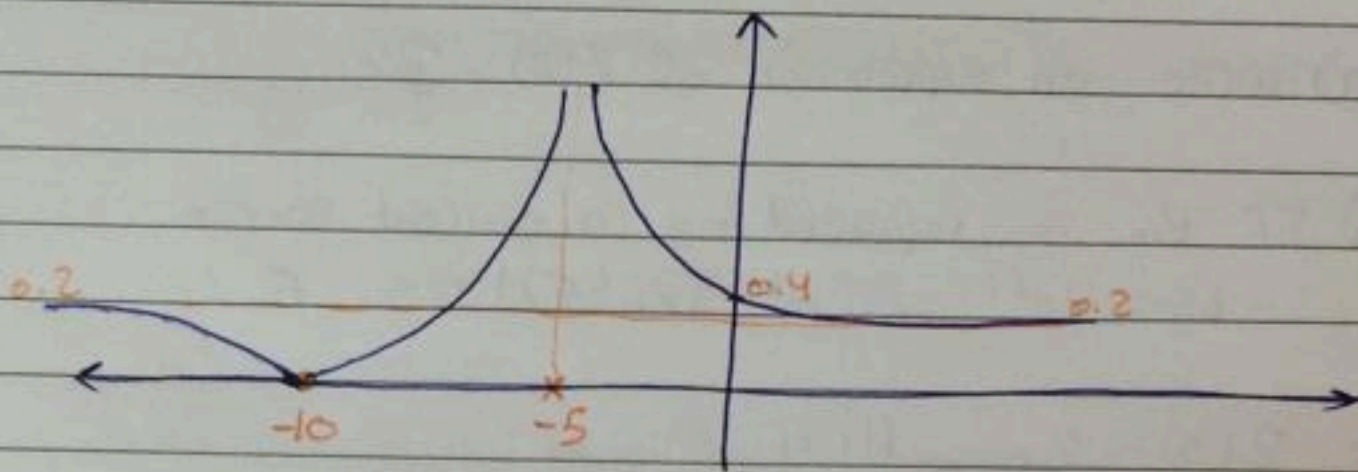
$$\left(\frac{s}{2} + 10\right) V_{in} - 2 V_x = I_{in}$$

$$\left(-\frac{s}{2} + 5\right) V_{in} + \left(\frac{s}{2} + 5\right) V_x = 0$$

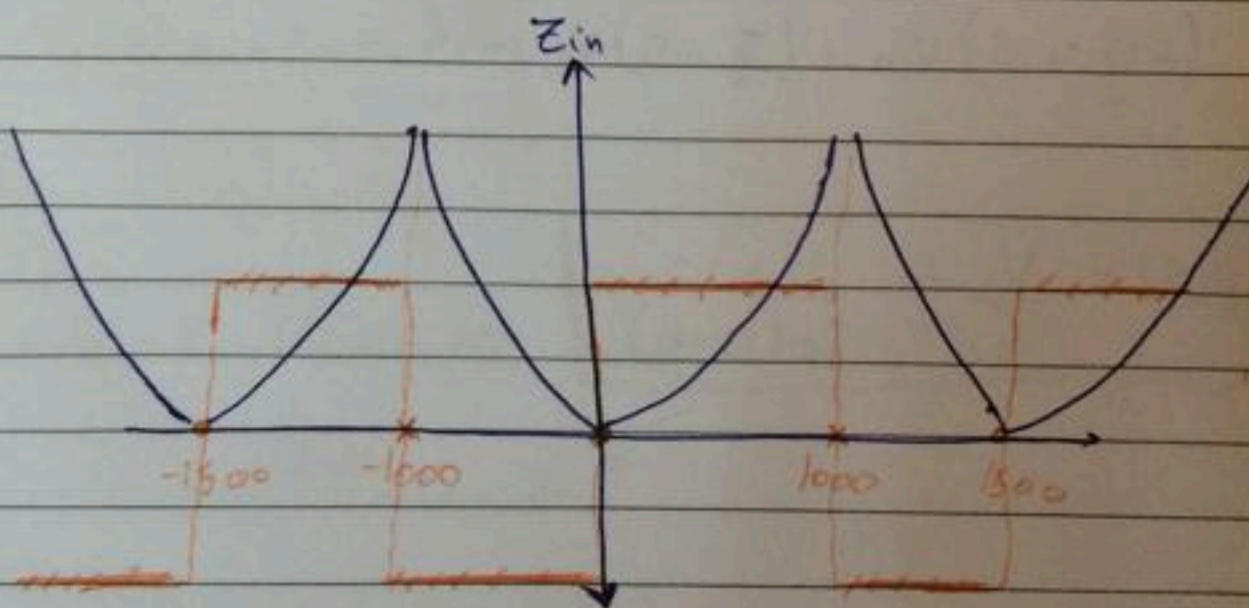
$$H(s) = \frac{V_x}{I_{in}} = \frac{s+10}{20(s+5)}$$

(b) $V_{in} = H(s) I_{in}$

$$= \frac{(s+10) \times 4}{20(s+5)}$$

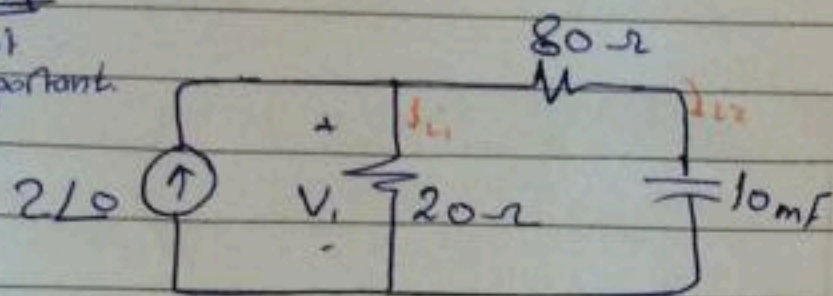


$$Z_{in} = \frac{j0.08\omega(\omega - 1500)(\omega + 500)}{(\omega - 1000)(\omega + 1000)}$$



(14)

exp
not
important



$$|V_1| \text{ vs. } \omega$$

$$\angle V_1 \text{ vs. } \omega$$

$$V_1 = 32 \sqrt{\frac{\omega^2 + 1.5625}{\omega^2 + 1}} = \frac{32(\omega - j1.25)}{\omega - j1}$$

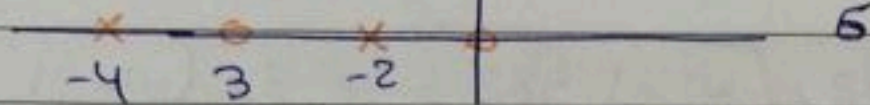
$$\angle V_1 = \tan^{-1} \frac{1.25}{\omega} - \tan^{-1} \frac{1}{\omega}$$

ω	V_1	$\angle V_1$	j_2
0	40	0	
1.25	39.6	-2.7	
2	33.8	-5.4	
∞	32	0	

exp

Find the $H(s)$

$$G(1) = 4$$

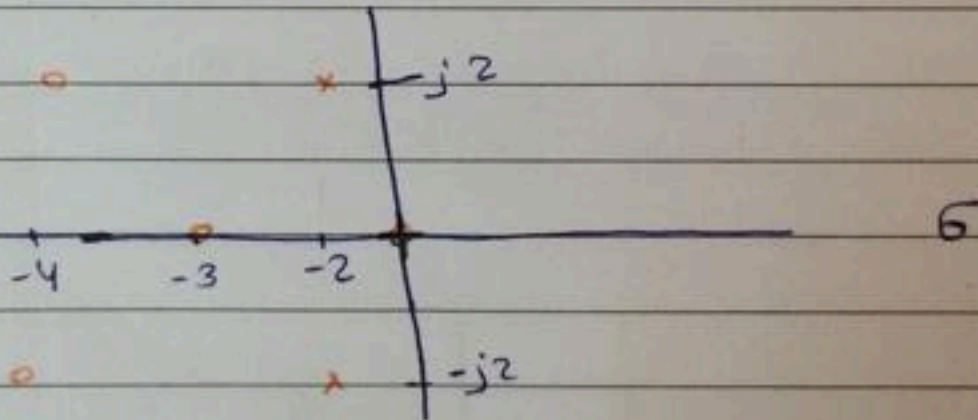


$$H(s) = K \frac{(s+z_1)(s+z_2)}{(s+p_1)(s+p_2)}$$

$$G(1) = 4 = K \frac{s(s+3)}{(s+2)(s+4)} \Big|_{s=1} = 15$$

$$G(s) = 15 \frac{s(s+3)}{(s+2)(s+4)}$$

exp



$$G(0) = 15$$

$$G(s) = K \frac{(s+3)(s+4+j2)(s+4-j2)}{(s+2+j2)(s+2-j2)} \Big|_{s=0} = 15$$

16

$$K=2$$

$$G(s) = \frac{2s^3 + 22s^2 + 88s + 120}{s^2 + 4s + 8}$$

exp Find Complex Frequency.

① $(2e^{-100t} + e^{-200t}) \sin 2000t$

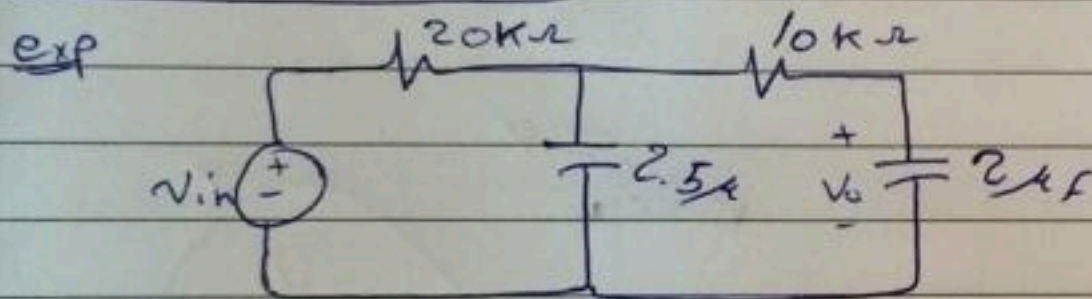
$-200 + j2000$, $-100 + j2000$

$-200 + j2000$, $-200 - j2000$

② $(2 - e^{-10t}) \cos(4t + \phi)$

$j4$, $-j4$

$-10 + j4$, $-10 - j4$



a) Plot $Z_{in}(s)$

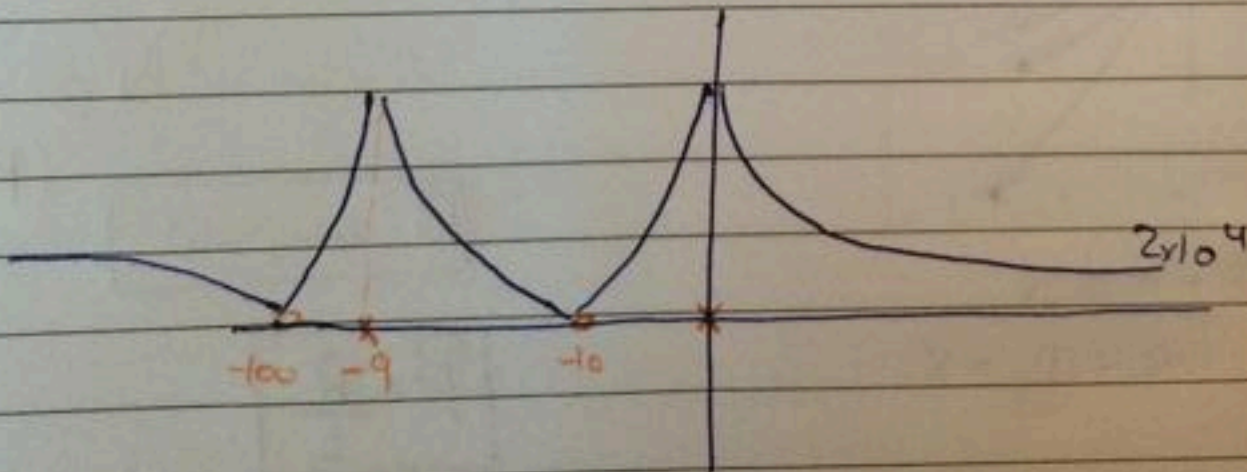
b) critical freq.

