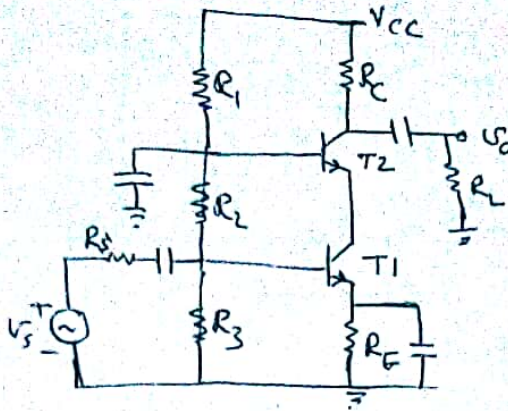


Q1) a) Draw the electronic circuit diagram of the cascode amplifier.



Write down its ac voltage gain .

$$A_V = -g_{m1} R_C \parallel R_L$$

3

1

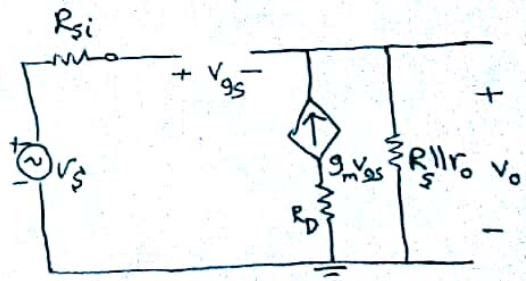
b) Draw the small signal low frequency circuit model of a common drain n-channel MOSFET , and use it to calculate the voltage gain when  $R_1 \parallel R_2 = \infty \Omega$  ,  $R_s$  ,  $R_{si}$  , and  $R_D$  are finite.

$$V_o = g_m V_{gs} (R_s \parallel r_o)$$

$$V_s = V_{gs} + g_m (R_s \parallel r_o) V_{gs}$$

$$\therefore V_{gs} = \frac{1}{1 + g_m (R_s \parallel r_o)} V_s$$

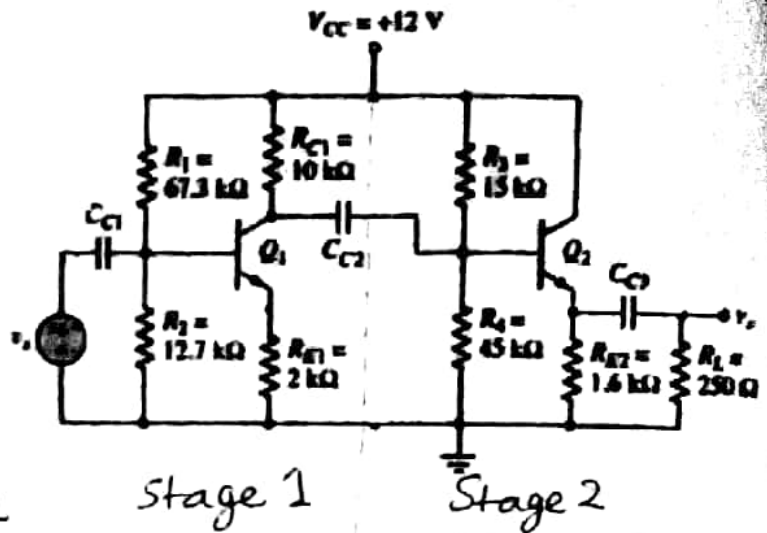
$$\therefore A_V = \frac{V_o}{V_s} = \frac{g_m (R_s \parallel r_o)}{1 + g_m (R_s \parallel r_o)}$$



4

Q2 Given the two-stage amplifier shown opposite with the following information:

$I_{BQ1} = 3.2 \mu A$ ,  
 $I_{BQ2} = 20 \mu A$ . Furthermore, for both transistors,  $\beta_{dc} = \beta_{ac} = \beta = 99$ ,  
 $V_T = 25 \text{ mV}$  and  $V_{CE} = \infty V$ .  
 Calculate the overall voltage gain.



