

25
30

1

University of Jordan

School of Engineering

Department of Electrical Engineering

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Digital Signal Processing , Second Exam



Number of questions : 4

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37

Q1. Consider the sequence $g[n]$ with DTFT $G(e^{j\omega})$. Making use of DTFT theorems, determine the inverse DTFTs of the following functions of $G(e^{j\omega})$ in terms of $g[n]$:

a) $X_1(e^{j\omega}) = e^{-j6\omega} G(e^{j\omega})$

$$X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x[n] e^{-j\omega n}$$

b) $X_2(e^{j\omega}) = G(e^{j(\omega+0.5\pi)})$

$$x[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\omega}) e^{j\omega n} d\omega.$$

c) $X_3(e^{j\omega}) = 3G(e^{j\omega}) + 4G(e^{-j\omega})$

d) $X_4(e^{j\omega}) = \frac{dG(e^{j\omega})}{d\omega} \rightarrow \frac{1}{j} \pi g[n].$

(a) $x[n] \xleftrightarrow{\text{DTFT}} X(e^{j\omega}).$
 $g[n-n_0] \xleftrightarrow{\text{DTFT}} e^{-jn_0\omega} G(e^{j\omega})$

~~no = 6.~~ time shift theorem
~~g[n-6] = x[n]~~

$$x[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} e^{-jn\omega} G(e^{j\omega}) e^{j\omega n} d\omega.$$

$$= \frac{1}{2\pi} \int_{-\pi}^{\pi} G(e^{j\omega}) e^{jn\omega - j(\omega - 6\omega)} d\omega.$$

(b) $e^{-jn\omega} g[n] \xleftrightarrow{\text{DTFT}} G(e^{j(\omega+0.5\pi)})$ frequency shift
 $x[n] = g[n] e^{-jn(0.5\pi)}$

(c) $a g[n] + b g[-n] \xleftrightarrow{\text{DTFT}} a G(e^{j\omega}) + b G(e^{-j\omega}).$ linear theorem

~~a = 3~~

~~b = 4.~~

~~g[-n].~~ $\xleftrightarrow{\text{DTFT}} G(e^{-j\omega})$

~~$\sum x[n]$~~ $x[n] = a g[n] + b g[-n].$

$x[n] = 3 g[n] + 4 g[-n].$

Q2. For the sequence $x[n] = \{-1, -3\}$,

7.5

- Find the D_2 matrix.
- Find $X[k]$, $k = 0, 1$. (The DFT of $x[n]$).
- Find the D_2^{-1} matrix.
- Find the IDFT of $X[k]$.

[a] $D_2 = \begin{bmatrix} 1 & 1 \\ 1 & w_2 \end{bmatrix} = \begin{bmatrix} 1 & 1 \\ 1 & -1 \end{bmatrix}$

$$W_2^{-1} = e^{-j\frac{2\pi k n}{N}} = e^{-j\frac{2\pi (1)(1)}{2}} = e^{-j\pi} = (-1)$$

[b] $X[k] = D_N x[n]$.

$$\begin{bmatrix} 1 & 1 \\ 1 & -1 \end{bmatrix} \begin{bmatrix} -1 \\ 3 \end{bmatrix} = \begin{bmatrix} 2 \\ -4 \end{bmatrix} \quad \begin{matrix} -1+3 \\ -1-3 \end{matrix}$$

[c] $D_2^{-1} = \frac{1}{N} D_2^* = \frac{1}{2} \begin{bmatrix} 1 & 1 \\ 1 & -1 \end{bmatrix} = \begin{bmatrix} 1/2 & 1/2 \\ 1/2 & -1/2 \end{bmatrix}$

[d] $x[n] = D_2^{-1} X[k]$

$$\begin{bmatrix} 1/2 & 1/2 \\ 1/2 & -1/2 \end{bmatrix} \begin{bmatrix} 2 \\ -4 \end{bmatrix} = \begin{bmatrix} -1 \\ 3 \end{bmatrix} \quad \begin{matrix} \frac{1}{2} \times 2 + \frac{1}{2} \times -4 = 1 - 2 = 1 \\ \frac{1}{2} \times 2 + -\frac{1}{2} \times -4 = 1 + 2 = 3 \end{matrix}$$

RSS .

Q3. The transfer function of a causal LTI discrete-time system is given by

$$H(z) = \frac{0.5 - 0.3z^{-1}}{1 - 0.6z^{-1}}$$

$$\frac{10}{5} = 5 \sqrt{10}$$

a) Determine the impulse response $h[n]$ of the system.b) Determine the output of the system $y[n]$ if the input $x[n] = (0.2)^n u[n]$.