c) evaluate

$Z(-10)$, $Z(-95)$

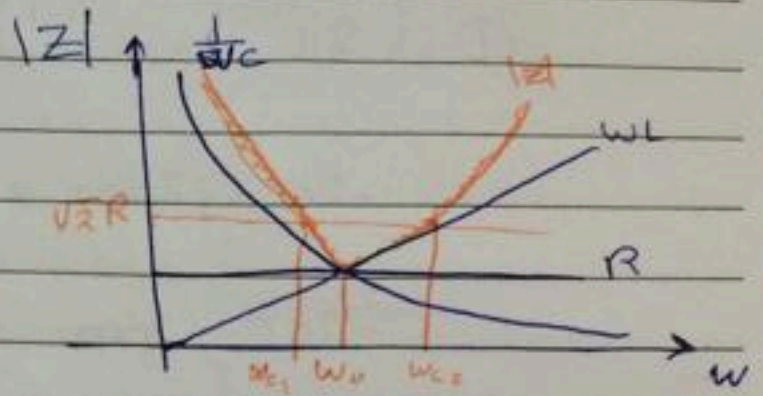
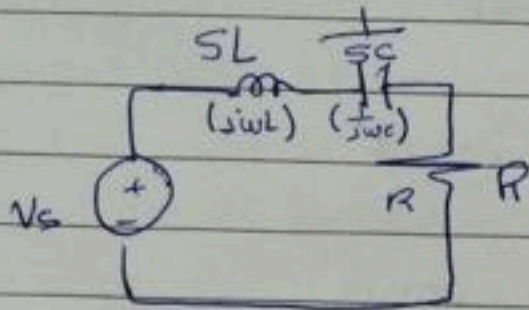
$Z(-50)$, $Z(10 \mu P)$

① $Z_{in} = 2 \times 10^4 \frac{(s+10)(s+100)}{s(s+90)}$



Resonance:-

R, L, C, Series and Parallel



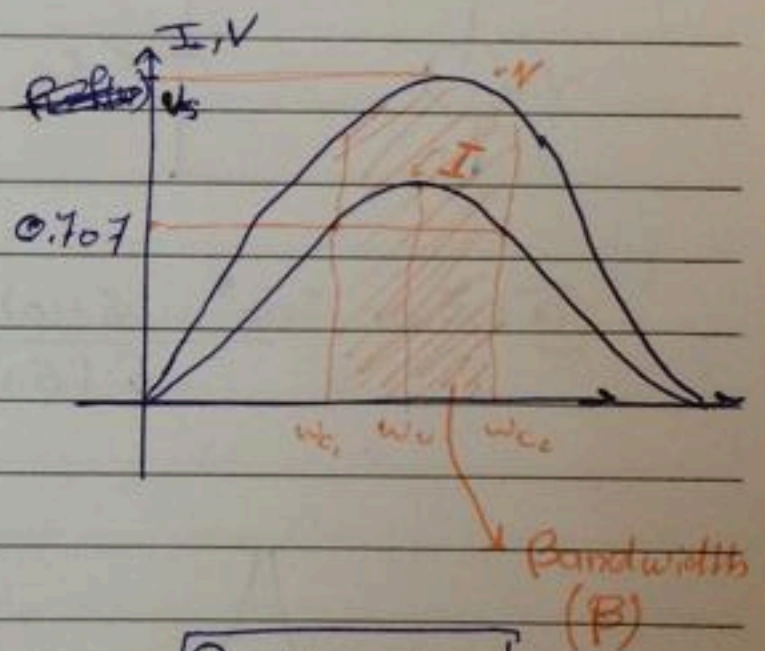
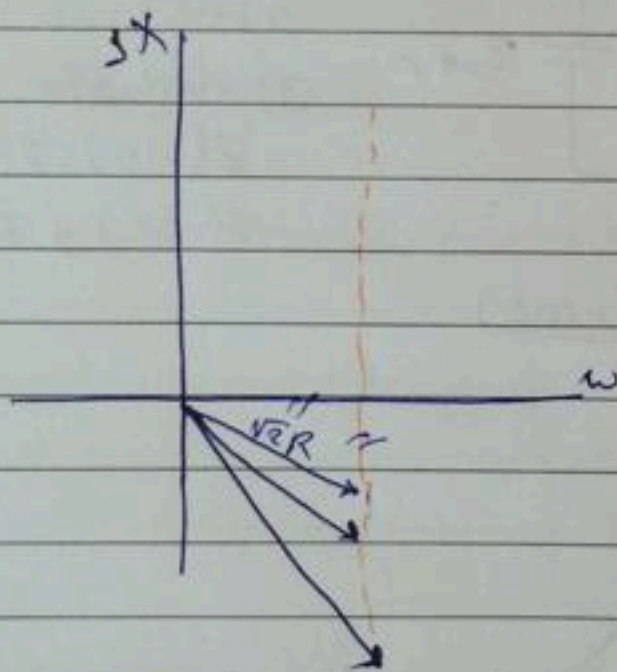
$$Z = R + j(\omega L - \frac{1}{\omega C})$$

$$|Z| = \sqrt{R^2 + (\omega L - \frac{1}{\omega C})^2}$$

ω_{c1} : Lower cut-off freq.

ω_{c2} : upper " " "

$$\omega L - \frac{1}{\omega C} = 0 \Rightarrow \omega = \frac{1}{\sqrt{LC}}$$



$$Z = R \pm jX$$

$$\omega L - \frac{1}{\omega C} = X$$

$$B = \omega_{c2} - \omega_{c1}$$

$$B = \frac{R}{L}$$

$$\omega^2 LC - 1 = \omega CX$$

$$\omega^2 LC - CX\omega - 1 = 0$$

$$\omega^2 - \frac{CX\omega}{LX} - \frac{1}{LC} = 0$$

81

$$w^2 - \frac{X}{L} w - \frac{1}{LC} = 0$$

$$w = \sqrt{\left(\frac{X}{2L}\right)^2 + \frac{1}{LC}} + \frac{X}{2L}$$

$$R = \pm X$$

$$w_{c1, c2} = \sqrt{\left(\frac{R}{2L}\right)^2 + \frac{1}{LC}} \pm \frac{R}{2L}$$

* Quality Factor:-

$$Q_0 = \frac{w_0 L}{R} = \frac{1}{w_0 C R} = \frac{\sqrt{L} \sqrt{L}}{R \sqrt{L} \sqrt{C}} = \frac{1}{R} \sqrt{\frac{L}{C}}$$

$$Q_0 = \frac{w_0}{\beta}$$

$$\Rightarrow V_L = V_s \frac{wL}{Z}$$

$$\Rightarrow V_C = V_s \frac{1}{wCZ}$$

$$\bullet V_L(w_0) = V_s \frac{w_0 L}{R} = Q_0 V_s$$

$$\bullet V_C(w_0) = V_s \frac{1}{w_0 C R} = Q_0 V_s$$

$$\bullet V_L(w_{c1, c2}) = V_s \frac{w_{c1} L}{\sqrt{2} R} = 0.707 V_{max}$$

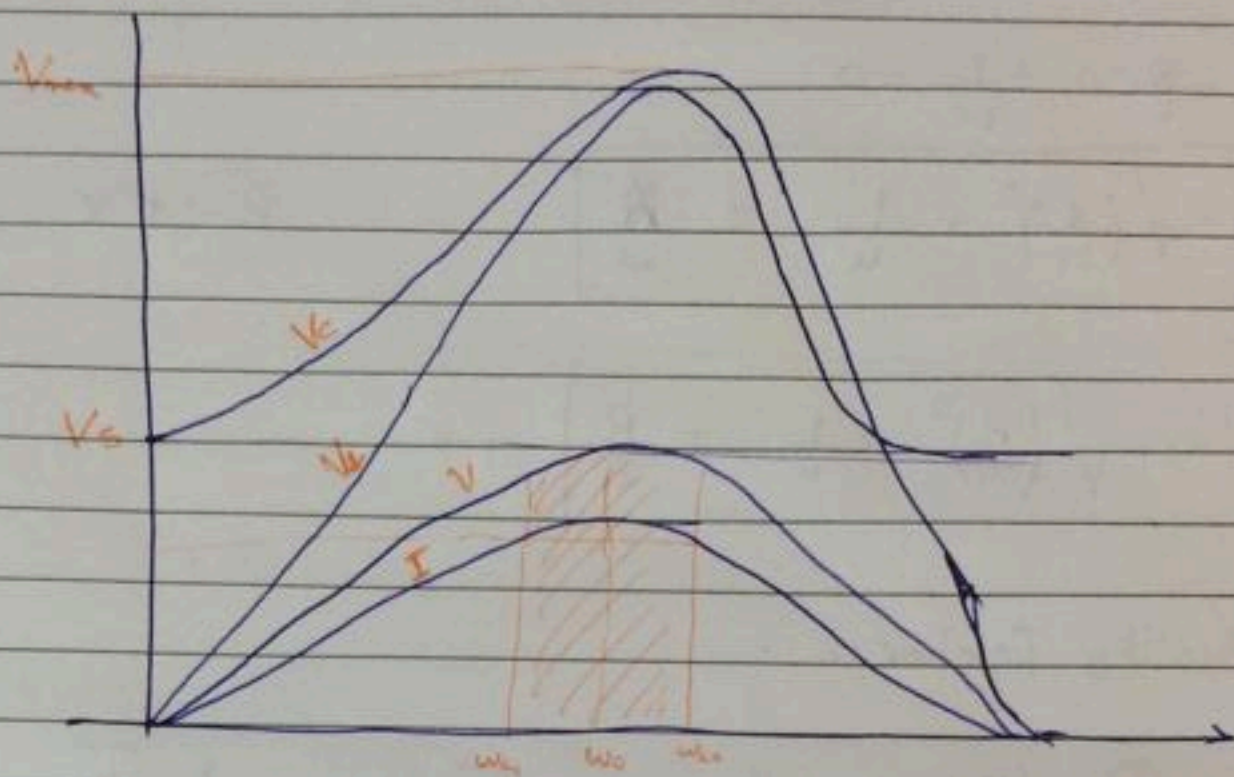
$$\bullet V(w_{c1}) = 0.707 V_{max}$$

$$\bullet V_L(w=\infty) = V_s$$

$$\bullet V(0) = V_s$$

$$\bullet V_L(0) = 0$$

$$\bullet V(\infty) = 0$$



$$Q_s = \frac{X_s}{R_s} \quad , \quad Q_p = \frac{R_p}{X_p}$$

proof $Z = R + j(\omega L - \frac{1}{\omega C})$

$$\frac{Z}{R} = \left(1 + j \left(\frac{\omega L}{R} + \frac{1}{\omega C R} \right) \right) * \frac{\omega_0}{\omega}$$