$$r_{\pi 1} = \frac{V_T}{I_{BQ1}} = \frac{25 \text{ mV}}{3.2 \mu A} = 7.8125 \text{ k}\Omega$$

$$r_{\pi 2} = \frac{25 \text{ mV}}{20 \mu A} = 1.25 \text{ k}\Omega$$

$$r_o = \frac{V_A}{I_{CQ}} = \frac{\infty}{I_{CQ}} = \infty \Omega$$

Stage 1 is CE

$$R_{L1} = r_{\pi 1} + (1 + \beta) R_{E1} = 7.8125 + 100 \times 2 = 207.8125 \text{ k}\Omega$$

$$R_{i1} = R_1 \parallel R_2 \parallel R_{L1} = 67.3 \parallel 12.7 \parallel 207.8125 = 10.16 \text{ k}\Omega$$

$$A'_{V1} = -\beta \frac{R_{C1}}{R_{L1}} = -99 \frac{10}{207.8125} = -4.764$$

$$R_{o1} = R_{C1} = 10 \text{ k}\Omega$$

$$V_o = \frac{R_L}{R_L + R_{o2}} A'_{V2} \frac{R_{L2}}{R_{L2} + R_{o1}} A'_{V1} \frac{R_{i1}}{R_{i1} + R_s} V_s$$

$$\frac{V_o}{V_s} = \frac{0.25}{0.25 + 112} \times 0.99255 \frac{10.5}{10.5 + 10} \times -4.764 \frac{10.16}{10.16 + 0}$$

$$= \frac{0.25}{0.362} \times 0.99255 \frac{10.5}{20.5} \times -4.764 \times 1$$

$$= -1.6729$$

Stage 2 is CC

$$R_{L2} = r_{\pi 2} + (1 + \beta)(r_o \parallel R_E) = 1.25 + 160 = 161.25 \text{ k}\Omega$$

$$R_{i2} = R_1 \parallel R_2 \parallel R_{L2} = 15 \parallel 65 \parallel 161.25 = 10.5164 \text{ k}\Omega$$

$$A'_{V2} = \frac{(1 + \beta)(r_o \parallel R_E)}{R_{i2}}$$

$$= \frac{160}{161.25} = 0.99255$$

$$R_{o2} = \frac{r_{\pi 2} + R_{o1}}{1 + \beta}$$

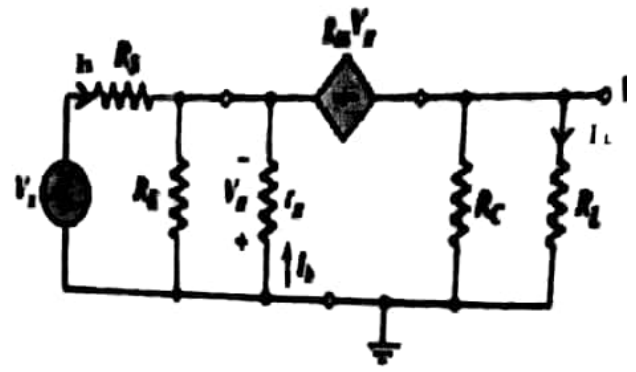
$$= \frac{11.25}{100} \text{ k}\Omega = 112.5 \Omega$$

③

Q3) a) Given the circuit shown.

i) Classify the circuit using the following words and abbreviations which fit the circuit:

pMOSFET, nMOSFET, pnp, npn, CE, CC, CB, CS, CD, CG, low frequency, high frequency, loaded, unloaded.