(a)

$$\frac{0.5}{1 - 0.6z^{-1}} \quad \boxed{\begin{array}{c} 0.5 \\ 0.5 - 0.3z^{-1} \\ 0.5 - 0.3z^{-1} \end{array} \ominus} \quad 7.5$$

$$H(z) = 0.5 + \frac{0}{1 - 0.6z^{-1}} = 0.5,$$

$$h[n] = 0.5 \delta[n].$$

(b)

$$y[n] = h[n]x[n], \quad X(z) = \frac{1}{1 - 0.2z^{-1}}$$

$$y(z) = H(z)X(z) = \left(\frac{0.5 - 0.3z^{-1}}{1 - 0.6z^{-1}} \right) \cdot \left(\frac{1}{1 - 0.2z^{-1}} \right).$$

$$Y(z) = \frac{0.5 - 0.3z^{-1}}{(1 - 0.6z^{-1})(1 - 0.2z^{-1})}$$

$$Y(z) = \frac{A_1}{(1 - 0.6z^{-1})} + \frac{A_2}{(1 - 0.2z^{-1})}$$

$$\boxed{z^{-1} = \frac{10}{6}} \quad A_1 \Big|_{z=0.6} = \frac{0.5 - 0.3(\frac{10}{6})}{(1 - 0.2 \cdot \frac{10}{6})} = 0$$

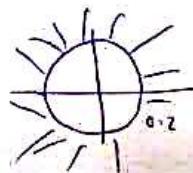
$$\boxed{z^{-1} = \frac{10}{2}} \quad A_2 \Big|_{z=0.2} = \frac{0.5 - 0.3(\frac{10}{2})}{(1 - 0.6 \cdot \frac{10}{2})} = \frac{-1}{-2} = 0.5$$

$$Y(z) = \frac{0}{1 - 0.6z^{-1}} + \frac{0.5}{1 - 0.2z^{-1}} = \frac{0.5}{1 - 0.2z^{-1}}$$

RSS $|z| > 1.01$
 $|z| > 0.2$

(b)

$$y[n] = 0.5(0.2)^n M[n].$$



~~Q4~~

$$X(z) = \sum_{n=-\infty}^{\infty} x[n] z^{-n}$$

$$x[n] = \frac{1}{N} \sum_{z=-\infty}^{\infty} X(z) z^n.$$

4

Q4. The continuous-time signal $x_a(t) = 2 \cos(10\pi t) + 3 \sin(24\pi t)$ is sampled at a 40 Hz rate. The generated sequence $x[n]$ is an input to a system with a transfer function

$$H(z) = (z - r_1)(z - r_1^*)(z - r_2)(z - r_2^*) \rightarrow \text{4}$$

Find r_1 and r_2 such that the output of the system $y[n]$ is equal zero.

$$H(z) = \frac{Y(z)}{X(z)}, \quad H[n] = \frac{y[n]}{x[n]}.$$

$$H[n] =$$

~~X(z)~~ = ~~H(z)~~

$$H(z) \cdot X(z) = y(z)$$

$$= 0.$$

$$X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x[n] e^{-j\omega n}$$

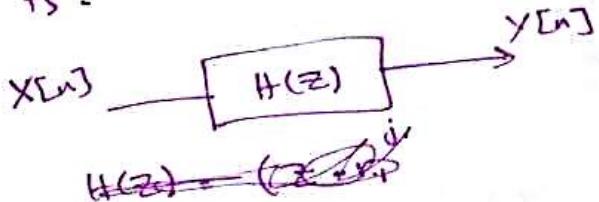
$$H(e^{j\omega}) = H(z) \Big|_{z=e^{j\omega}} = (z - r_1)(z - r_1^*)(z - r_2)(z - r_2^*) \quad z = e^{j\omega}$$

$$\text{At } \omega = 10\pi: \quad 2 \left[e^{+j10\pi} + e^{-j10\pi} \right] + 3 \left[\frac{e^{+j10\pi} - e^{-j10\pi}}{2j} \right] = e^{j10\pi} + e^{-j10\pi} + \left[\frac{3}{2j} e^{j10\pi} - \frac{3}{2j} e^{-j10\pi} \right]$$

Q4: Sol:

$$x_a(t) = 2 \cos(10\pi t) + 3 \sin(24\pi t).$$

$$f_s = 40 \text{ Hz}.$$



$$\omega_1 = 10\pi \cdot \frac{1}{40} = \frac{\pi}{4} = 0.25\pi \text{ rad/sample.}$$

$$\omega_2 = 24\pi \cdot \frac{1}{40} = 0.6\pi \text{ rad/sample}$$

$$r_1 = e^{j\omega_1} = e^{j0.25\pi}$$

$$r_2 = e^{j\omega_2} = e^{j0.6\pi}.$$

$$H(z) = (z - r_1^*) (z - r_1) (z - r_2^*) (z - r_2)$$
$$= (z - r_1^*) e^{j(0.25)} (z - r_1) (z - r_2^*) e^{j(0.6)}$$