$$= 1 + j \left(\frac{\omega_0 L}{R} \cdot \frac{\omega}{\omega_0} - \frac{1}{\omega_0 C R} \cdot \frac{\omega_0}{\omega} \right)$$

$$= 1 + jQ \left(\frac{\omega}{\omega_0} - \frac{\omega_0}{\omega} \right)$$

$$= 1 + jQ \frac{\omega^2 - \omega_0^2}{\omega_0 \omega} = 1 + jQ \frac{(\omega - \omega_0)(\omega + \omega_0)}{\omega_0 \omega}$$

$$\boxed{\omega \approx \omega_0}$$

$$= 1 + jQ \frac{(\omega - \omega_0) 2\omega_0}{\omega_0^2} = 1 + jQ \frac{2(\omega - \omega_0)}{\omega_0}$$

(20)

$$\frac{Z}{R} = 1 + j \frac{\omega - \omega_0}{\beta/2} = 1 + jN$$

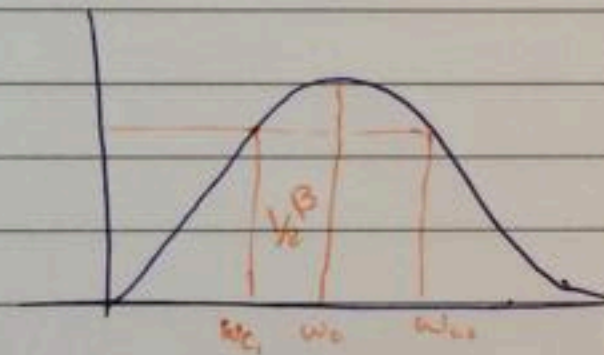
$$\omega_{c1, c2} = \sqrt{\left(\frac{R}{2L}\right)^2 + \frac{1}{LC}} \mp \frac{R}{2L}$$

$$= \sqrt{\left(\frac{\omega_0}{2Q}\right)^2 + \omega_0^2} \mp \frac{\beta}{2}$$

$$\omega_{c1, c2} = \omega_0 \sqrt{\left(\frac{1}{2Q}\right)^2 + 1} \mp \frac{\beta}{2}$$

$$\underline{Q \gg 5}$$

$$\omega_{c1, c2} = \omega_0 \mp \frac{\beta}{2}$$

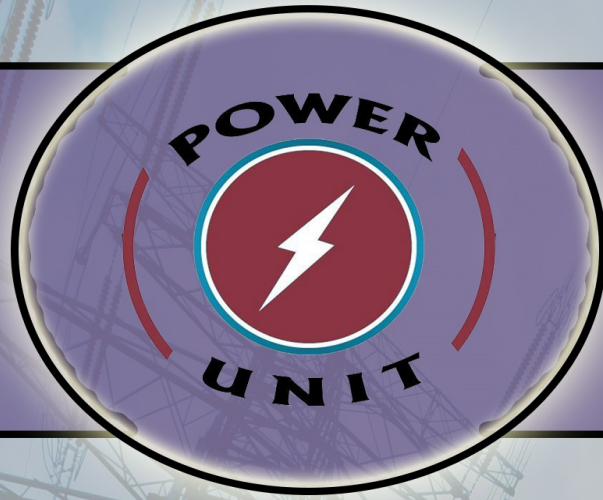


$$Z(s) = R + sL + \frac{1}{sC}$$

$$= \frac{RLs + s^2LC + 1}{sC} = \frac{LCs^2 + RLs + \frac{1}{C}}{Cs}$$

$$= \frac{\left(s^2 + \frac{R}{L}s + \frac{1}{LC}\right)L}{Cs}$$

(21)



Circuits II Notebook

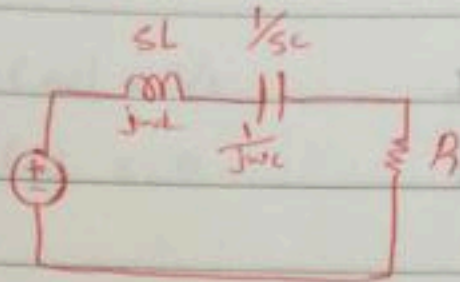
Dr. Nabeel Tawalbeh

By . Yazan Abawi

بأفكارنا نبدع

* Resonance Frequency *

* RLC series & Parallel *

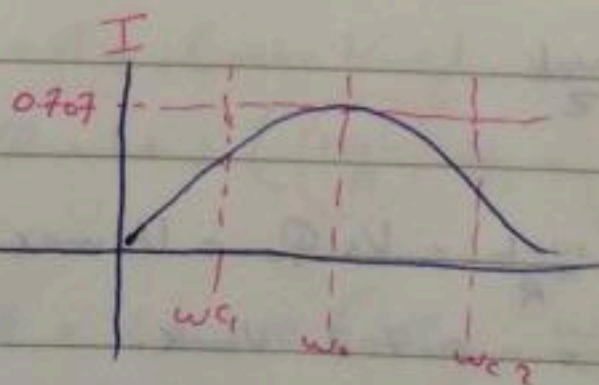
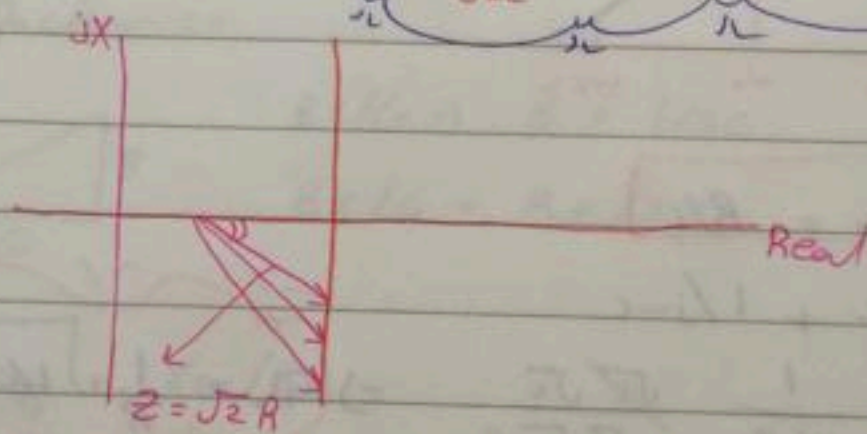


$$* Z = R + j(\omega L - \frac{1}{\omega C})$$

$$* R, \omega L, \frac{1}{\omega C}$$

$$* Z = \sqrt{R^2 + (\omega L - \frac{1}{\omega C})^2}$$

$$* \omega L - \frac{1}{\omega C} = 0 \Rightarrow \omega_0 = \frac{1}{\sqrt{LC}} \Rightarrow \text{resonant frequency}$$



1 at 0 or ∞ = 0. C ⇒ 0

$$I_{max} = V/R$$

$$Z = R + jX$$

$$\omega L - \frac{1}{\omega C} = X$$

$$\omega^2 LC - \omega C X \omega - 1 = 0$$

$$\omega^2 - \frac{CX}{LC} \omega - \frac{1}{LC} = 0$$

$$\omega^2 - \frac{X}{L} \omega - \frac{1}{LC} = 0$$

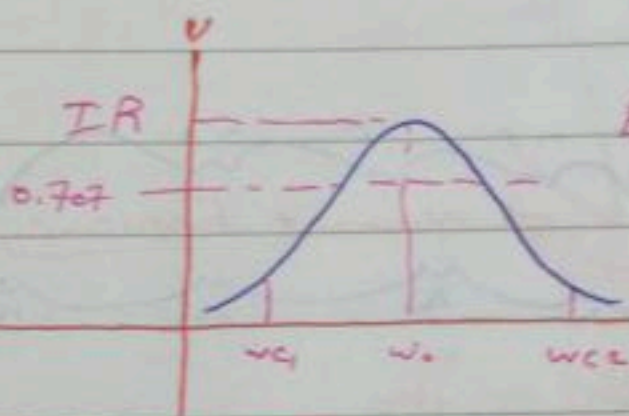
$$\omega = \sqrt{\left(\frac{X}{2L}\right)^2 + \frac{1}{LC}} + \frac{X}{2L}$$

$$X = -R, +R \Rightarrow R = \pm X$$

$$\text{if } X = -R$$

$$\omega_{c1, c2} = \sqrt{\left(\frac{R}{2L}\right)^2 + \frac{1}{LC}} \mp \frac{R}{2L}$$

lower -ve freq
upper +ve freq



$$\text{Bandwidth (B)} = \omega_{c2} - \omega_{c1}$$

Band Pass Filter

$$* \quad B = \omega_{c2} - \omega_{c1} = R/L$$

$$* \quad Z = R + j\omega L + 1/j\omega C$$

$$* \quad Q_0 = \frac{\omega_0 L}{R} = \frac{1}{\omega_0 R C} \frac{\sqrt{L} \sqrt{L}}{\sqrt{R} \sqrt{C} R} \Rightarrow Q_0 = \frac{1}{R} \sqrt{L/C}$$

$$Q_0 = \frac{\omega_0}{B}$$

$$V_L = V_s - \frac{\omega_0 L}{Z}$$

$$\text{at resonance } V_L(\omega_0) = V_s \frac{\omega_0 L}{R} = V_s Q_0 = V_{L \text{ max}}$$

$$V_L(\omega_{c1}, \omega_{c2}) = V_s \frac{\omega_0 L}{\sqrt{2} R} = 0.707 V_{L \text{ max}}$$