This circuit represent

a loaded, low frequency, npn, common base (CB) amplifier

(2)

ii) Derive the relationship which gives the current gain defined as  $A_i = \frac{I_L}{I_s}$ .

$$I_L = - \frac{R_C}{R_C + R_L} g_m V_{\pi}$$

$$-I_s = \frac{V_{\pi}}{R_E \parallel r_{\pi}} + g_m V_{\pi} = \left( \frac{1}{R_E \parallel r_{\pi}} + g_m \right) V_{\pi}$$

$$\therefore A_i = \frac{I_L}{I_s} = \frac{-I_L}{-I_s} = \frac{\frac{R_C}{R_C + R_L} g_m}{g_m + \frac{1}{R_E \parallel r_{\pi}}}$$

$$= \frac{R_C}{R_C + R_L} \frac{g_m (R_E \parallel r_{\pi})}{1 + g_m (R_E \parallel r_{\pi})} < 1$$

(3)

b) In an npn transistor amplifier circuit involving coupling and bypass capacitors, and operated at high frequency.

i) What capacitances account for the high frequency response?

the internal capacitances of the transistor:  $C_{\pi}$  and  $C_{\mu}$

ii) What capacitors account for the low frequency response?

(2)

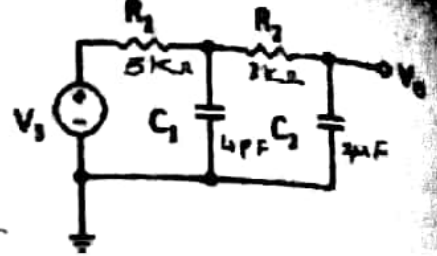
the coupling and bypass capacitors

7



Q4

a) consider the circuit shown, where  $R_1 = 5K\Omega$ ,  $R_2 = 3K\Omega$ ,  $C_1 = 4pF$ ,  $C_2 = 2\mu F$ . If the method of approximate time constants applies, use it to calculate  $f_L$  and  $f_H$ . Hence, or otherwise, calculate the midband gain.



Since the resistances  $R_1$  and  $R_2$  are of the same order and  $C_1$  and  $C_2$  are order of magnitude different, then the time constants are well far apart. Hence, the method applies.

$$\tau_L = (R_1 + R_2)C_2 = 8 \times 10^3 \times 2 \times 10^{-6} = 16 \text{ ms}$$

$$\Rightarrow f_L = \frac{1}{2\pi\tau_L} = \frac{10^3}{2\pi \times 16} = 9.947 \text{ Hz}$$

$$\tau_F = (R_1 || R_2)C_1 = (3k || 5k)C_1 = 1.875 \times 10^3 \times 4 \times 10^{-12} = 7.5 \times 10^{-9} \text{ s} = 7.5 \text{ ns}$$

$$\Rightarrow f_H = \frac{1}{2\pi\tau_F} = \frac{10^9}{2\pi \times 7.5} = \frac{1000}{15\pi} \text{ MHz} = 21.221 \text{ MHz}$$

Mid-band gain is obtained when  $Z_{C_1} = \infty \Omega$  &  $Z_{C_2} = 0 \Omega$ . Hence equals to zero volt.



b) A particular transistor short circuit current gain is given by  $A_i = \frac{0.05 - j 10^{-12}f}{0.0005 + j 10^{-11}f}$

Represent  $A_i$  in the form  $A_i = \frac{K(1 - j \frac{f}{f_z})}{(1 + j \frac{f}{f_p})}$

What are the values of  $K$ ,  $f_z$ , and  $f_p$ ,

$$A_i = \frac{0.05 (1 - j \frac{10^{-12}f}{0.05})}{0.0005 (1 + j \frac{10^{-11}f}{0.0005})}$$

$$= 100 \frac{1 - j \frac{f}{\frac{0.05}{10^{-12}}}}{1 + j \frac{f}{\frac{0.0005}{10^{-11}}}}$$

$$= 100 \frac{1 - j \frac{f}{50 \times 10^9}}{1 + j \frac{f}{50 \times 10^6}}$$

$$\Rightarrow K = 100$$

$$f_z = 50 \times 10^9 = 50 \text{ GHz}$$

$$f_p = 50 \times 10^6 = 50 \text{ MHz}$$

Draw the Bode diagram. Don't draw the phase plot.