$$V_L(\omega = \infty) = V_s$$

$$V_L(\omega = 0) = 0$$

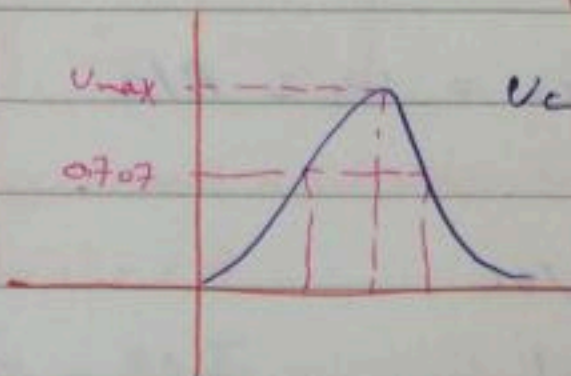
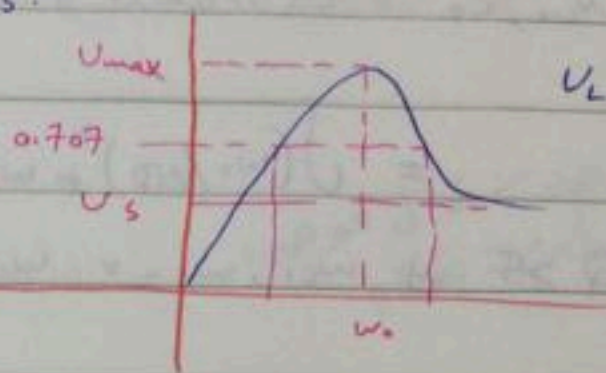
$$\text{at } 0 \Rightarrow V_L = 0$$

$$V_L = V_s \cdot \frac{1}{\omega_0 L Z}$$

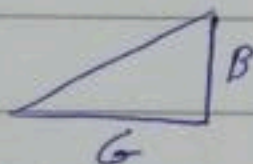
$$V_c(\omega) = V_s * \frac{1}{\omega \cdot C R} = Q_0 \cdot V_s$$

$$V_c(\infty) = V_s$$

$$V_c(0) = 0.$$



* Capacitor \Rightarrow



$$B \text{ C/G} \Rightarrow B = 1/X_C$$

$$B \text{ C/G} = B_P / X_P$$

$$Q_{\text{series}} = X_s / R_s$$

$$Q_{\text{parallel}} = R_P / X_P$$

$$Z = R + j(\omega L - 1/\omega C) = R \left(1 + j \left[\frac{\omega L}{R} - \frac{1}{\omega C R} \right] \right) \frac{\omega_0}{\omega}$$

$$Z/R = 1 + j \left(\frac{\omega_0 L}{R} \cdot \frac{\omega}{\omega_0} - \frac{1}{\omega_0 C R} \cdot \frac{\omega_0}{\omega} \right)$$

$$Z/R = 1 + jQ \left(\frac{\omega}{\omega_0} - \frac{\omega_0}{\omega} \right) = 1 + jQ \frac{\omega^2 - \omega_0^2}{\omega \omega_0}$$

$$= \frac{1 + jQ (\omega - \omega_0) (\omega + \omega_0)}{\omega \omega_0}$$

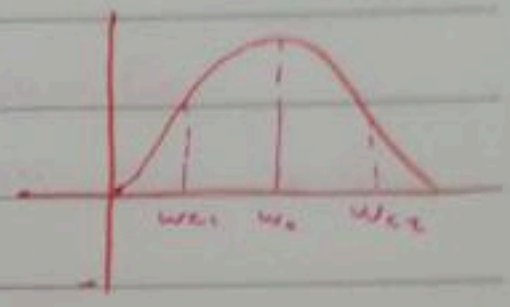
$$* \omega = \omega_0 \Rightarrow 1 + jQ^2 \frac{(\omega - \omega_0)}{\omega_0} = 1 + j \frac{\omega - \omega_0}{B/2} = 1 + jN$$

$$\omega_{c1, c2} = \sqrt{(R/2L)\beta + 1/LC} \mp R/2L$$

$$= \sqrt{(\omega_0/2Q)^2 + \omega^2} \mp \beta/2$$

$$Q \geq 5 \Rightarrow \omega_{c1, c2} = \omega_0 \sqrt{(1/2Q)^2 + 1} \mp \beta/2$$

$$Q \geq 5 : \omega_{c1, c2} = \omega_0 \mp \beta/2$$



$$Z(s) = R + Ls + 1/sC$$

$$Z(s) = \frac{RLs + sLC + 1}{sC}$$

$$\omega_{c1, c2} = \omega_0 \mp \frac{1}{2} \beta$$

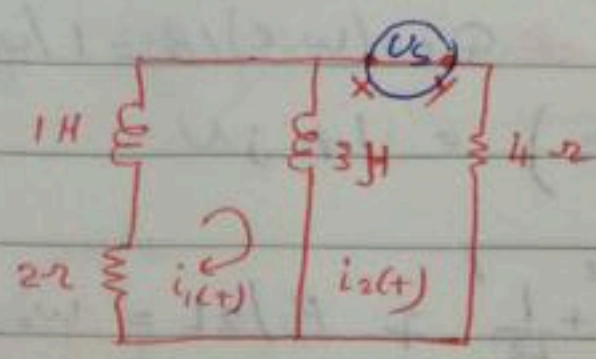
$$= \frac{LCS^2 + RLS + 1}{sC} \div LC$$

$$= \frac{s^2 + \frac{R^2}{C} + \frac{1}{LC}}{Cs/L}$$

ROCO

ROCO

* قال افترسك فارس Complex Frequency



* هار سقال فوجهور
بأسنة الكساي

* evaluate $i_2(t)$ by install a suitable source across Points

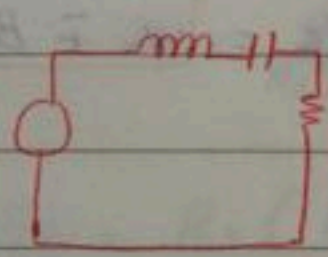
$x, y \quad i_1(0) = i_2(0) = 1A$

$$U_s = (4 + 3s) I_2 - 3s I_1$$

$$0 = -3s I_1 + (2 + 4s) I_2$$

$$I_2 = \frac{(2 + 4s) U_s}{3 \left(s^2 + \frac{22}{3} s + \frac{8}{3} \right)}$$

* * * * *

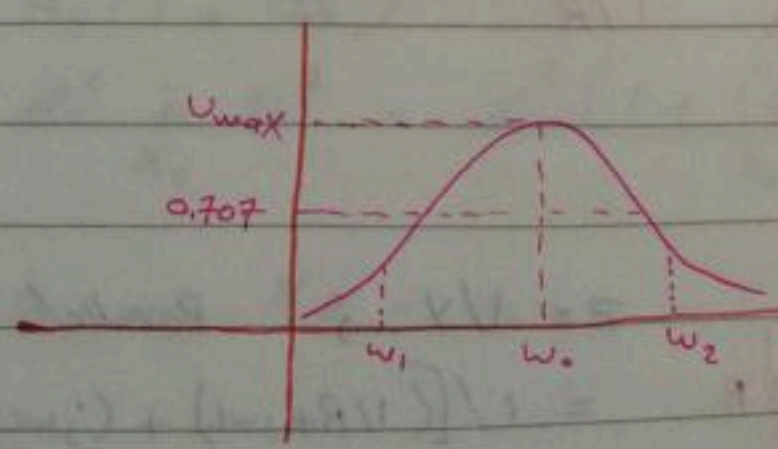


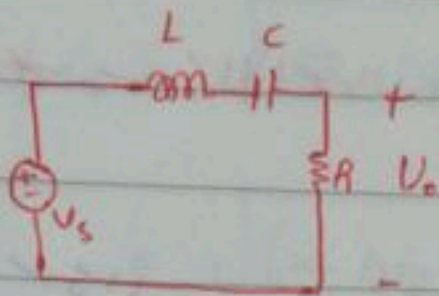
$$U_c = U_s * (X_c / Z)$$

$$U_c(\omega_0) = U_s * [(1/\omega_0 C) / R] = \phi_0 U_s$$

$$U_c(0) = U_s \quad U_c(\infty) = 0$$

$$U_c(\omega_{c1}) = 0.707 U_{max}$$





Series

* $Z = R + j(\omega L - 1/\omega C)$

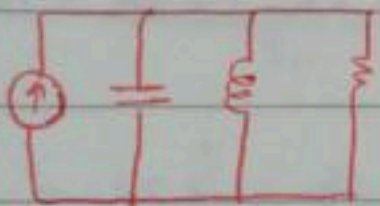
* $Q_{oS} = (\omega C)/R = 1/\omega_0 C R = \frac{1}{R} \sqrt{\frac{L}{C}}$

* $Z = 1 + j \left(\frac{\omega - \omega_0}{B/2} \right) = 1 + jN$

* $\omega_{c1}, \omega_{c2} = \sqrt{(R/2L)^2 + \frac{1}{LC}} \mp R/2L = \omega_0 \mp B/2$

* $B = R/C = \omega_0 / Q_0$

Parallel



$G + jB_C - jB_L$

$Y = G + j(\omega C - 1/\omega L) = 1 + jQ \left(\frac{\omega}{\omega_0} - \frac{\omega_0}{\omega} \right)$

$Q_{oP} = \omega_0 C R = \frac{R}{\omega_0 L} = R \sqrt{C/L} \Rightarrow$ *series قانون*

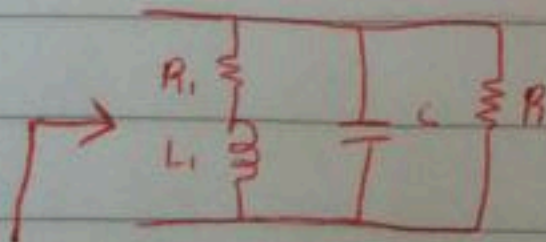
$\omega_{c1}, \omega_{c2} = \sqrt{(1/2RC)^2 + 1/LC} \mp 1/2RC = \omega_0 \mp B/2$
 $= \omega_0 \sqrt{1 + (1/2Q)^2} \mp 1/2Q$

$B = \frac{1}{RC} = \frac{\omega_0}{Q_0}$

* $G \rightarrow$ Conductance $1/R$

* $B \rightarrow$ Susbetance

* $Y \rightarrow$ Admittance



$Z = 1/Y \rightarrow$ Parallel

$= 1 / \left[(1/R + j\omega L) + (j\omega C) + (1/R) \right]$

$= \text{Real} + \text{Imaginary}$

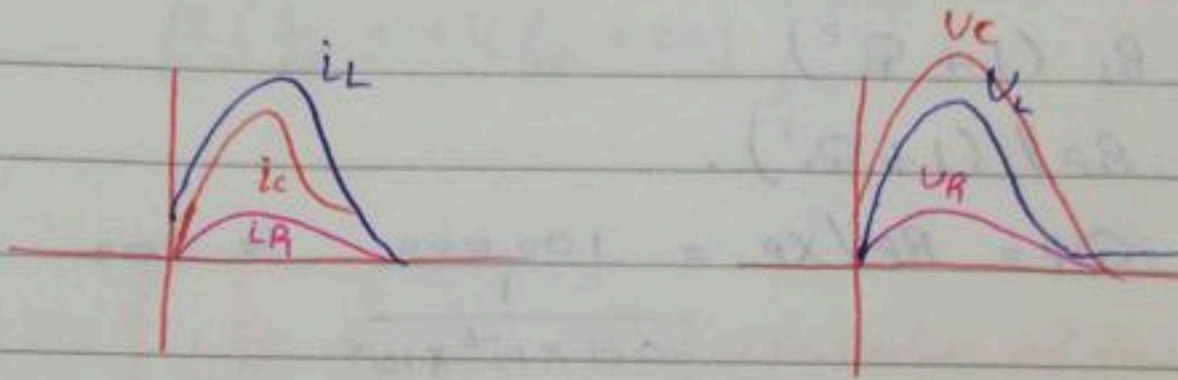
$$I_{mg} = 1/R_2 + j\omega C + 1/(R + j\omega L)$$

Req

$$\omega_0 = \sqrt{\frac{1}{LC} - \left(\frac{R_1}{L}\right)^2}$$

Pure RLC

* لا يكون موج



$$Z_s \rightarrow Z_p$$

للتحويل من s إلى p وبالضبط

$$* Z_s = R_s + jX_s \xrightarrow{\text{بني اوكلا}} Y$$

$$* Y = 1/Z = 1/(R_s + jX_s) = R_s / (R_s^2 + jX_s^2) - j \frac{X_c}{R_s^2 + X_s^2}$$

$$* Y = G_p - jB_p = \frac{1}{R_p} - j \frac{1}{X_p}$$

$$* \frac{1}{R_p} = \frac{R_s}{R_s^2 + X_s^2} \Rightarrow R_p = \frac{R_s^2 + X_s^2}{R_s} = R_s \left(1 + \left(\frac{X_s}{R_s}\right)^2\right)$$

$$R_p = R_s (1 + Q^2)$$

* هاتين اهم معادلتين موضحتين في الشيفاه

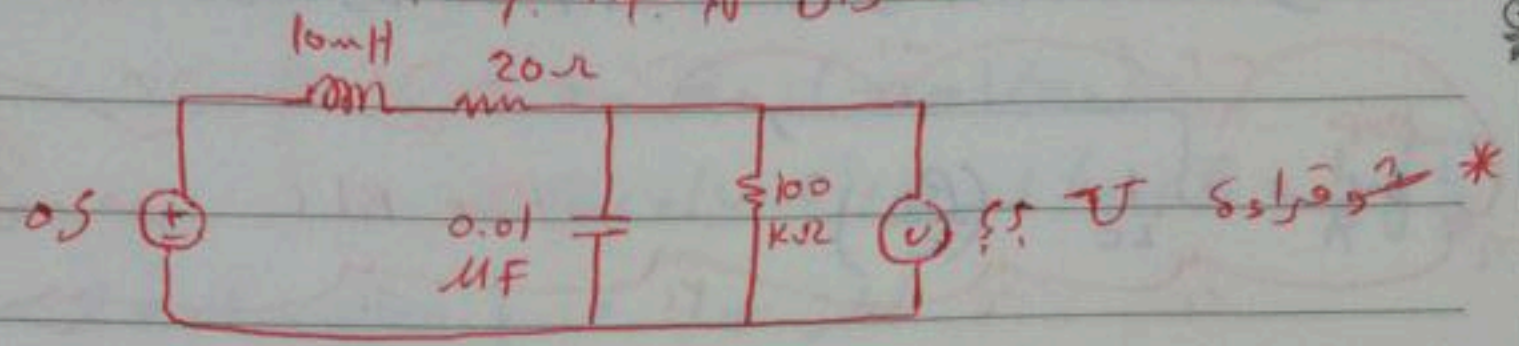
$$\text{for } Q \geq 5 \rightarrow R_p = R_s \cdot Q^2$$

$$R_p = R_s (1 + Q^2)$$

$$X_p = \frac{R_s^2 + X_s^2}{X_s} = \frac{X_s^2 \left(1 + \left(\frac{R_s}{X_s}\right)^2\right)}{X_s} = X_s \left(1 + \frac{1}{Q^2}\right)$$

$$* Q \geq 5 \Rightarrow X_p = X_s$$

قال في كتاباً جدياً



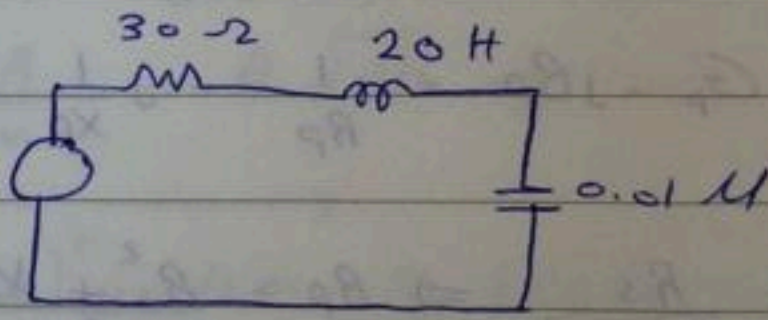
$$R_p = R_s (1 + Q_o^2)$$

$$R_s = R_p / (1 + Q_o^2)$$

$$Q_p = R_p / X_p = \frac{100000}{0.01 \times 10^{-6} \times 10^5} = 100$$

$$\omega_o = \frac{1}{\sqrt{0.01 \times 1.0 \times 10^{-9}}} = 10^5$$

بعد التحويل



ROCO

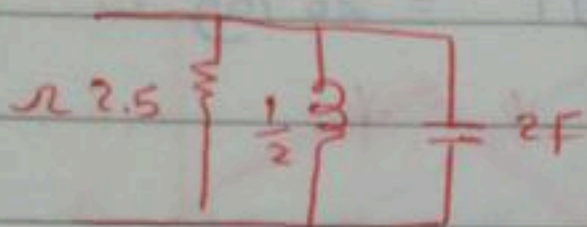
ROCO

* Scalling

$$H(s) = \frac{2s}{2(s+10)(s+20)} \Rightarrow \text{Func. كمال ال } K_m \text{ وانقسم كل } s \text{ بـ } K_F$$

$$H(s) = \left[\frac{2(s/K_F)}{2\left(\frac{s}{K_F} + 10\right)\left(\frac{s}{K_F} + 20\right)} \right] K_m$$

*



Find the value of each element after scaled by $K_F = 5 \times 10^6$

$$K_m = 200$$

$$R_{new} = 500$$

$$L_{new} = 200 \text{ MH}$$

$$C_{new} = 200 \text{ pF}$$

$$= L \frac{K_m}{K_F} = \frac{2}{K_m K_F}$$

* Bode Plot *

$$H(s) = \frac{k(s+z_1)}{(s+p_1)(s+p_2)}$$

$$H(j\omega) = \frac{k(j\omega + z_1)}{(j\omega + p_1)(j\omega + p_2)} = \frac{k z_1 (1 + j\omega/z_1)}{p_1 p_2 (1 + j\omega/p_1)(1 + j\omega/p_2)}$$

$$= k \frac{z_1}{p_1 p_2} \cdot \frac{|1 + \omega/z_1|}{|1 + \omega/p_1| |1 + \omega/p_2|}$$

$$|H(j\omega)| = k \cdot \frac{|1 + j\omega/z_1|}{|1 + j\omega/p_1| |1 + j\omega/p_2|}$$

$$H_{dB} = 20 \log |H(j\omega)| = 20 \log k + 20 \log \left| 1 + j \frac{\omega}{z} \right| - 20 \log \left| 1 + j \omega / p_1 \right| - 20 \log \left| 1 + j \omega / p_2 \right|$$

$$\angle H(j\omega) = \tan^{-1} \omega / z_1 - \tan^{-1} \omega / p_1 - \tan^{-1} \omega / p_2$$

الزاوية موجبة

$$H(j\omega) = j\omega$$

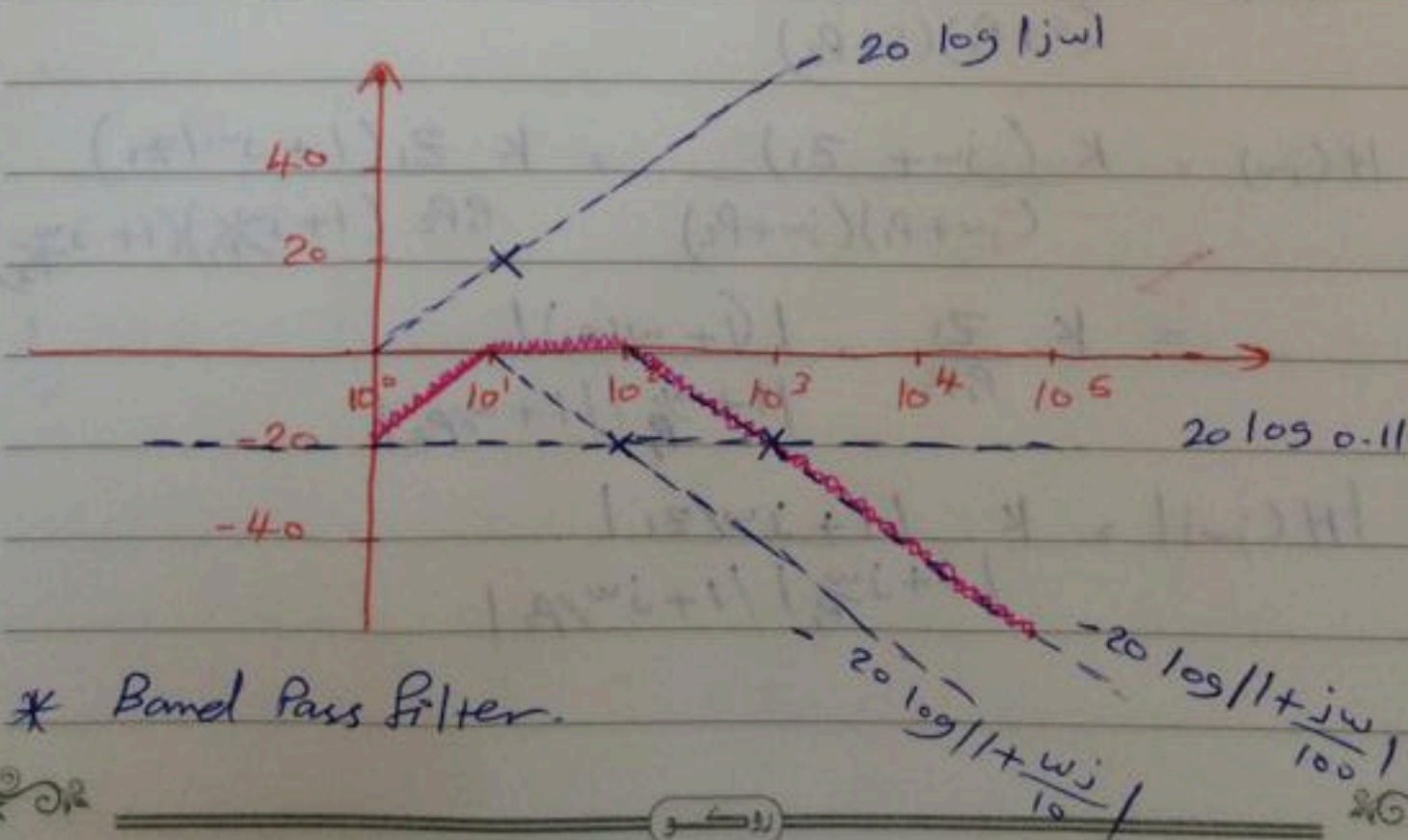
$$H_{dB} = A_{dB} = 20 \log |H(j\omega)| = 20 \log \omega$$

* * * * *

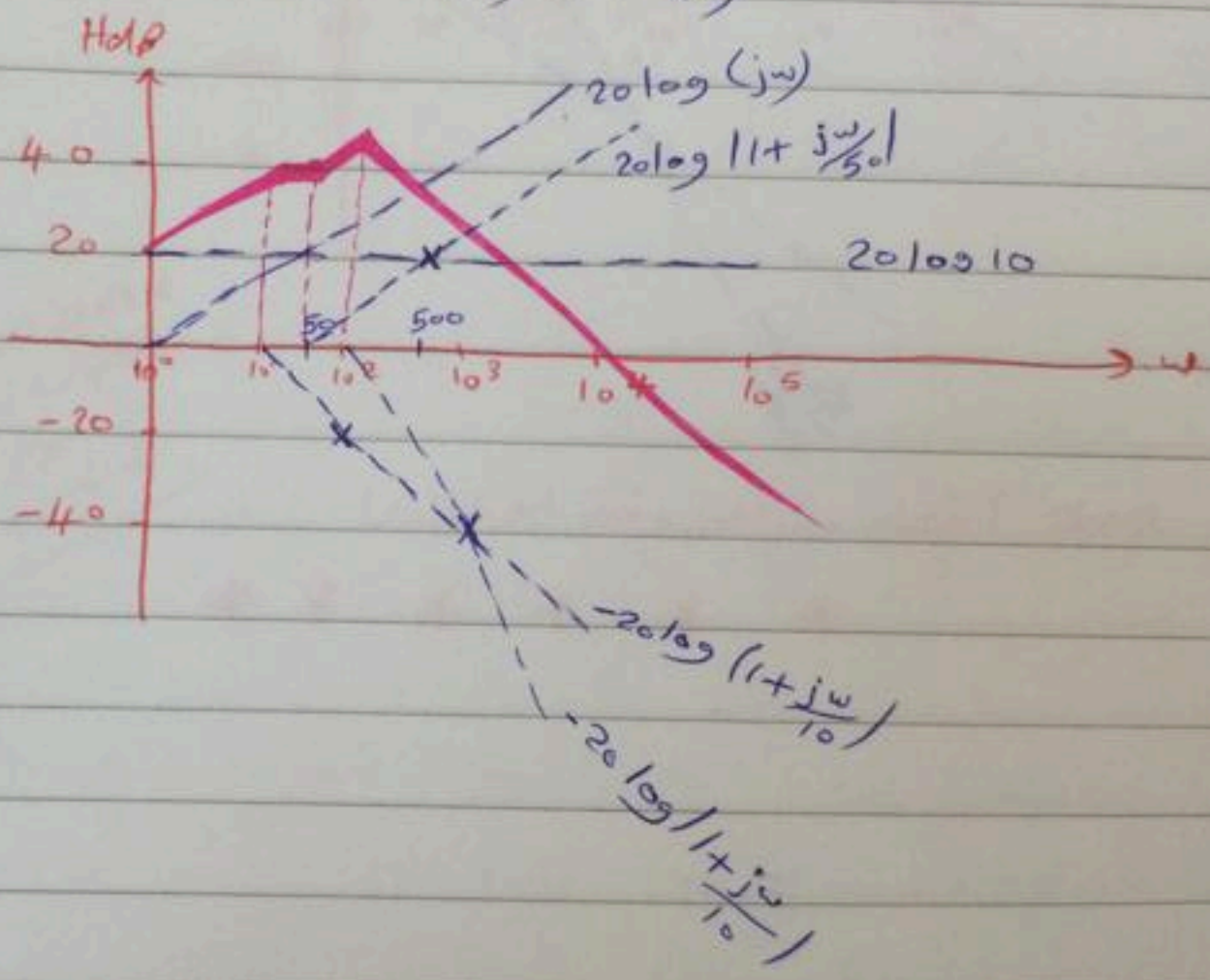
$$H(s) = \frac{110s}{s^2 + 110s + 1000} = \frac{110s}{(s+10)(s+100)}$$

$$H(s) = \frac{110s}{100 * 10 (1 + \frac{s}{10})(1 + \frac{s}{100})} \Rightarrow \text{Standard form}$$

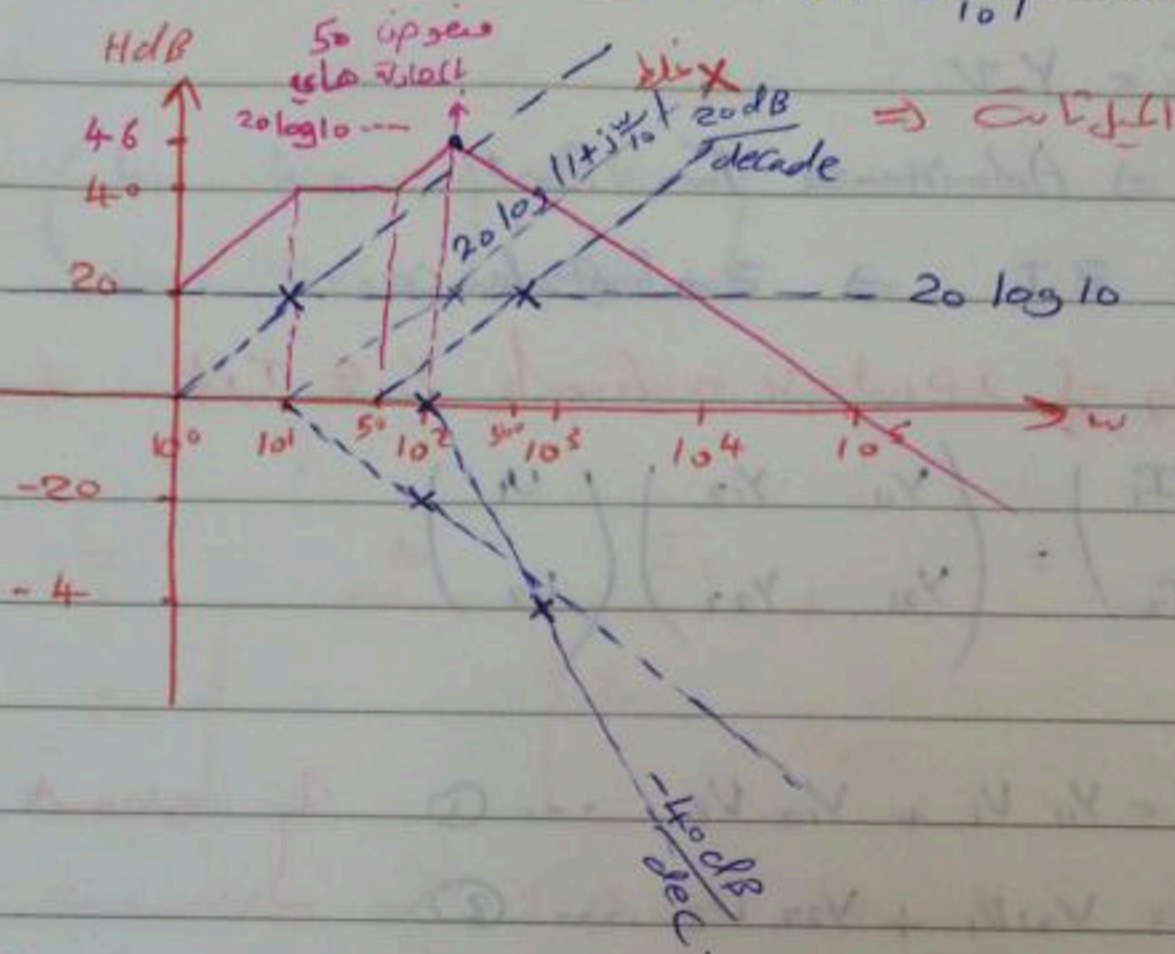
$$= 0.11 \frac{s}{(1 + \frac{s}{10})(1 + \frac{s}{100})}$$



$$\begin{aligned}
 H(s) &= \frac{2 \times 10^4 s (s + 50)}{(s + 100)(s^2 + 110s + 10000)} \\
 &= \frac{2 \times 10^4 \times 50 s (1 + s/50)}{100 \times 10 \times 100 (1 + s/100)(1 + s/10)} \\
 &= 10 \frac{(s + s/50)}{(1 + s/100)(1 + s/10)}
 \end{aligned}$$



$$H(s) = \frac{100s(1 + s/50)}{(1 + s/10)(1 + s/100)^2} = 20 \log 10 + 20 \log(s) + 20 \log(1 + \frac{s}{50}) - 20 \log |1 + \frac{s}{10}| - 2 \times 20 \log |1 + \frac{s}{100}|$$

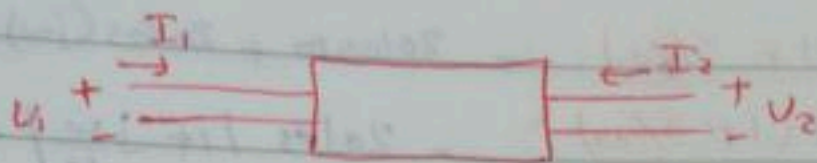


* العنصر الثاني
الكتلة عند 100
السؤال الثاني
ال 10 - 50

* Phase أجهته وازا من وقت افناضها

* * * * *

* 2-Point Network *



* $I = YV$

* $Y \Rightarrow$ Admittance Two Port.

* $V = ZI \Rightarrow Z \Rightarrow$ Impedance.

* Key of 2-Port is to find Y, Z, T, h .

$$\begin{pmatrix} I_1 \\ I_2 \end{pmatrix} = \begin{pmatrix} Y_{11} & Y_{12} \\ Y_{21} & Y_{22} \end{pmatrix} \begin{pmatrix} V_1 \\ V_2 \end{pmatrix}$$

* $I_1 = Y_{11} V_1 + Y_{12} V_2 \dots \textcircled{1}$ } shunt cct

* $I_2 = Y_{21} V_1 + Y_{22} V_2 \dots \textcircled{2}$

* $SC_{\text{out}}^{\text{port}}$, i.e. $V_2 = 0 \rightarrow Y_{11} = I_1 / V_1 \quad Y_{21} = I_2 / V_1$

* $SC_{\text{in}}^{\text{port}}$, i.e. $V_1 = 0 \rightarrow Y_{22} = I_2 / V_2 \quad Y_{12} = I_1 / V_2$

* for impedance go

$II \Leftarrow I_2 = 0 \Rightarrow O.C.$

• $V_1 = Z_{11} I_1 + Z_{12} I_2 \dots \textcircled{1}$ } $Z_{11} = V_1 / I_1, Z_{21} = V_2 / I_1$

• $V_2 = Z_{21} I_1 + Z_{22} I_2 \dots \textcircled{2}$ } $Z_{22} = V_2 / I_2, Z_{12} = V_1 / I_2$

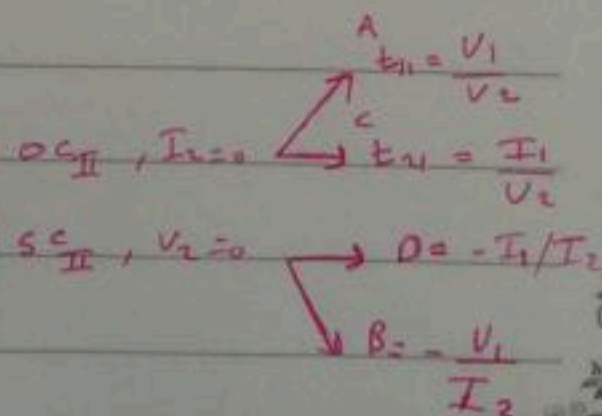
* for Transfer 2-Port go

$I = I_1 = 0 \Rightarrow O.C.$

$$\begin{pmatrix} V_1 \\ I_1 \end{pmatrix} = \begin{pmatrix} A & B \\ C & D \end{pmatrix} \begin{pmatrix} V_2 \\ -I_2 \end{pmatrix}$$

$V_1 = t_{11} V_2 - t_{12} I_2 \dots \textcircled{1}$

$I_1 = t_{21} V_2 - t_{22} I_2 \dots \textcircled{2}$



* $V_1 = h_{11} I_1 + h_{12} U_2$

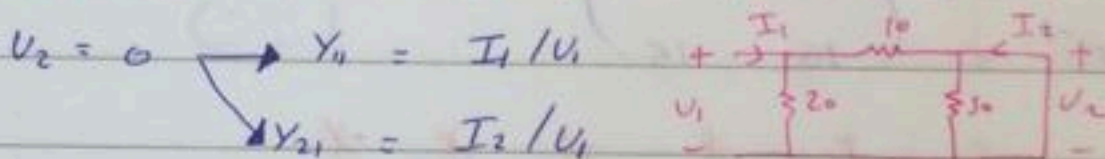
$I_2 = h_{21} I_1 + h_{22} U_2$

* $\begin{pmatrix} I_1 \\ U_2 \end{pmatrix} = \begin{pmatrix} h_{11} & h_{12} \\ h_{21} & h_{22} \end{pmatrix} \begin{pmatrix} U_1 \\ I_2 \end{pmatrix}$

* * * * *



* Find all parameters of 2-port network



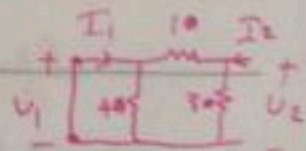
$Y_{11} = I_1 / U_1 = 1 / (10 // 20) \Rightarrow$ *30 اهمه لانه 30 اهمه Parallel لانه*

$Y_{21} = \frac{I_2}{U_1} = Y_{11} \cdot \frac{I_2}{I_1}$
 $= Y_{11} \cdot \frac{-I_1 (20/30)}{I_1} = -2/3 Y_{11}$

اعتبرناه الريس لانو لانو SiC والسبب لانو اجهه موصلي

$U_1 = 0 \rightarrow Y_{22} = I_2 / U_2 = 1 / (10 // 30)$

$Y_{12} = I_1 / U_2 = I_1 / U_2 + \frac{I_2}{I_2}$
 $= Y_{22} \cdot \frac{-I_2 (30/40)}{I_2} = -3/4 Y_{22}$

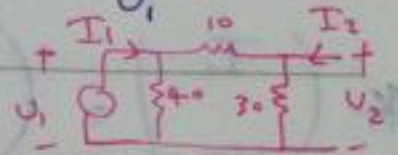


$\begin{pmatrix} I_1 \\ I_2 \end{pmatrix} = \begin{pmatrix} 3/20 & -1/10 \\ -1/10 & 4/30 \end{pmatrix} \begin{pmatrix} U_1 \\ U_2 \end{pmatrix} \Rightarrow Z = Y^{-1}$ *طريقة 1*

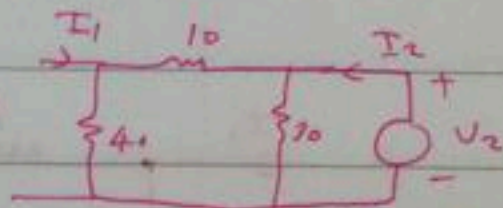
$$* Z_{11} = \left. \frac{V_1}{I_1} \right|_{I_2=0} = 40/20 = 80/6$$

$$* Z_{21} = \left. \frac{V_2}{I_1} \right|_{I_2=0} = \frac{V_2}{I_1} \cdot \frac{U_1}{U_1} = Z_{11} \cdot \frac{U_2}{U_1}$$

$$= Z_{11} \cdot \frac{U_1 (30/40)}{U_1}$$



$$* Z_{22} = \left. \frac{V_2}{I_2} \right|_{I_1=0} = 30/30 = 15 \Omega$$



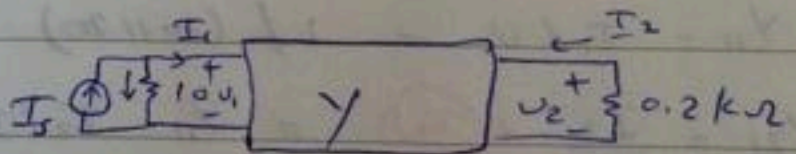
$$* Z_{12} = \frac{U_1}{I_2} \cdot \frac{U_2}{U_2} = Z_{22} \cdot \frac{U_1}{U_2} = Z_{22} \cdot \frac{U_2 (20/30)}{U_2}$$

$$= 15 \cdot \frac{20}{30} = 10 \Omega$$

$$* Z = Y^{-1} = \begin{pmatrix} 80/6 & 10 \\ 10 & 15 \end{pmatrix}$$

* * * * *

$$Y = \begin{pmatrix} 0.2 & 0.05 \\ 0.04 & 0.06 \end{pmatrix}$$



Find $G_u = U_2/U_1$

$G_i = I_2/I_1$

$\frac{P_2}{P_1} = G_p$

Z_{in}, Z_{out}

$Z_{in} = \frac{U_1}{I_1}$

$Z_{out} = \frac{U_2}{I_2}$

$$I_1 = 0.2 V_1 - 0.05 V_2 \dots \textcircled{1}$$

$$I_2 = 0.04 V_1 + 0.06 V_2 \dots \textcircled{2}$$

$$I_2 = -\frac{V_2}{200} = -0.005 V_2 \dots \textcircled{3} \quad \text{Condition} \quad \text{فيها} \quad \text{ال} \quad \text{output} \leftarrow \text{load} \quad \text{is}$$

$$I_1 = I_s - 0.1 V_1 \dots \textcircled{4} \rightarrow \text{Kcl Loop}$$

* $\textcircled{3}$ بعوضها بـ $\textcircled{2}$ و $\textcircled{4}$ بـ $\textcircled{1}$

$$3 \rightarrow 2 \quad I_2 = 0.04 V_1 + 0.065 V_2 = 0 \rightarrow \text{فيها الطرف الثاني}$$

$$4 \rightarrow \textcircled{1} \quad I_s = 0.3 V_1 - 0.05 V_2$$

$$V_1 = \begin{pmatrix} 0 & \dots \\ I_s & \dots \\ \dots & \dots \end{pmatrix} = \dots V_s$$

* $\text{فيها} \quad \text{Mode} \quad 5-1$

بال Calculator $\text{فيها} \quad \text{فيها}$

$$V_2 = -80/43 V_s$$

$$V_1 = 130/43 V_s \Rightarrow \frac{V_2}{V_1} = \frac{-80 * 43}{43 * 130} = -8/13$$

$$I_1 = 0.2 V_1 - 0.05 V_2$$

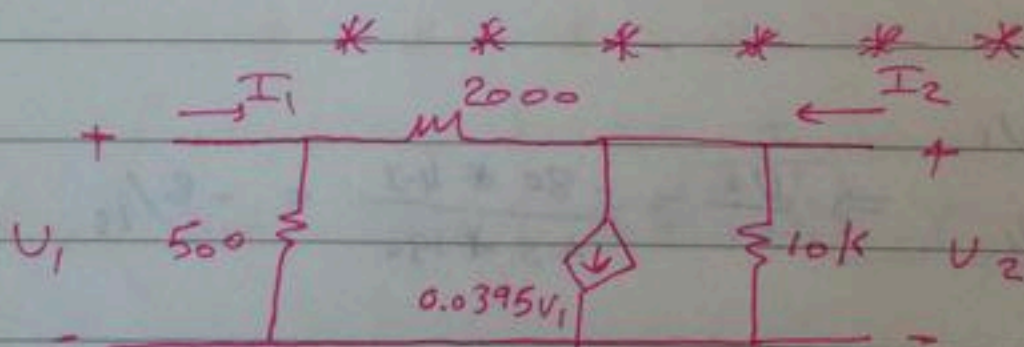
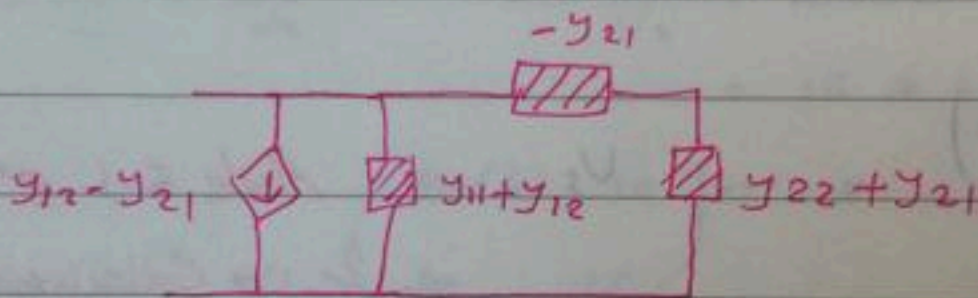
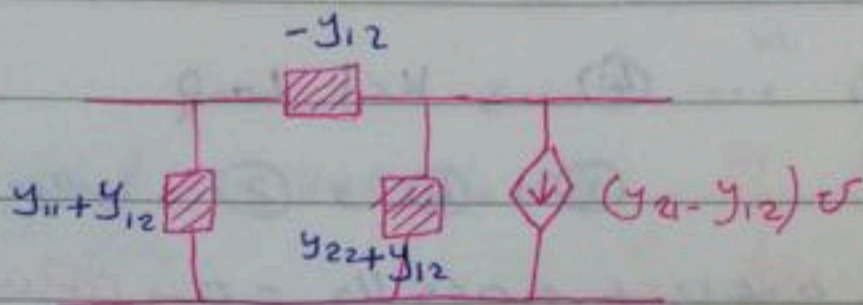
$$I_2 = -V_2/0.2 = -80/-43 * 1/0.2 =$$

$$I_1 = I_s - 0.1 V_1$$

$$I_1 = Y_{11} V_1 + Y_{12} V_2 \quad \text{--- (1)}$$

$$I_2 = Y_{21} V_1 + Y_{22} V_2 + \underbrace{Y_{12} V_1 - Y_{12} V_1}_{\text{ما بتأثرنا ضايفناها}}$$

$$I_2 - (Y_{21} - Y_{12}) V_1 = Y_{12} V_1 + Y_{22} V_2 \quad \text{--- (2)}$$



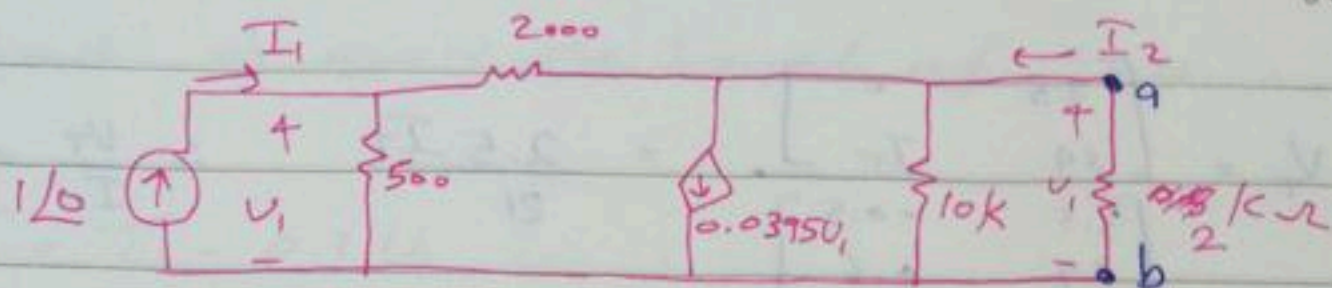
$$Y_{12} = -1/2000 = -0.5 \text{ S}$$

$$Y_{11} + Y_{12} = 1/500 \Rightarrow Y_{11} = 1/500 - (-0.5) = 2.5 \text{ mS}$$

$$Y_{22} + Y_{12} = 1/10000 \Rightarrow Y_{22} = 1/10000 - (-0.5) = 0.6 \text{ mS}$$

$$(Y_{21} - Y_{12}) V_1 = 0.0395 V_1$$

$$Y_{21} = 0.0395 + Y_{12} = 0.0395 * 1000 - 0.5 = 39 \text{ mS}$$



هناك نفس سؤال المرة الماضية على طلب فيه Z_{out} , Z_{in} , G_i , G_v

$$I_1 = 2.5V_1 - 0.5V_2 \quad \text{--- (1)}$$

$$I_2 = 39V_1 + 0.6V_2 \quad \text{--- (2)}$$

$$I_1 = 1 \text{ mA } V_1 \quad \text{--- (3)}$$

$$I_2 = -2V_2$$

$$2.5V_1 - 0.5V_1 = 1 \quad \text{--- (5)}$$

$$39V_1 + 2.6V_2 = 0 \quad \text{--- (6)}$$

$$V_1 = 0.1 \quad V_2 = -1.5 \text{ V}$$

mode 5-1 *تجربة*

$$G_v = -15$$

$$G_i = \frac{-2(-1.5)}{1} = 3$$

$$G_p = G_v G_i = 45$$

$$Z_{in} = V_1 / I_1 = 0.1 \text{ k}\Omega$$

* $a-b \rightarrow o.c \leftarrow V_{oc}$ الجهد Z_{out} له

$I_{s.c}$ الجهد $s.c$ له

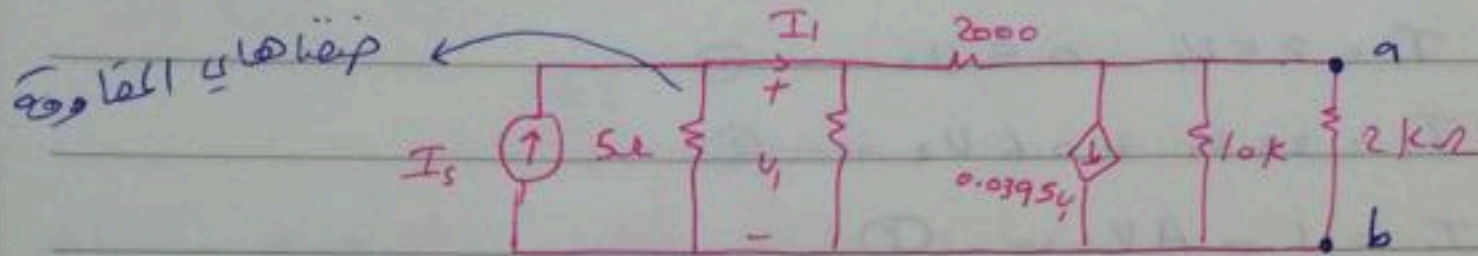
$$I_1 = 2.5V_1 - 0.5V_2$$

$$I_2 = 39V_1 + 0.6V_2$$

$$2.5V_1 - 0.5V_T = 0$$

$$39V_1 + 0.6V_T = I_T$$

$$V_T = \frac{\begin{bmatrix} 2.5 & 0 \\ 39 & I_T \end{bmatrix}}{\begin{bmatrix} 2.5 & -0.5 \\ 39 & 0.6 \end{bmatrix}} = \frac{2.5 I_T}{21} \quad \therefore \frac{V_T}{I_T} = 119 \Omega$$



$$I_1 = I_s - 0.2 V_1 \quad \text{--- (1)}$$

$$I_2 = -2 V_2 \quad \text{--- (2)}$$

$$2.7 V_1 - 0.5 V_2 = I_s \quad \text{--- (3)}$$

$$39 V_1 + 2.6 V_2 = 0 \quad \text{--- (4)}$$

$$V_1 = 0.09 I_s$$

* بعضنا قد يقول $I_1 = I_s$ بالافتقار الى اطلاع

$$V_2 = -1.47 I_s$$

معنى الجواب بتلاتة I_s

$$G_V = -1.47 / 0.09 I_s = -16.3$$

$$I_1 = 0.98 I_s$$

$$I_2 = 2.94 I_s$$

$$G_i = 3.1$$

$$G_p = -(-16.3)(3.1) = 50.53$$

$$Z_{in} = 0.09 I_s / 0.98 I_s = 0.091$$

$Z_{out} \Rightarrow$ remove the branch of a-b

$$I_1 = I_s - 0.2 U_1 \dots 1$$

for $Z_{out} \Rightarrow I_1 = -0.2 U_1$

يعود في المعادلات الى I_1

$$I_1 = 2.5 U_1 - 0.5 U_2$$

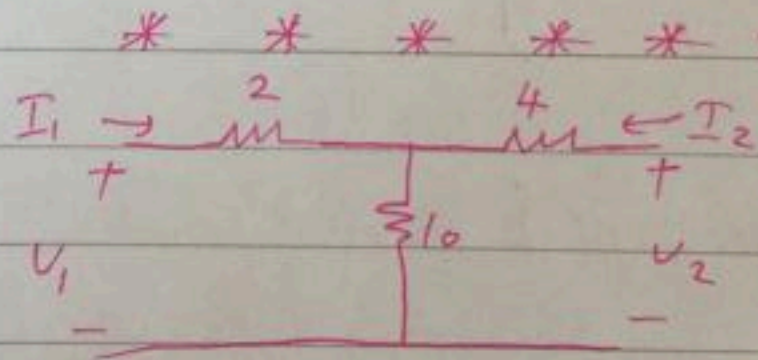
$$I_2 = 39 U_1 + 0.6 U_2$$

$$2.7 U_1 - 0.5 U_2 = 0$$

$$39 U_1 + 0.6 U_2 = I_T$$

$$Z_{out} = \left(\frac{352}{45} \right) = -1.47 I_s$$

تفسير $\Rightarrow \frac{2.7}{(524/25)}$



$$U_1 = A U_2 - B I_2$$

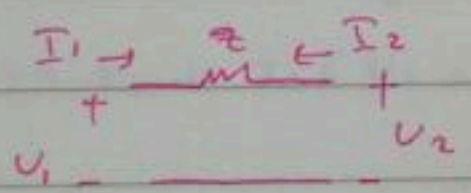
$$I_1 = C U_2 - D I_2$$

* o.c. II $\rightarrow A = U_1 / U_2$ * s.c. II $\rightarrow D = \frac{-I_1}{I_2}$
 $\rightarrow C = I_1 / I_2$ $\rightarrow B = -U_1 / I_2$

$$\begin{pmatrix} 1 & 2 \\ 0 & 1 \end{pmatrix} \begin{pmatrix} 1 & 0 \\ 1/10 & 1 \end{pmatrix} \begin{pmatrix} 1 & 4 \\ 0 & 1 \end{pmatrix}$$

$$\begin{pmatrix} 1.2 & 2 \\ 0.1 & 1 \end{pmatrix} \begin{pmatrix} 1 & 4 \\ 0 & 1 \end{pmatrix} = \begin{pmatrix} 1.2 & 6.8 \\ 0.1 & 1.4 \end{pmatrix}$$

$$\begin{pmatrix} V_1 \\ I_1 \end{pmatrix} = \begin{pmatrix} 1.2 & 6.8 \\ 0.1 & 1.4 \end{pmatrix} \begin{pmatrix} V_2 \\ -I_2 \end{pmatrix}$$



$$V_1 = V_2 = zI_2$$

$$I_1 = -I_2$$

$$T = \begin{pmatrix} 1 & z \\ 0 & 1 \end{pmatrix}$$

شروع

* في سؤال كبير بالفانيل على كل الكتيبات هار ← شوقو من الكتاب الى شرح التفصيلي

* شوقو علاقات ال Parameters مع بعضهم من الكتيبات